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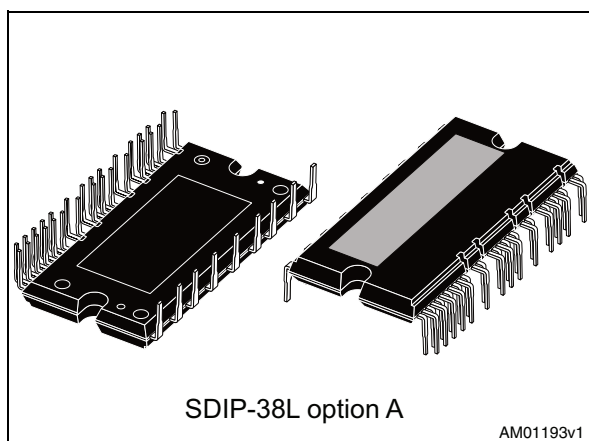
sales@integrated-circuit.com



STGIPL14K60, STGIPL14K60-S

SLLIMM™ (small low-loss intelligent molded module)
IPM, 3-phase inverter, 15 A, 600 V short-circuit rugged IGBT

Datasheet - production data



- 5 kΩ NTC for temperature control
- UL Recognized: UL1557 file E81734

Applications

- 3-phase inverters for motor drives
- Home appliances, such as washing machines, refrigerators, air conditioners and sewing machines

Description

These intelligent power modules provides a compact, high performance AC motor drive in a simple, rugged design. Combining ST proprietary control ICs with the most advanced short-circuit-rugged IGBT system technology, this device is ideal for 3-phase inverters in applications such as home appliances and air conditioners. SLLIMM™ is a trademark of STMicroelectronics.

Features

- IPM 15 A, 600 V 3-phase IGBT inverter bridge including control ICs for gate driving and free-wheeling diodes
- Short-circuit rugged IGBTs
- $V_{CE(sat)}$ negative temperature coefficient
- 3.3 V, 5 V, 15 V CMOS/TTL inputs comparators with hysteresis and pull down/pull up resistors
- Undervoltage lockout
- Internal bootstrap diode
- Interlocking function
- Smart shut down function
- Comparators for fault protection against overtemperature and overcurrent
- Op amps for advanced current sensing
- DBC substrate leading to low thermal resistance
- Isolation rating of 2500 V_{rms}/min

Table 1. Device summary

Order code	Marking	Package	Packaging
STGIPL14K60	GIPL14K60	SDIP-38L option A	Tube
STGIPL14K60-S	GIPL14K60-S	SDIP-38L option B	Tube

Contents

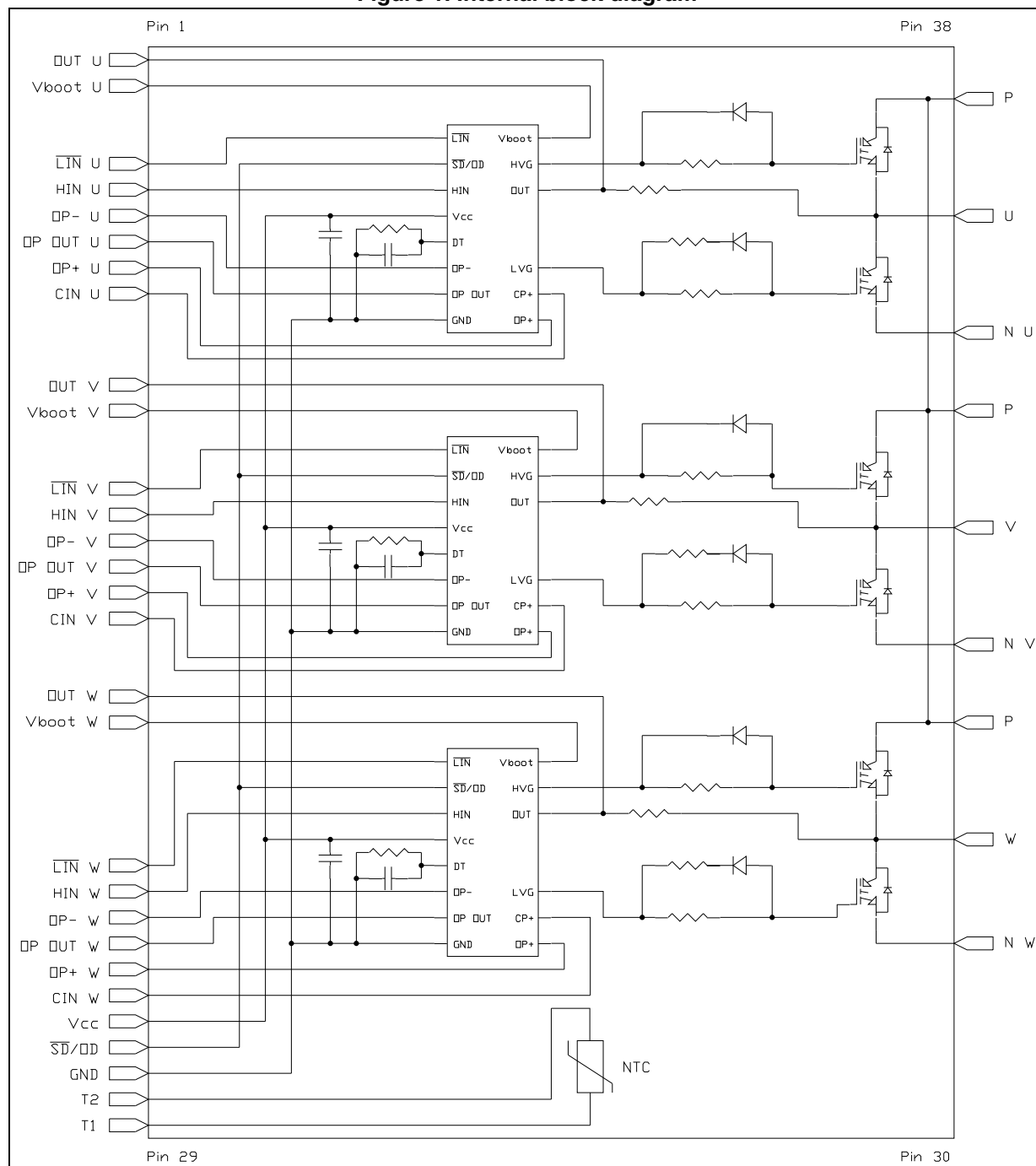
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Internal block diagram and pin configuration

1 Internal block diagram and pin configuration

Figure 1. Internal block diagram



Internal block diagram and pin configuration

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Table 2. Pin description

Pin	Symbol	Description
1	OUT _U	High side reference output for U phase
2	V _{boot U}	Bootstrap voltage for U phase
3	$\overline{\text{LIN}}_{\text{U}}$	Low side logic input for U phase
4	HIN _U	High side logic input for U phase
5	OP _{-U}	Op amp inverting input for U phase
6	OP _{OUT U}	Op amp output for U phase
7	OP _{+U}	Op amp non inverting input for U phase
8	CIN _U	Comparator input for U phase
9	OUT _V	High side reference output for V phase
10	V _{boot V}	Bootstrap voltage for V phase
11	$\overline{\text{LIN}}_{\text{V}}$	Low side logic input for V phase
12	HIN _V	High side logic input for V phase
13	OP _{-V}	Op amp inverting input for V phase
14	OP _{OUT V}	Op amp output for V phase
15	OP _{+V}	Op amp non inverting input for V phase
16	CIN _V	Comparator input for V phase
17	OUT _W	High side reference output for W phase
18	V _{boot W}	Bootstrap voltage for W phase
19	$\overline{\text{LIN}}_{\text{W}}$	Low side logic input for W phase
20	HIN _W	High side logic input for W phase
21	OP _{-W}	Op amp inverting input for W phase
22	OP _{OUT W}	Op amp output for W phase
23	OP _{+W}	Op amp non inverting input for W phase
24	CIN _W	Comparator input for W phase
25	V _{CC}	Low voltage power supply
26	$\overline{\text{SD}} / \text{OD}$	Shut down logic input (active low) / open drain (comparator output)
27	GND	Ground
28	T ₂	NTC thermistor terminal 2
29	T ₁	NTC thermistor terminal 1
30	N _W	Negative DC input for W phase
31	W	W phase output
32	P	Positive DC input
33	N _V	Negative DC input for V phase
34	V	V phase output

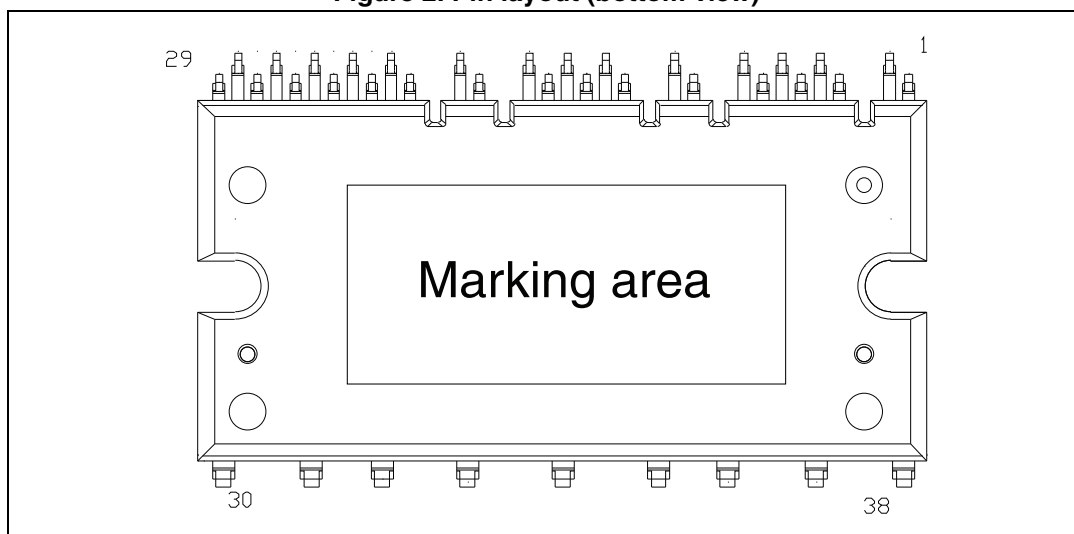
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Internal block diagram and pin configuration

Table 2. Pin description (continued)

Pin	Symbol	Description
35	P	Positive DC input
36	N _U	Negative DC input for U phase
37	U	U phase output
38	P	Positive DC input

Figure 2. Pin layout (bottom view)



Electrical ratings

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2 Electrical ratings

2.1 Absolute maximum ratings

Table 3. Inverter part

Symbol	Parameter	Value	Unit
V_{PN}	Supply voltage applied between P- N_U , N_V , N_W	450	V
$V_{PN(surge)}$	Supply voltage (surge) applied between P- N_U , N_V , N_W	500	V
V_{CES}	Each IGBT collector emitter voltage ($V_{IN}^{(1)} = 0$)	600	V
$\pm I_C^{(2)}$	Each IGBT continuous collector current at $T_C = 25^\circ\text{C}$	15	A
$\pm I_{CP}^{(3)}$	Each IGBT pulsed collector current	30	A
P_{TOT}	Each IGBT total dissipation at $T_C = 25^\circ\text{C}$	44	W
t_{scw}	Short circuit withstand time, $V_{CE} = 0.5 V_{(BR)CES}$ $T_j = 125^\circ\text{C}$, $V_{CC} = V_{boot} = 15\text{ V}$, $V_{IN(1)} = 0 \div 5\text{ V}$	5	μs

1. Applied between HIN_i , $\overline{LIN_i}$ and GND for $i = U, V, W$

2. Calculated according to the iterative formula:

$$I_C(T_C) = \frac{T_{j(max)} - T_C}{R_{thj-c} \times V_{CE(sat)(max)}(T_{j(max)}, I_C(T_C))}$$

3. Pulse width limited by max junction temperature

Table 4. Control part

Symbol	Parameter	Min.	Max.	Unit
V_{OUT}	Output voltage applied between OUT_U , OUT_V , OUT_W - GND	$V_{boot} - 21$	$V_{boot} + 0.3$	V
V_{CC}	Low voltage power supply	- 0.3	21	V
V_{CIN}	Comparator input voltage	- 0.3	$V_{CC} + 0.3$	V
V_{op+}	OPAMP non-inverting input	- 0.3	$V_{CC} + 0.3$	V
V_{op-}	OPAMP inverting input	- 0.3	$V_{CC} + 0.3$	V
V_{boot}	Bootstrap voltage	- 0.3	620	V
V_{IN}	Logic input voltage applied between HIN , $\overline{LIN}$ and GND	- 0.3	15	V
$V_{SD/OD}$	Open drain voltage	- 0.3	15	V
dV_{OUT}/dt	Allowed output slew rate		50	V/ns

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Electrical ratings

Table 5. Total system

Symbol	Parameter	Value	Unit
V_{ISO}	Isolation withstand voltage applied between each pin and heatsink plate (AC voltage, $t = 60\text{sec.}$)	2500	V
T_j	Power chips operating junction temperature	-40 to 150	°C
T_C	Module case operation temperature	-40 to 125	°C

2.2 Thermal data

Table 6. Thermal data

Symbol	Parameter	Value	Unit
$R_{th(j-c)}$	Thermal resistance junction-case single IGBT	2.8	°C/W
	Thermal resistance junction-case single diode	5	°C/W

Electrical characteristics

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3 Electrical characteristics

$T_j = 25\text{ }^{\circ}\text{C}$ unless otherwise specified.

Table 7. Inverter part

Symbol	Parameter	Test condition	Value			Unit
			Min.	Typ.	Max.	
$V_{CE(sat)}$	Collector-emitter saturation voltage	$V_{CC} = V_{Boot} = 15\text{ V}$, $V_{IN}^{(1)} = 0 \div 5\text{ V}$, $I_C = 7\text{ A}$	-	2.1	2.5	V
		$V_{CC} = V_{Boot} = 15\text{ V}$, $V_{IN}^{(1)} = 0 \div 5\text{ V}$, $I_C = 7\text{ A}$, $T_j = 125\text{ }^{\circ}\text{C}$	-	1.8		
I_{CES}	Collector-cut off current ($V_{IN}^{(1)} = 0$ "logic state")	$V_{CE} = 550\text{ V}$ $V_{CC} = V_{boot} = 15\text{ V}$	-		150	μA
V_F	Diode forward voltage	$V_{IN}^{(1)} = 0$ "logic state", $I_C = 7\text{ A}$	-		2.1	V
Inductive load switching time and energy						
t_{on}	Turn-on time	$V_{DD} = 300\text{ V}$, $V_{CC} = V_{boot} = 15\text{ V}$, $V_{IN}^{(1)} = 0 \div 5\text{ V}$, $I_C = 7\text{ A}$ (see Figure 3)	-	270		ns
$t_{c(on)}$	Crossover time (on)		-	130		
t_{off}	Turn-off time		-	320		
$t_{c(off)}$	Crossover time (off)		-	110		
t_{rr}	Reverse recovery time		-	130		
E_{on}	Turn-on switching losses		-	150		μJ
E_{off}	Turn-off switching losses		-	90		

1. Applied between $HINi$, $\overline{LINi}$ and GND for $i = U, V, W$ ($\overline{LIN}$ inputs are active-low).

Note: t_{on} and t_{off} include the propagation delay time of the internal drive. $t_{C(ON)}$ and $t_{C(OFF)}$ are the switching time of IGBT itself under the internally given gate driving condition.

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Electrical characteristics

Figure 3. Switching time test circuit

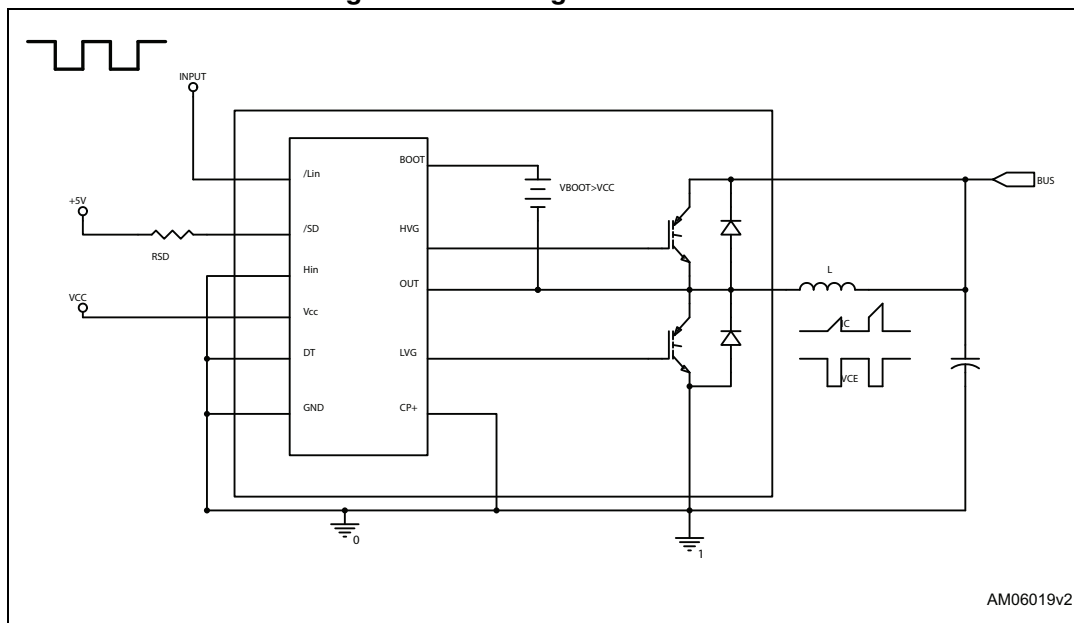


Figure 4. Switching time definition

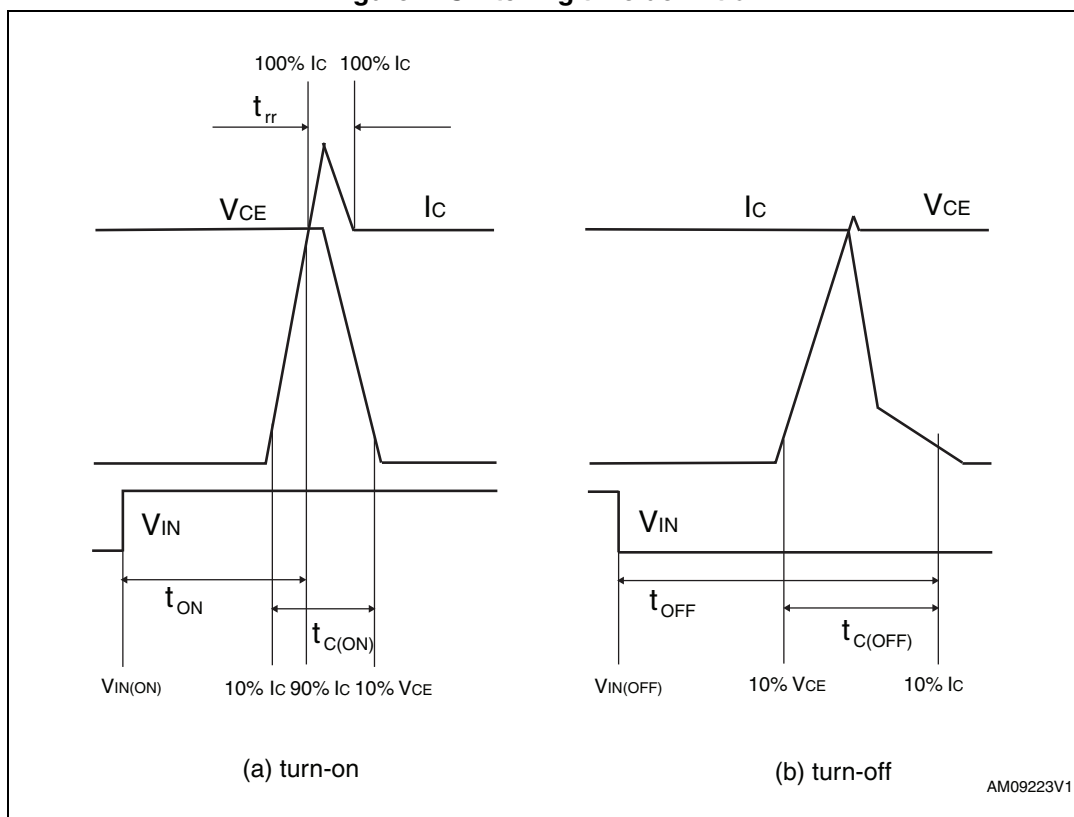


Figure 4 "Switching time definition" refers to $\overline{\text{HIN}}$ inputs (active high). For $\overline{\text{LIN}}$ inputs (active low), V_{IN} polarity must be inverted for turn-on and turn-off.

Electrical characteristics

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3.1 Control part

Table 8. Low voltage power supply ($V_{CC} = 15\text{ V}$ unless otherwise specified)

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
V_{CC_hys}	V_{CC} UV hysteresis		1.2	1.5	1.8	V
V_{CC_thON}	V_{CC} UV turn ON threshold		11.5	12	12.5	V
V_{CC_thOFF}	V_{CC} UV turn OFF threshold		10	10.5	11	V
I_{qccu}	Undervoltage quiescent supply current	$V_{CC} = 10\text{ V}$ $\overline{SD}/\overline{OD} = 5\text{ V}$; $\overline{LIN} = 5\text{ V}$; $H_{IN} = 0$, $C_{IN} = 0$			450	μA
I_{qcc}	Quiescent current	$V_{CC} = 15\text{ V}$ $\overline{SD}/\overline{OD} = 5\text{ V}$; $\overline{LIN} = 5\text{ V}$ $H_{IN} = 0$, $C_{IN} = 0$			3.5	mA
V_{ref}	Internal comparator (CIN) reference voltage		0.5	0.54	0.58	V

Table 9. Bootstrapped voltage ($V_{CC} = 15\text{ V}$ unless otherwise specified)

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
V_{BS_hys}	V_{BS} UV hysteresis		1.2	1.5	1.8	V
V_{BS_thON}	V_{BS} UV turn ON threshold		11.1	11.5	12.1	V
V_{BS_thOFF}	V_{BS} UV turn OFF threshold		9.8	10	10.6	V
I_{QBSU}	Undervoltage V_{BS} quiescent current	$V_{BS} < 9\text{ V}$ $\overline{SD}/\overline{OD} = 5\text{ V}$; $\overline{LIN}$ and $H_{IN} = 5\text{ V}$; $C_{IN} = 0$		70	110	μA
I_{QBS}	V_{BS} quiescent current	$V_{BS} = 15\text{ V}$ $\overline{SD}/\overline{OD} = 5\text{ V}$; $\overline{LIN}$ and $H_{IN} = 5\text{ V}$; $C_{IN} = 0$		200	300	μA
$R_{DS(on)}$	Bootstrap driver on resistance	LVG ON		120		W

Table 10. Logic inputs ($V_{CC} = 15\text{ V}$ unless otherwise specified)

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
V_{il}	Low logic level voltage		0.8		1.1	V
V_{ih}	High logic level voltage		1.9		2.25	V
I_{HINh}	HIN logic "1" input bias current	$H_{IN} = 15\text{ V}$	110	175	260	μA
I_{HINl}	HIN logic "0" input bias current	$H_{IN} = 0\text{ V}$			1	μA
I_{LINl}	$\overline{LIN}$ logic "1" input bias current	$\overline{LIN} = 0\text{ V}$	3	6	20	μA
I_{LINh}	$\overline{LIN}$ logic "0" input bias current	$\overline{LIN} = 15\text{ V}$			1	μA
I_{SDh}	$\overline{SD}$ logic "0" input bias current	$\overline{SD} = 15\text{ V}$	30	120	300	μA

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Electrical characteristics

Table 10. Logic inputs ($V_{CC} = 15\text{ V}$ unless otherwise specified) (continued)

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{SDI}	$\overline{SD}$ logic "1" input bias current	$\overline{SD} = 0\text{ V}$			3	μA
Dt	Dead time	see Figure 9		600		ns

Table 11. OPAMP characteristics ($V_{CC} = 15\text{ V}$ unless otherwise specified)

Symbol	Parameter	Test condition	Min.	Typ.	Max.	Unit
V_{io}	Input offset voltage	$V_{ic} = 0\text{ V}$, $V_o = 7.5\text{ V}$			6	mV
I_{io}	Input offset current	$V_{ic} = 0\text{ V}$, $V_o = 7.5\text{ V}$		4	40	nA
I_{ib}	Input bias current ⁽¹⁾			100	200	nA
ζ_{icm}	Input common mode voltage range		0			V
ζ_{OL}	Low level output voltage	$R_L = 10\text{ k}\Omega$ to V_{CC}		75	150	mV
ζ_{OH}	High level output voltage	$R_L = 10\text{ k}\Omega$ to GND	14	14.7		V
I_o	Output short circuit current	$\Sigma\text{upply}\epsilon$, $\zeta_{id} = +1$; $\zeta_o = 0\text{ }\zeta$	16	30		mA
		Sink, $\zeta_{id} = -1$; $\zeta_o = \zeta_{XX}$	50	80		mA
SR	Slew rate	$V_i = 1 \div 4\text{ V}$; $C_L = 100\text{ pF}$; unity gain	2.5	3.8		V/ μs
GBWP	Gain bandwidth product	$V_o = 7.5\text{ V}$	8	12		MHz
A_{vd}	Large signal voltage gain	$R_L = 2\text{ k}\Omega$	70	85		dB
SVR	Supply voltage rejection ratio	vs. V_{CC}	60	75		dB
CMRR	Common mode rejection ratio		55	70		dB

1. The direction of input current is out of the IC.

Table 12. Sense comparator characteristics ($V_{CC} = 15\text{ V}$ unless otherwise specified)

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
$I_{ib(i)}$	Input bias current	$V_{CIN(i)} = 1\text{ V}$, $i = U, V \text{ o } W$	-		3	μA
V_{ol}	Open-drain low-level output voltage	$I_{od} = 3\text{ mA}$	-		0.5	V
t_{d_comp}	Comparator delay	$\overline{SD}/OD$ pulled to 5 V through 100 k Ω resistor	-	90	130	ns
SR	Slew rate	$C_L = 180\text{ pF}$; $R_{pu} = 5\text{ k}\Omega$	-	60		V/ μsec
t_{sd}	Shut down to high / low side driver propagation delay	$V_{OUT} = 0$, $V_{boot} = V_{CC}$, $V_{IN} = 0$ to 3.3 V	50	125	200	ns
t_{isd}	Comparator triggering to high / low side driver turn-off propagation delay	Measured applying a voltage step from 0 V to 3.3 V to pin CIN_i	50	200	250	

Electrical characteristics

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Table 13. Truth table

Condition	Logic input (V_I)			Output	
	$\overline{SD/OD}$	LIN	HIN	LVG	HVG
Shutdown enable half-bridge 3-state	L	X	X	L	L
Interlocking half-bridge 3-state	H	L	H	L	L
0 "logic state" half-bridge 3-state	H	H	L	L	L
1 "logic state" low side direct driving	H	L	L	H	L
1 "logic state" high side direct driving	H	H	H	L	H

Note: X: don't care

Figure 5. Maximum $I_{C(RMS)}$ current vs. switching frequency ⁽¹⁾

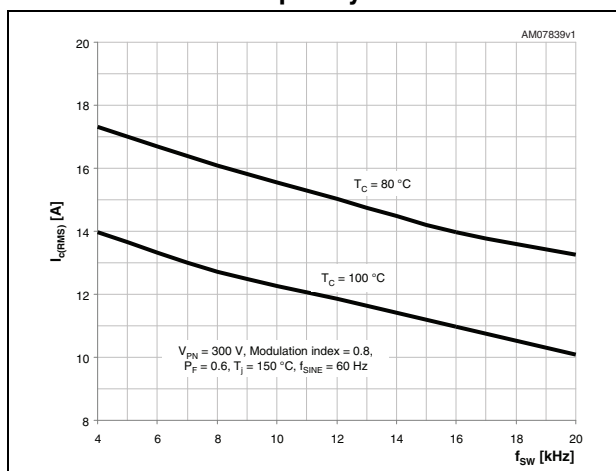
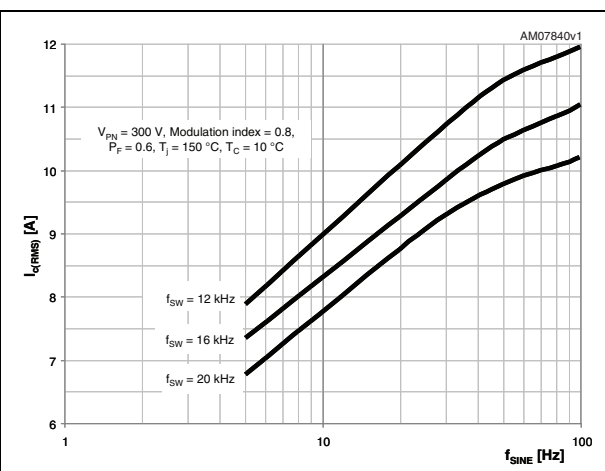


Figure 6. Maximum $I_{C(RMS)}$ current vs. f_{sine} ⁽¹⁾



1. Simulated curves refer to typical IGBT parameters and maximum R_{thj-c} .

3.1.1 NTC thermistor

Table 14. NTC thermistor

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit.
R_{25}	Resistance	$T = 25^{\circ}\text{C}$		5		$k\Omega$
R_{125}	Resistance	$T = 125^{\circ}\text{C}$		300		Ω
B	B-constant	$T = 25^{\circ}\text{C}$ to 85°C		3340		K
T	Operating temperature		-40		125	$^{\circ}\text{C}$

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Electrical characteristics

Equation 1: resistance variation vs. temperature

$$R(T) = R_{25} \cdot e^{B \left(\frac{1}{T} - \frac{1}{298} \right)}$$

Where T are temperatures in Kelvin.

Figure 7. NTC resistance vs. temperature

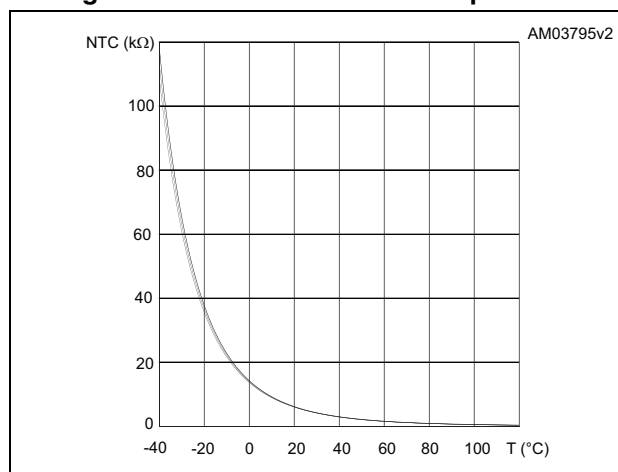
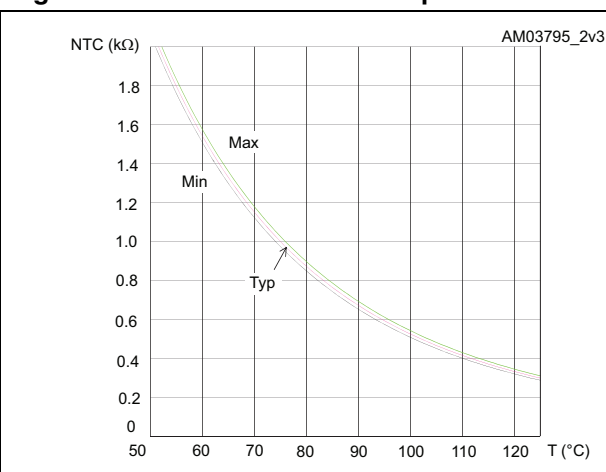


Figure 8. NTC resistance vs. temperature zoom

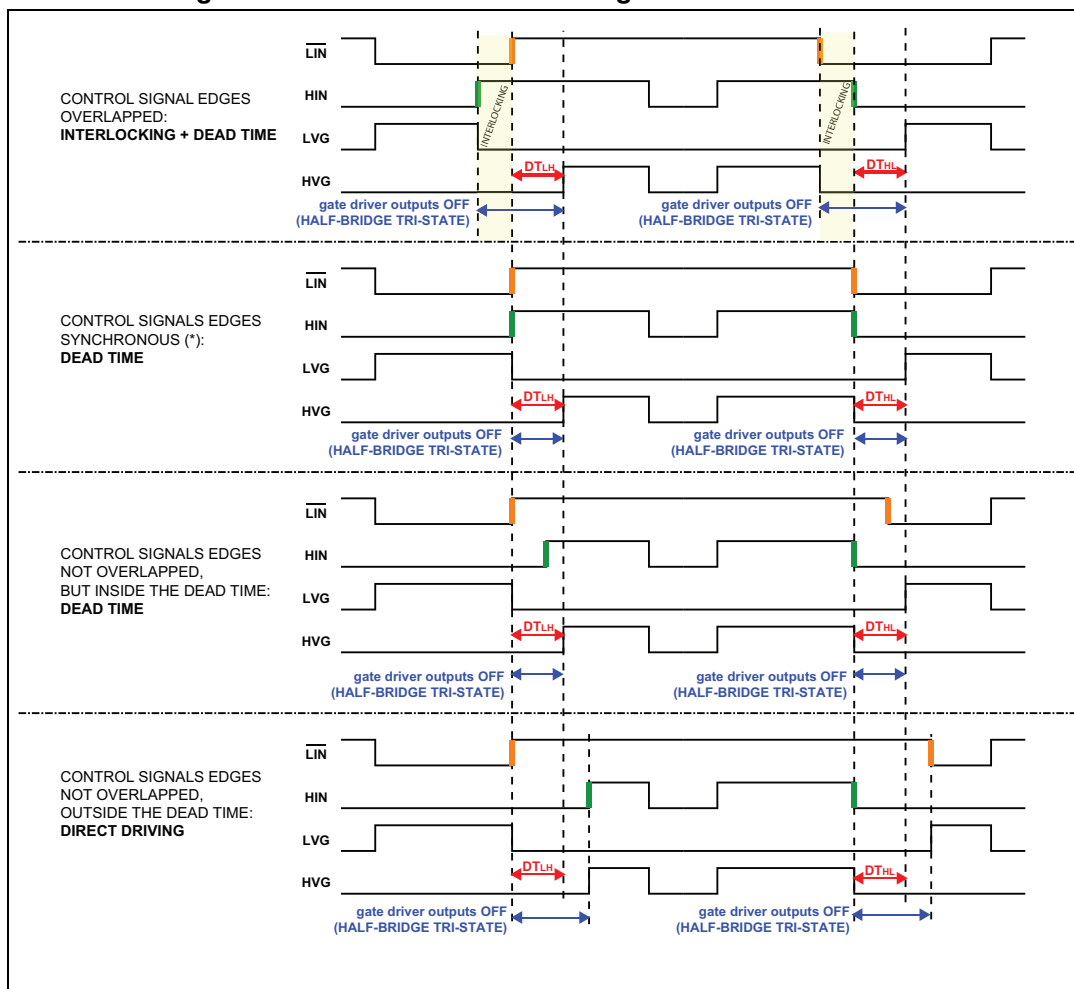


Electrical characteristics

STGIPL14K60, STGIPL14K60-S

3.2 Waveforms definitions

Figure 9. Dead time and interlocking waveforms definitions



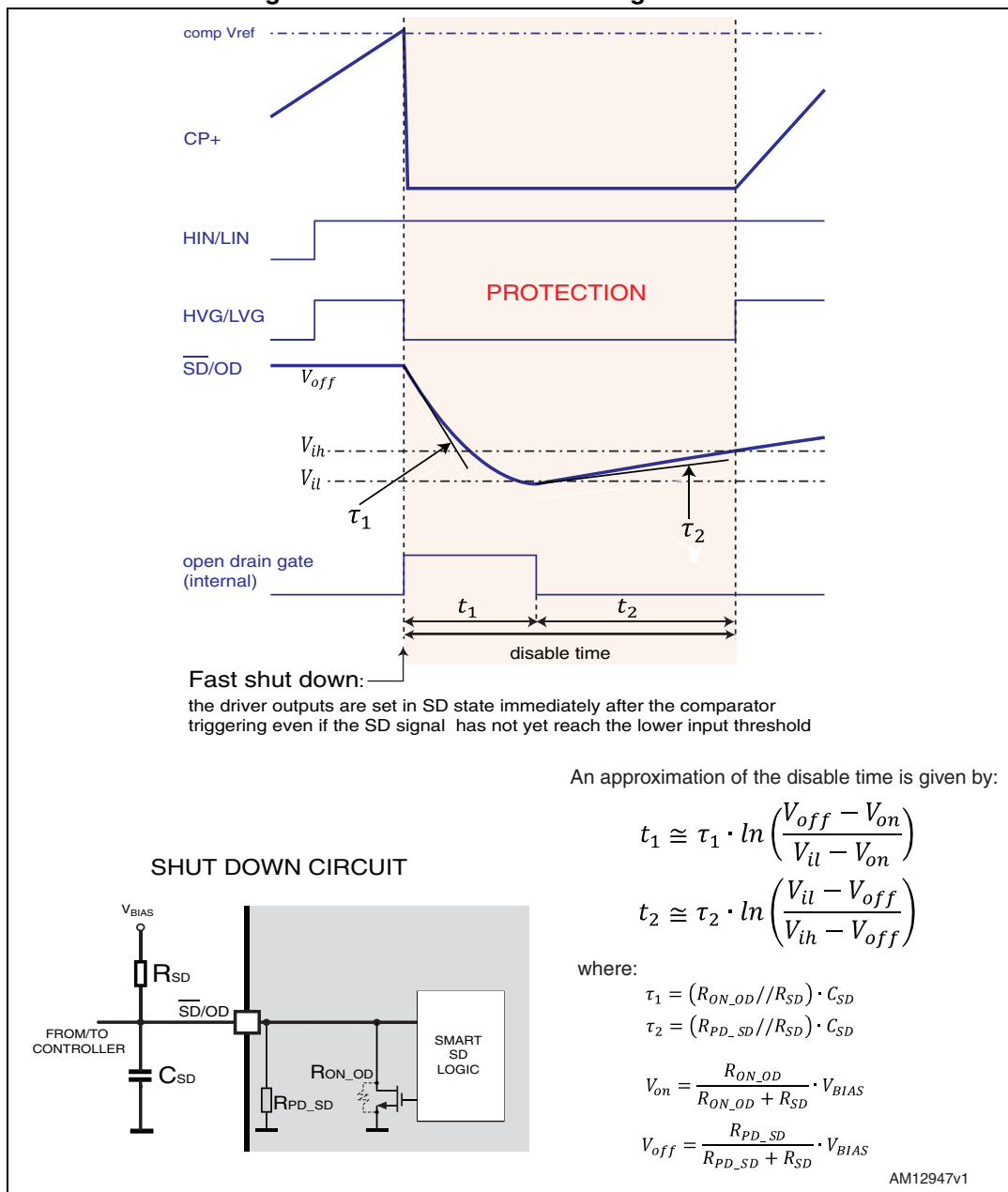
4 Smart shutdown function

The devices integrate a comparator for fault sensing purposes. The comparator has an internal voltage reference V_{REF} connected to the inverting input, while the non-inverting input, available on pin (CIN), can be connected to an external shunt resistor in order to implement a simple over-current protection function. When the comparator triggers, the device is set in shutdown state and both its outputs are set to low-level leading the halfbridge in tri-state. In the common overcurrent protection architectures the comparator output is usually connected to the shutdown input through a RC network, in order to provide a mono-stable circuit, which implements a protection time that follows the fault condition. Our smart shutdown architecture allows to immediately turn-off the output gate driver in case of overcurrent, the fault signal has a preferential path which directly switches off the outputs. The time delay between the fault and the outputs turn-off is no more dependent on the RC values of the external network connected to the shutdown pin. At the same time the DMOS connected to the open-drain output (pin SD/OD) is turned on by the internal logic which holds it on until the shutdown voltage is lower than the logic input lower threshold (V_{il}). Finally, the smart shutdown function provides the possibility to increase the real disable time without increasing the constant time of the external RC network.

Smart shutdown function

STGIPL14K60, STGIPL14K60-S

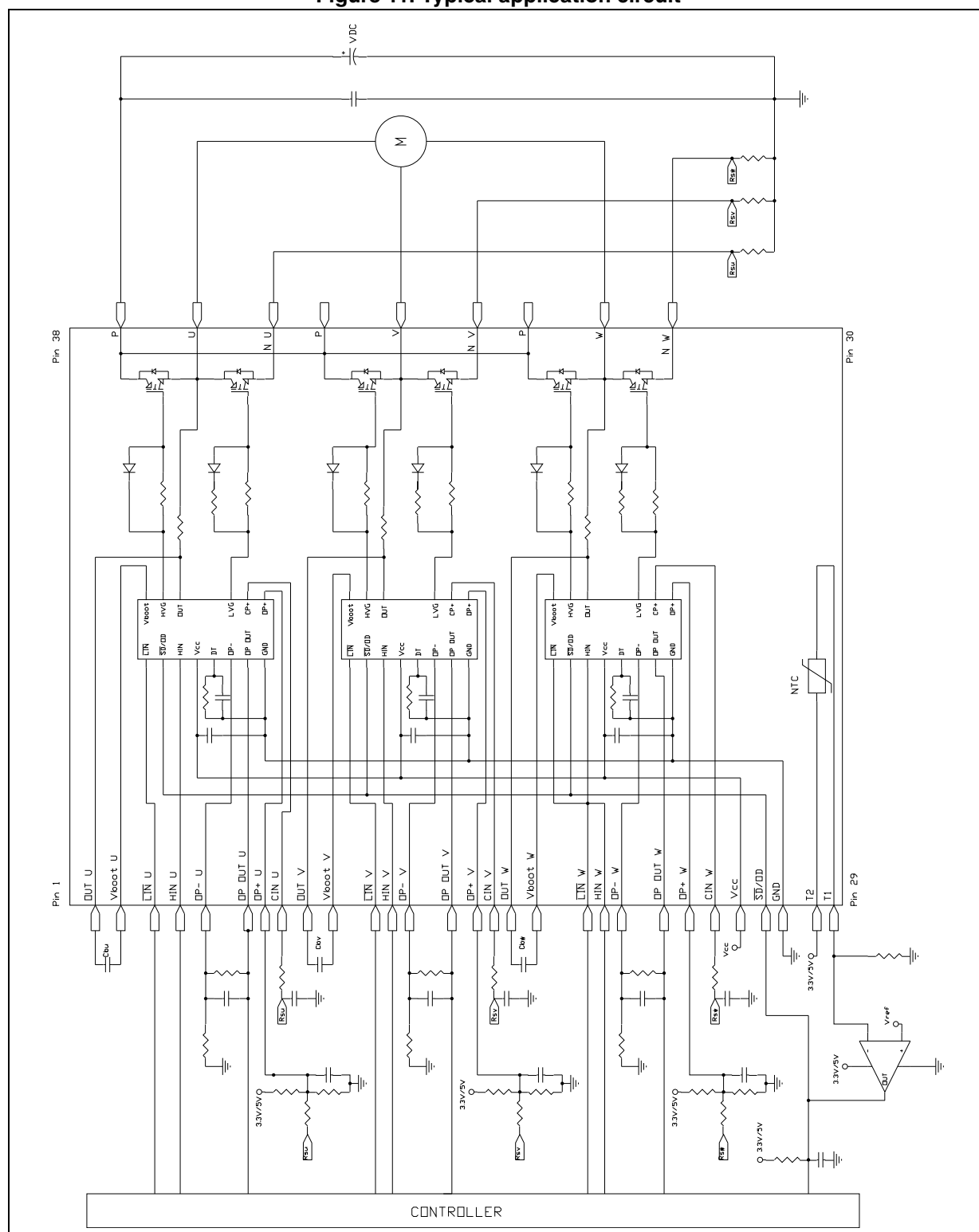
Figure 10. Smart shutdown timing waveforms



Please refer to [Table 12](#) for internal propagation delay time details.

5 Applications information

Figure 11. Typical application circuit



Applications information

STGIPL14K60, STGIPL14K60-S

5.1 Recommendations

- Input signal HIN is active high logic. A 85k Ω (typ.) pull down resistor is built-in for each high side input. If an external RC filter is used, for noise immunity, pay attention to the variation of the input signal level.
- Input signal $\overline{\text{LIN}}$ is active low logic. A 720 k Ω (typ.) pull-up resistor, connected to an internal 5 V regulator through a diode, is built-in for each low side input.
- To prevent the input signals oscillation, the wiring of each input should be as short as possible.
- By integrating an application specific type HVIC inside the module, direct coupling to MCU terminals without any opto-coupler is possible.
- Each capacitor should be located as nearby the pins of IPM as possible.
- Low inductance shunt resistors should be used for phase leg current sensing.
- Electrolytic bus capacitors should be mounted as close to the module bus terminals as possible. Additional high frequency ceramic capacitor mounted close to the module pins will further improve performance.
- The $\overline{\text{SD}}/\text{OD}$ signal should be pulled up to 5 V / 3.3 V with an external resistor (see [Section 4: Smart shutdown function](#) for detailed info).

Table 15. Recommended operating conditions

Symbol	Parameter	Conditions	Value			Unit
			Min.	Typ.	Max.	
V _{PN}	Supply Voltage	Applied between P-Nu, Nv, Nw		300	400	V
V _{CC}	Control supply voltage	Applied between V _{CC} -GND	13.5	15	18	V
V _{BS}	High side bias voltage	Applied between V _{BOOTi} -OUT _i for i = U, V, W	13		18	V
t _{dead}	Blanking time to prevent Arm-short	For each input signal	1			μs
f _{PWM}	PWM input signal	-40°C < T _c < 100°C -40°C < T _j < 125°C			20	kHz
T _C	Case operation temperature				100	°C

Note: For further details refer to AN3338.

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Package information

6 Package information

In order to meet environmental requirements, ST offers these devices in different grades of ECOPACK® packages, depending on their level of environmental compliance. ECOPACK® specifications, grade definitions and product status are available at: www.st.com. ECOPACK® is an ST trademark.

Please refer to dedicated technical note TN0107 for mounting instructions.

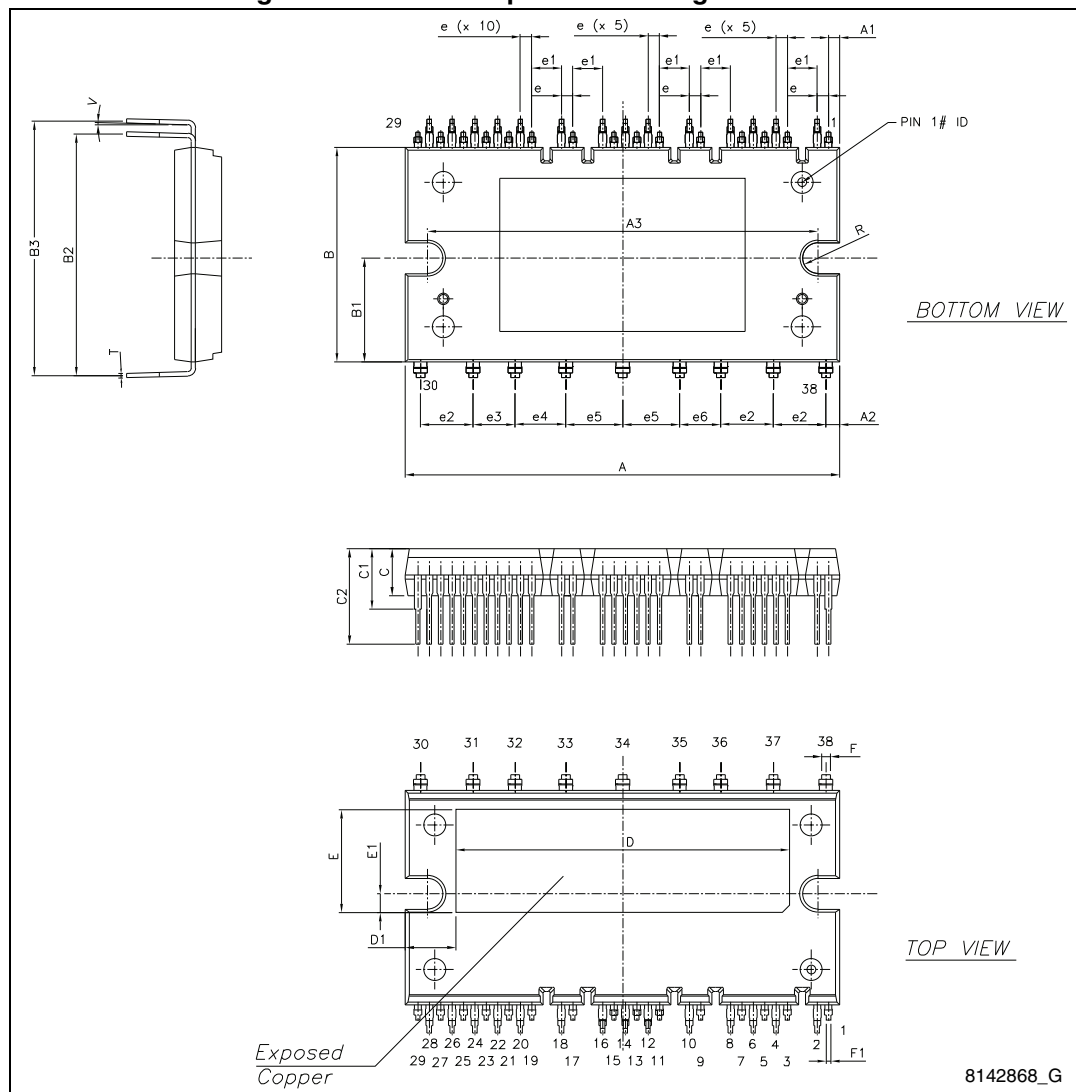
Table 16. SDIP-38L option A mechanical data

Dimensions	mm.		
	Min.	Typ.	Max.
A	49.10	49.60	50.10
A1	1.10	1.30	1.50
A2	1.40	1.60	1.80
A3	44.10	44.60	45.10
B	24.00	24.50	25.00
B1	11.25	11.85	12.45
B2	27.10	27.60	28.10
B3	28.60	29.10	29.60
C	5.00	5.40	6.00
C1	6.50	7.00	7.50
C2	10.35	10.85	11.35
e	1.10	1.30	1.50
e1	3.20	3.40	3.60
e2	5.80	6.00	6.20
e3	4.60	4.80	5.00
e4	5.60	5.80	6.00
e5	6.30	6.50	6.70
e6	4.50	4.70	4.90
D		38.10	
D1		5.75	
E		11.80	
E1		2.15	
F	0.85	1.00	1.15
F1	0.35	0.50	0.65
R	1.55	1.75	1.95
T	0.45	0.55	0.65
V	0°		6°

Package information

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Figure 12. SDIP-38L option A drawing dimensions



STGIPL14K60, STGIPL14K60-S

Package information

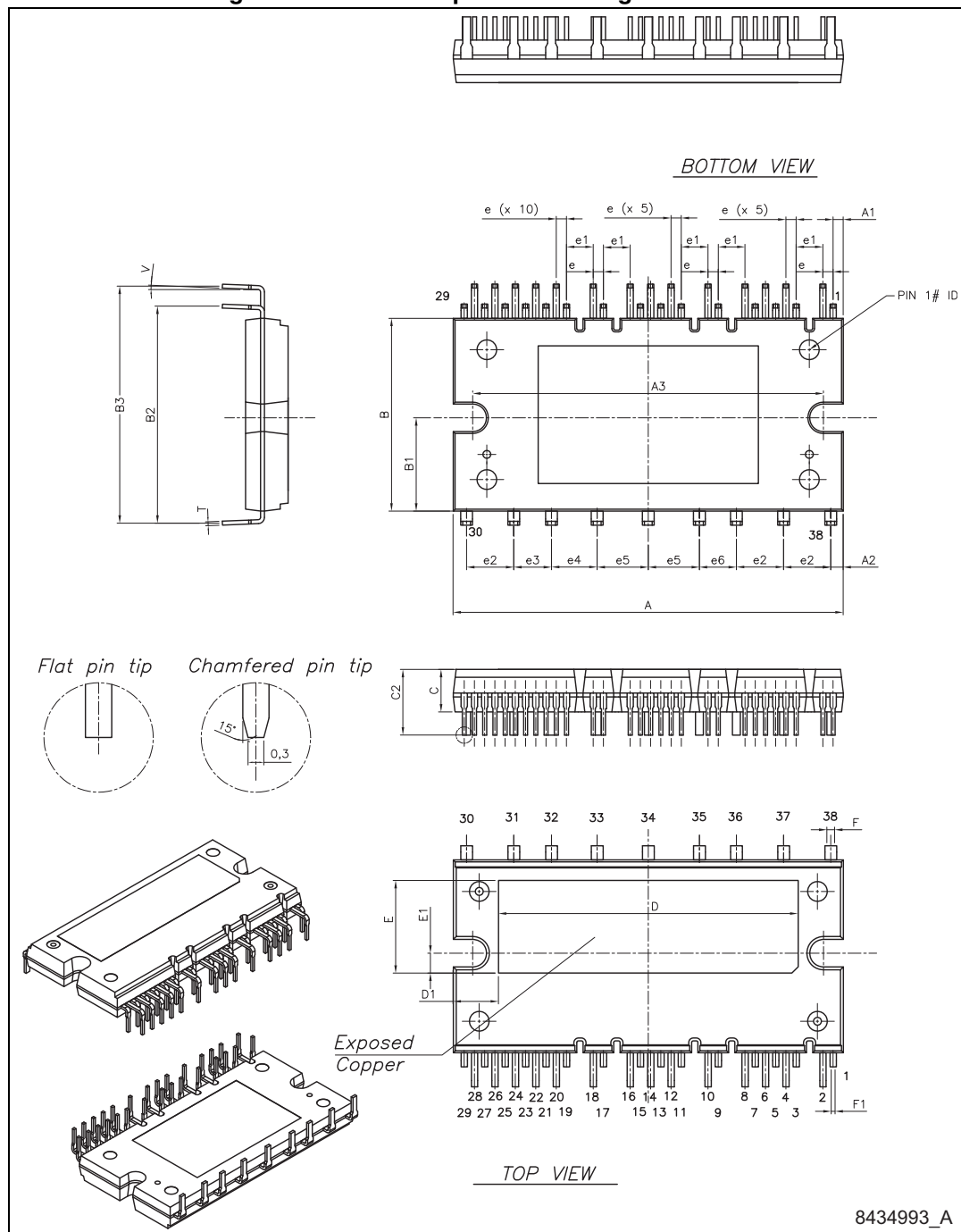
Table 17. SDIP-38L option B mechanical data

Dimensions	mm.		
	Min.	Typ.	Max.
A	49.10	49.60	50.10
A1	1.10	1.30	1.50
A2	1.40	1.60	1.80
A3	44.10	44.60	45.10
B	24.00	24.50	25.00
B1	11.25	11.85	12.45
B2	27.10	27.60	28.10
B3	29.65	30.15	30.65
C	5.00	5.40	6.00
C2	8.15	8.35	8.55
e	1.10	1.30	1.50
e1	3.20	3.40	3.60
e2	5.80	6.00	6.20
e3	4.60	4.80	5.00
e4	5.60	5.80	6.00
e5	6.30	6.50	6.70
e6	4.50	4.70	4.90
D		38.10	
D1		5.75	
E		11.80	
E1		2.15	
F	0.85	1.00	1.15
F1	0.35	0.50	0.65
R	1.55	1.75	1.95
T	0.45	0.55	0.65
V	0°		6°

Package information

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Figure 13. SDIP-38L option B drawing dimensions

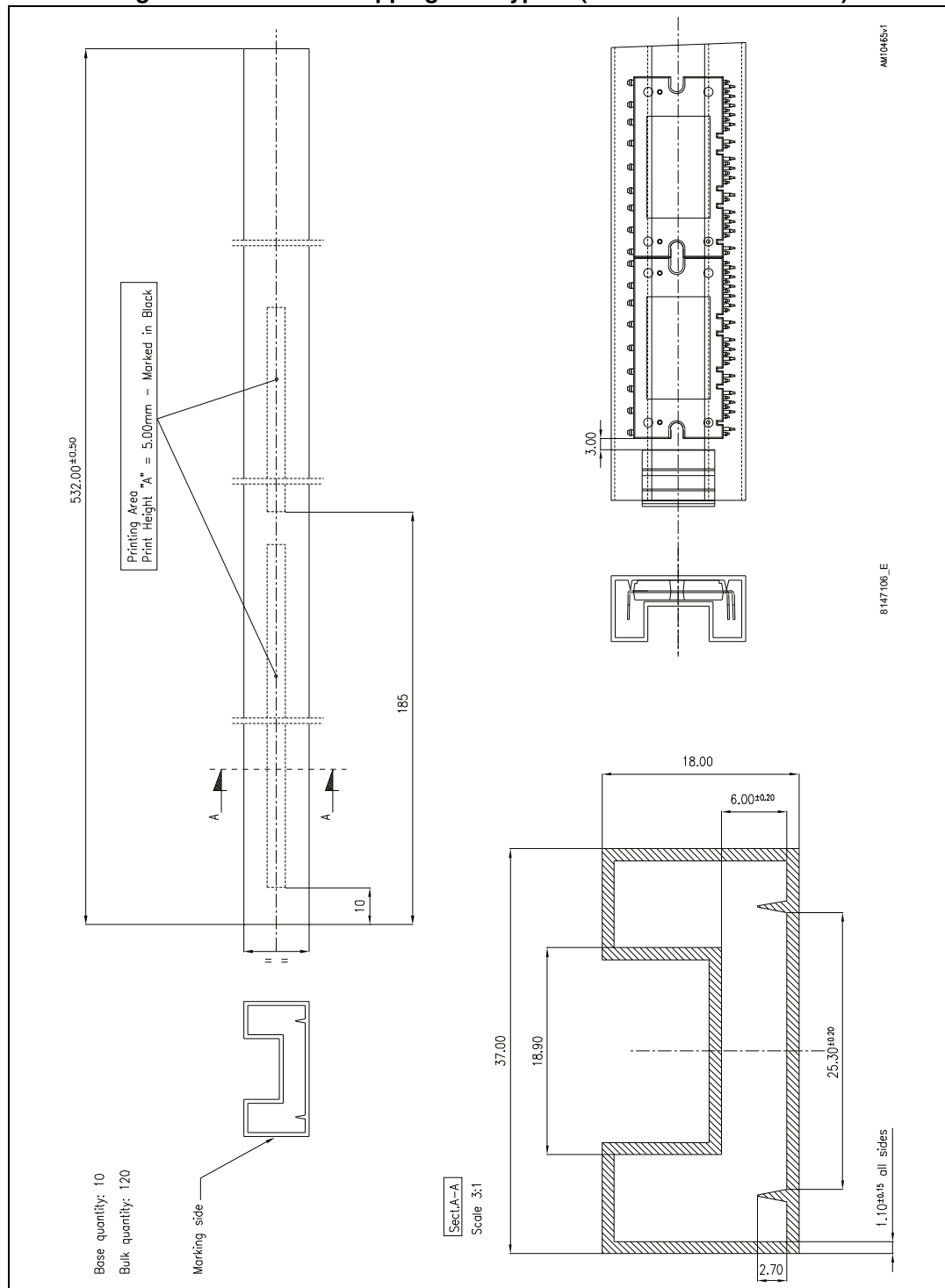


STGIPL14K60, STGIPL14K60-S

Packaging mechanical data

7 Packaging mechanical data

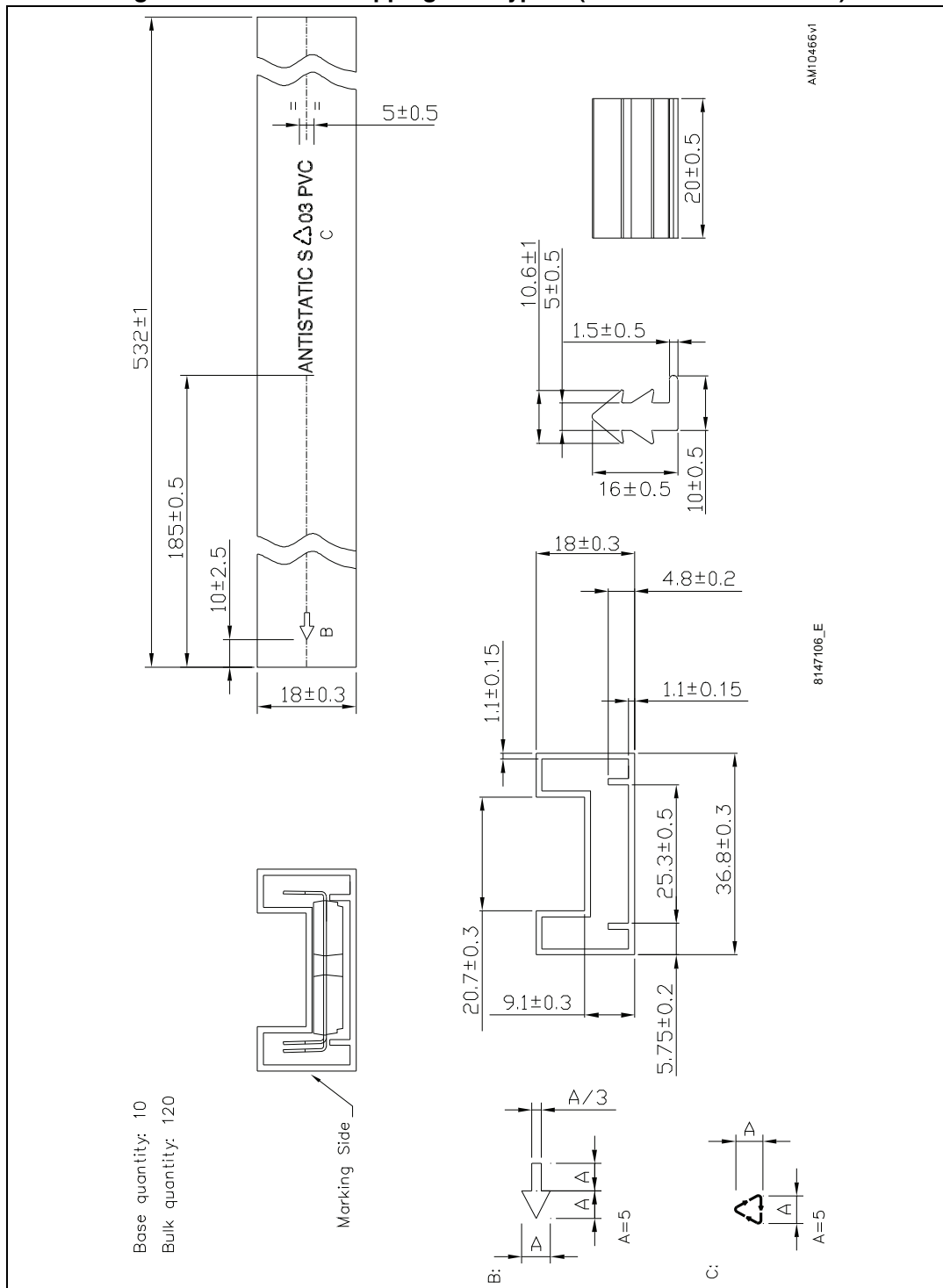
Figure 14. SDIP-38L shipping tube type A (dimensions are in mm.)



Packaging mechanical data

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Figure 15. SDIP-38L shipping tube type B (dimensions are in mm.)



8 Revision history

Table 18. Document revision history

Date	Revision	Changes
16-Apr-2009	1	Initial release
29-Mar-2010	2	Inserted Figure 5 , Figure 6 and Section 4: Smart shutdown function . Updated Section 3.1: Control part and package mechanical data, Section 6 . Minor text changes to improve readability.
14-Jun-2010	3	Document status promoted from preliminary data to datasheet. Updated Table 7: Inverter part , Figure 5: Maximum IC(RMS) current vs. switching frequency and Figure 6: Maximum IC(RMS) current vs. fsine(1) .
21-Sep-2010	4	Updated: Table 3 , 5 , 8 , 9 , 10 and 12 . Modified: Figure 5 and Figure 6 .
09-Mar-2011	5	Updated title with SLLIMM™ in cover page, added SDIP-38L tube dimensions Figure 14 .
04-Nov-2011	6	Updated title with SLLIMM™ (small low-loss intelligent molded module) IPM, 3-phase inverter - 15 A, 600 V short-circuit rugged IGBT in cover page and SDIP-38L mechanical data Table 16 on page 19 , Figure 12 .
28-Aug-2012	7	Modified: Min. and Max. value Table 4 on page 6 . Updated: Figure 14 . Added: Figure 15 .
04-Mar-2013	8	Added: Figure 7 and Figure 8
13-Mar-2013	9	Modified: Figure 8 on page 13
17-Jun-2013	10	Updated: Figure 9: Dead time and interlocking waveforms definitions .
17-Oct-2013	11	Added device STGIPL14K60-S and modified Table 1: Device summary accordingly. Updated Section 6: Package information and Section 7: Packaging mechanical data . Minor text changes.

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