

Excellent Integrated System Limited

Stocking Distributor

Click to view price, real time Inventory, Delivery & Lifecycle Information:

Maxim Integrated MAX1240ACSA

For any questions, you can email us directly: <u>sales@integrated-circuit.com</u>



Distributor of Maxim Integrated: Excellent Integrated System Limited Datasheet of MAX1240ACSA - IC ADC LP 12-BIT SERIAL 8-SOIC Contact us: sales@integrated-circuit.com Website: www.integrated-circuit.com



+2.7V, Low-Power, 12-Bit Serial ADCs in 8-Pin SO

+2.7V to +3.6V (MAX1240)

+2.7V to +5.25V (MAX1241)

Internal 2.5V Reference (MAX1240)

3mW (73ksps, MAX1241) 66µW (1ksps, MAX1241)

5µW (power-down mode)

Small Footprint: 8-Pin PDIP/SO Packages

SPI/QSPI/MICROWIRE 3-Wire Serial Interface

Low Power: 3.7µW (73ksps, MAX1240)

Single-Supply Operation:

12-Bit Resolution

Internal Track/Hold

Internal Clock

Features

General Description

The MAX1240/MAX1241 low-power, 12-bit analog-todigital converters (ADCs) are available in 8-pin packages. The MAX1240 operates with a single +2.7V to +3.6V supply, and the MAX1241 operates with a single +2.7V to +5.25V supply. Both devices feature a 7.5µs successive-approximation ADC, a fast track/hold (1.5µs), an on-chip clock, and a high-speed, 3-wire serial interface.

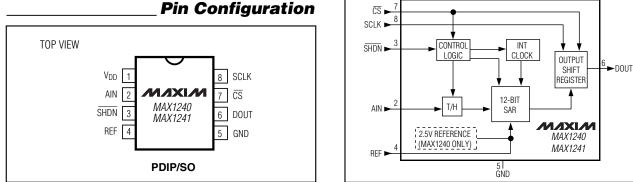
Power consumption is only 37mW (V_{DD} = 3V) at the 73ksps maximum sampling speed. A 2µA shutdown mode reduces power at slower throughput rates.

The MAX1240 has an internal 2.5V reference, while the MAX1241 requires an external reference. The MAX1241 accepts signals from OV to VREF, and the reference input range includes the positive supply rail. An external clock accesses data from the 3-wire interface, which connects directly to standard microcontroller I/O ports. The interface is compatible with SPI™, QSPI™, and MICROWIRE™.

Excellent AC characteristics and very low power combined with ease of use and small package size make these converters ideal for remote-sensor and dataacquisition applications, or for other circuits with demanding power consumption and space requirements. The MAX1240/MAX1241 are available in 8-pin PDIP and SO packages.

Applications

- Battery-Powered Systems Portable Data Logging
- Isolated Data Acquisition
- Process Control
- Instrumentation



SPI and QSPI are trademarks of Motorola, Inc. MICROWIRE is a trademark of National Semiconductor Corp.

Maxim Integrated Products 1

MAX1240/MAX124 **Ordering Information**

		,	
PART*	TEMP RANGE	PIN- PACKAGE	INL (LSB)
MAX1240ACPA+	0°C to +70°C	8 PDIP	±1/2
MAX1240BCPA+	0°C to +70°C	8 PDIP	±1
MAX1240CCPA+	0°C to +70°C	8 PDIP	±1
MAX1240ACSA+	0°C to +70°C	8 SO	±1/2
MAX1240BCSA+	0°C to +70°C	8 SO	±1
MAX1240CCSA+	0°C to +70°C	8 SO	±1
MAX1240CC/D	0°C to +70°C	Dice*	±1
MAX1240AESA/V+**	-40°C to +85°C	8 SO	±1/2
MAX1240BESA/V+	-40°C to +85°C	8 SO	±1

Ordering Information continued at end of data sheet.

*Dice are specified at $T_A = +25^{\circ}C$, DC parameters only. **Future product—contact factory for availability. /V denotes an automotive qualified part.

+Denotes a lead(Pb)-free/RoHS-compliant package.

V_{DD}

Functional Diagram

For pricing, delivery, and ordering information, please contact Maxim Direct at 1-888-629-4642, or visit Maxim's website at www.maxim-ic.com.



ABSOLUTE MAXIMUM RATINGS

V _{DD} to GND AIN to GND REF to GND Digital Inputs to GND	0.3V to (V _{DD} + 0.3V) 0.3V to (V _{DD} + 0.3V)
DOUT to GND DOUT Current	0.3V to (V _{DD} + 0.3V)
Continuous Power Dissipation (TA = +70° Plastic DIP (derate 9.09mW/°C above - SO (derate 5.88mW/°C above +70°C). CERDIP (derate 8.00mW/°C above +70°C)	+70°C)727mW 471mW

Operating Temperature Ranges	
MAX1240_C_A/MAX1241_C_A	0°C to +70°C
MAX1240_E_ A/MAX1241_E_ A	40°C to +85°C
MAX1240_MJA/MAX1241_MJA	55°C to +125°C
Storage Temperature Range	60°C to +150°C
Lead Temperature (soldering, 10s)	+300°C
Soldering Temperature (reflow)	
PDIP, SO	+260°C
CDIP	+250°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS	
DC ACCURACY (Note 1)		I						
Resolution				12			Bits	
	INII	MAX124_A	MAX124_A			±0.5		
Relative Accuracy (Note 2)	INL	MAX124_B/C				±1.0	– LSB	
Differential Nonlinearity	DNL	No missing codes over	r temperature			±1	LSB	
Offset Error		MAX124_A			±0.5	±3.0	LSB	
		MAX124_B/C			±0.5	±4.0		
Gain Error (Note 3)					±0.5	±4.0	LSB	
Gain Temperature Coefficient					±0.25		ppm/°C	
DYNAMIC SPECIFICATIONS (1)	0kHz sine-w	ave input, 0V to 2.500Vp	o-p, 73ksps, f _{SCLK} = 2.	1MHz)				
Signal-to-Noise Plus	SINAD	MAX124_A/B		70			dB	
Distortion Ratio	SINAD	MAX124_C			71.5			
Total Harmonic Distortion	THD	Lin to the 5th harmonic	MAX124_A/B			-80	dB	
			MAX124_C		-88			
Spurious-Free Dynamic Range	SFDR	MAX124_A/B		80			dB	
Spunous-Free Dynamic Range	JEDN	MAX124_C			88			
Small-Signal Bandwidth		-3dB rolloff			2.25		MHz	
Full-Power Bandwidth					1.0		MHz	
CONVERSION RATE	·							
Conversion Time	tCONV			5.5		7.5	μs	
Track/Hold Acquisition Time	tacq					1.5	μs	
Throughput Rate		f _{SCLK} = 2.1MHz				73	ksps	
Aperture Delay	tapr	Figure 8			30		ns	
Aperture Jitter					<50		ps	
ANALOG INPUT								
Input Voltage Range				0		V _{REF}	V	
Input Capacitance					16		pF	

M/IXI/M



ELECTRICAL CHARACTERISTICS (continued)

 $(V_{DD} = +2.7V \text{ to } +3.6V \text{ (MAX1240}); V_{DD} = +2.7V \text{ to } +5.25V \text{ (MAX1241)}; 73 \text{ksps}, f_{SCLK} = 2.1 \text{MHz} (50\% \text{ duty cycle}); MAX1240 - 4.7 \mu \text{F} \text{capacitor at REF pin, MAX1241} - external reference; V_{REF} = 2.500V applied to REF pin; T_A = T_{MIN} \text{ to } T_{MAX}; unless otherwise noted.}$

PARAMETER	SYMBOL	CONDITIONS	MIN	ΤΥΡ	MAX	UNITS
INTERNAL REFERENCE (MAX1	240 only)		ł			1
REF Output Voltage		$T_A = +25^{\circ}C$	2.480	2.500	2.520	V
REF Short-Circuit Current					30	mA
REF Temperature Coefficient		MAX1240AC/BC		±30	±50	
		MAX1240AE/BE		±30	±60	ppm/°C
REF Temperature Coemcient		MAX1240AM/BM		±30	±80	
		MAX1240C		±30		1
Load Regulation (Note 4)		0mA to 0.2mA output load		0.35		
Capacitive Bypass at REF			4.7			μF
EXTERNAL REFERENCE (VREF	= 2.500V)		I			
Input Voltage Range			1.00		V _{DD} + 50mV	V
Input Current				100	150	μA
Input Resistance			18	25		kΩ
REF Input Current in Shutdown		V _{SHDN} = 0V		±0.01	10	μA
Capacitive Bypass at REF			0.1			μF
DIGITAL INPUTS: SCLK, CS, S	HDN	I	I			1
SCLK, CS Input High Voltage		$V_{DD} \le 3.6V$	2.0			
		V _{DD} > 3.6V (MAX1241)	3.0			V
SCLK, CS Input Low Voltage	VIL				0.8	V
SCLK, CS Input Hysteresis	VHYST			0.2		V
SCLK, CS Input Leakage	lin	$V_{IN} = 0V \text{ or } V_{DD}$		±0.01	±1	μA
SCLK, CS Input Capacitance	CIN	(Note 5)			15	pF
SHDN Input High Voltage	V _{SH}		V _{DD} - 0.4	1		V
SHDN Input Low Voltage	VSL				0.4	V
SHDN Input Current		$V_{\overline{SHDN}} = 0V \text{ or } V_{DD}$			±4.0	μA
SHDN Input Mid Voltage	V _{SM}		1.1		V _{DD} - 1.1	V
SHDN Voltage, Unconnected	VFLT	SHDN = unconnected		$V_{DD}/2$		V
SHDN Max Allowed Leakage, Mid Input		SHDN = unconnected			±100	nA
DIGITAL OUTPUT: DOUT	1	I	I			1
Output Voltage Low	Ve	I _{SINK} = 5mA			0.4	- V
Output Voltage Low	Vol	I _{SINK} = 16mA			0.8	V
Output Voltage High	Voн	ISOURCE = 0.5mA	V _{DD} - 0.5	5		V
Three-State Leakage Current	١L	$\overline{\text{CS}} = \text{V}_{\text{DD}}$		±0.01	±10	μA
Three-State Output Capacitance	Соит	$\overline{\text{CS}} = \text{V}_{\text{DD}}$ (Note 5)			15	pF



ELECTRICAL CHARACTERISTICS (continued)

PARAMETERS	SYMBOL		CONDITION	S	MIN	TYP	MAX	UNITS
POWER REQUIREMENTS		1						-
Supply Voltage	V _{DD}	MAX1240			2.7		3.6	V
Supply Voltage		MAX1241			2.7		5.25	
			MAX1240A/B			1.4	2.0	
		Operating mode	MAX1240C	$V_{DD} = 3.6V$		1.4	3.5	
			MAX1241A/B	$V_{DD} = 3.6V$		0.9 1.5	1.5	mA
Supply Current				V _{DD} = 5.25V		1.6	2.5	
Supply Current	IDD		MAX1241C	V _{DD} = 3.6V		0.9	2.8	
			WAX 124 10	$V_{DD} = 5.25V$		1.6	3.8	
				V _{DD} = 3.6V		1.9	10	
				$V_{DD} = 5.25V$		3.5	15	μΑ
Supply Rejection	PSR	(Note 5)				±0.3		mV

TIMING CHARACTERISTICS (Figure 8)

 $(V_{DD} = +2.7V \text{ to } +3.6V \text{ (MAX1240)}; V_{DD} = +2.7V \text{ to } +5.25V \text{ (MAX1241)}; T_A = T_{MIN} \text{ to } T_{MAX}, \text{ unless otherwise noted.})$

PARAMETERS	SYMBOL	CONDITIONS		MIN	TYP	МАХ	UNITS
Acquisition Time	tacq	$\overline{\text{CS}} = \text{V}_{\text{DD}}$ (Note 6)		1.5			μs
SCLK Fall to Output Data Valid	too	Figure 1,	MAX124C/E	20		200	ns
	UDU IDU	t _{DO} C _{LOAD} = 50pF M	MAX124M	20		240	- 115
CS Fall to Output Enable	t _{DV}	Figure 1, C _{LOAD} = 50pF				240	ns
CS Rise to Output Disable	t _{TR}	Figure 2, C _{LOAD} = 50pF				240	ns
SCLK Clock Frequency	fsclk			0		2.1	MHz
SCLK Pulse Width High	tсн			200			ns
SCLK Pulse Width Low	tCL			200			ns
SCLK Low to CS Fall Setup Time	tcso			50			ns
DOUT Rise to SCLK Rise (Note 5)	tstr			0			ns
CS Pulse Width	t _{CS}			240			ns

Note 1: Tested at $V_{DD} = +2.7V$.

Note 2: Relative accuracy is the deviation of the analog value at any code from its theoretical value after the full-scale range and offset have been calibrated.

Note 3: MAX1240—internal reference, offset nulled; MAX1241—external reference (VREF = +2.500V), offset nulled.

Note 4: External load should not change during conversion for specified accuracy.

Note 5: Guaranteed by design. Not subject to production testing.

- **Note 6:** Measured as [V_{FS}(2.7V) V_{FS}(V_{DD(MAX})].
- Note 7: To guarantee acquisition time, t_{ACQ} is the maximum time the device takes to acquire the signal, and is also the minimum time needed for the signal to be acquired.



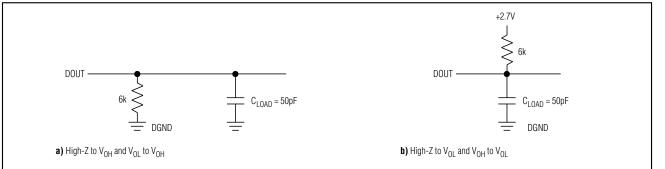


Figure 1. Load Circuits for DOUT Enable Time

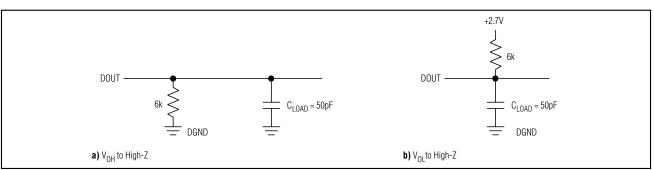
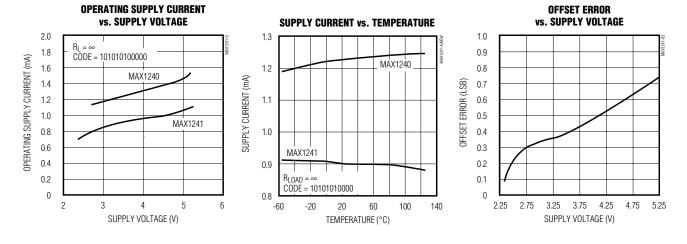


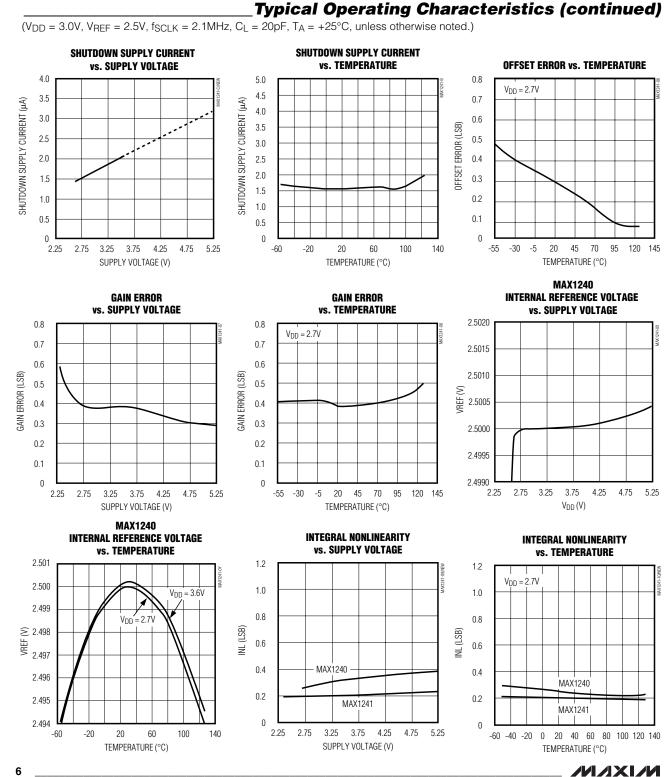
Figure 2. Load Circuits for DOUT Disable Time

Typical Operating Characteristics

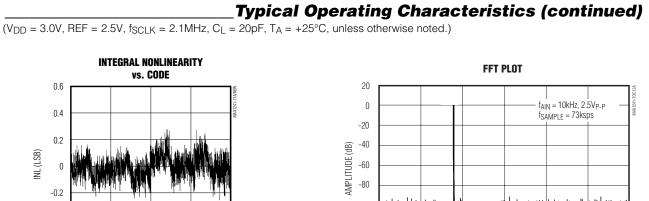
(V_{DD} = 3.0V, V_{REF} = 2.5V, f_{SCLK} = 2.1MHz, C_L = 20pF, T_A = +25°C, unless otherwise noted.)

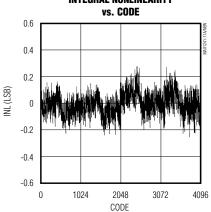


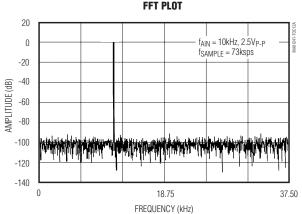












Pin Description

PIN	NAME	FUNCTION
1	VDD	Positive Supply Voltage: 2.7V to 3.6V, (MAX1240); 2.7V to 5.25V (MAX1241)
2	AIN	Sampling Analog Input, 0V to VREF range
3	SHDN	Three-Level Shutdown Input. Pulling SHDN low shuts the MAX1240/MAX1241 down to 15µA (max) supply current. Both the MAX1240 and MAX1241 are fully operational with either SHDN high or unconnected. For the MAX1240, pulling SHDN high enables the internal reference, and letting SHDN open disables the internal reference and allows for the use of an external reference.
4	REF	Reference Voltage for Analog-to-Digital Conversion. Internal 2.5V reference output for MAX1240; bypass with 4.7µF capacitor. External reference voltage input for MAX1241, or for MAX1240 with the internal reference disabled. Bypass REF with a minimum of 0.1µF when using an external reference.
5	GND	Analog and Digital Ground
6	DOUT	Serial Data Output. Data changes state at SCLK's falling edge. DOUT is high impedance when \overline{CS} is high.
7	CS	Active-Low Chip Select initiates conversions on the falling edge. When \overline{CS} is high, DOUT is high impedance.
8	SCLK	Serial Clock Input. SCLK clocks data out at rates up to 2.1MHz.



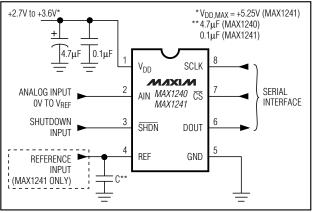


Figure 3. Operational Diagram

Detailed Description

Converter Operation

The MAX1240/MAX1241 use an input track/hold (T/H) and successive-approximation register (SAR) circuitry to convert an analog input signal to a digital 12-bit output. No external-hold capacitor is needed for the T/H. Figure 3 shows the MAX1240/MAX1241 in its simplest configuration. The MAX1240/MAX1241 convert input signals in the 0V to V_{REF} range in 9µs, including T/H acquisition time. The MAX1240's internal reference is trimmed to 2.5V, while the MAX1241 requires an external reference. Both devices accept voltages from 1.0V to V_DD. The serial interface requires only three digital lines (SCLK, \overline{CS} , and DOUT) and provides an easy interface to microprocessors (µPs).

The MAX1240/MAX1241 have two modes: normal and shutdown. Pulling SHDN low shuts the device down and reduces supply current below 10µA (V_{DD} \leq 3.6V), while pulling SHDN high or leaving it open puts the device into operational mode. Pulling CS low initiates a conversion. The conversion result is available at DOUT in unipolar serial format. The serial data stream consists of a high bit, signaling the end of conversion (EOC), followed by the data bits (MSB first).

Analog Input

Figure 4 illustrates the sampling architecture of the analog-to-digital converter's (ADC's) comparator. The fullscale input voltage is set by the voltage at REF.

Track/Hold

In track mode, the analog signal is acquired and stored in the internal hold capacitor. In hold mode, the T/H switch opens and maintains a constant input to the ADC's SAR section.

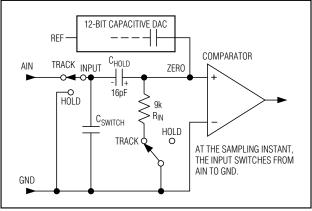


Figure 4. Equivalent Input Circuit

During acquisition, the analog input (AIN) charges capacitor C_{HOLD} . Bringing \overline{CS} low ends the acquisition interval. At this instant, the T/H switches the input side of C_{HOLD} to GND. The retained charge on C_{HOLD} represents a sample of the input, unbalancing node ZERO at the comparator's input.

In hold mode, the capacitive digital-to-analog converter (DAC) adjusts during the remainder of the conversion cycle to restore node ZERO to 0V within the limits of 12-bit resolution. This action is equivalent to transferring a charge from C_{HOLD} to the binary-weighted capacitive DAC, which in turn forms a digital representation of the analog input signal. At the conversion's end, the input side of C_{HOLD} switches back to AIN, and C_{HOLD} charges to the input signal again.

The time required for the T/H to acquire an input signal is a function of how quickly its input capacitance is charged. If the input signal's source impedance is high, the acquisition time lengthens and more time must be allowed between conversions. The acquisition time (t_{ACQ}) is the maximum time the device takes to acquire the signal, and is also the minimum time needed for the signal to be acquired. Acquisition time is calculated by:

$$t_{ACQ} = 9(R_S + R_{IN}) \times 16 pF$$

where ${\sf R}_{\sf IN}$ = 9k $\Omega,\,{\sf R}_{S}$ = the input signal's source impedance, and $t_{\sf ACQ}$ is never less than 1.5µs. Source impedances below 1k Ω do not significantly affect the ADC's AC performance.

Higher source impedances can be used if a 0.01μ F capacitor is connected to the analog input. Note that the input capacitor forms an RC filter with the input source impedance, limiting the ADC's input signal bandwidth.





External Reference

Input Bandwidth

The ADCs' input tracking circuitry has a 2.25MHz smallsignal bandwidth, so it is possible to digitize highspeed transient events and measure periodic signals with bandwidths exceeding the ADC's sampling rate by using undersampling techniques. To avoid aliasing of unwanted high-frequency signals into the frequency band of interest, anti-alias filtering is recommended.

Analog Input Protection

Internal protection diodes, which clamp the analog input to V_{DD} and GND, allow the input to swing from GND - 0.3V to V_{DD} + 0.3V without damage. However, for accurate conversions near full scale, the input must not exceed V_{DD} by more than 50mV, or be lower than GND by 50mV.

If the analog input exceeds 50mV beyond the supplies, limit the input current to 2mA.

Internal Reference (MAX1240)

The MAX1240 has an on-chip voltage reference trimmed to 2.5V. The internal reference output is connected to REF and also drives the internal capacitive DAC. The output can be used as a reference voltage source for other components and can source up to 400µA. Bypass REF with a 4.7µF capacitor. Larger capacitors increase wake-up time when exiting shutdown (see the section *Using SHDN to Reduce Supply Current*). The internal reference is enabled by pulling the SHDN pin high. Letting SHDN open disables the internal reference, as described in the *External Reference* section.

The MAX1240/MAX1241 operate with an external reference at the REF pin. To use the MAX1240 with an external reference, disable the internal reference by letting SHDN open. Stay within the +1.0V to V_{DD} voltage range to achieve specified accuracy. The minimum input impedance is 18k Ω for DC currents. During conversion, the external reference must be able to deliver up to 250µA of DC load current and have an output impedance of 10 Ω or less. The recommended minimum value for the bypass capacitor is 0.1µF. If the reference has higher output impedance or is noisy, bypass it close to the REF pin with a 4.7µF capacitor.

Serial Interface

Initialization after Power-Up and Starting a Conversion

When power is first applied, and if \overline{SHDN} is not pulled low, it takes the fully discharged 4.7μ F reference bypass capacitor up to 20ms to provide adequate charge for specified accuracy. With an external reference, the internal reset time is 10µs after the power supplies have stabilized. No conversions should be performed during these times.

To start a conversion, pull \overline{CS} low. At \overline{CS} 's falling edge, the T/H enters its hold mode and a conversion is initiated. After an internally timed conversion period, the end of conversion is signaled by DOUT pulling high. Data can then be shifted out serially with the external clock.



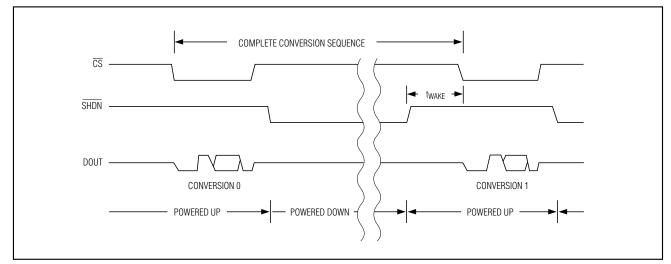


Figure 5. Shutdown Sequence



Using SHDN to Reduce Supply Current

Power consumption can be reduced significantly by shutting down the MAX1240/MAX1241 between conversions. Figure 6 shows a plot of average supply current versus conversion rate. Because the MAX1241 uses an external reference voltage (assumed to be present continuously), it "wakes up" from shutdown more quickly (in 4µs) and therefore provides lower average supply currents. The wake-up time (twakE) is the time from when SHDN is deasserted to the time when a conversion may be initiated (Figure 5). For the MAX1240, this time depends on the time in shutdown (Figure 7) because the external 4.7µF reference bypass capacitor loses charge slowly during shutdown.

External Clock The actual conversion does not require the external clock. This allows the conversion result to be read back at the μ P's convenience at any clock rate from up to 2.1MHz. The clock duty cycle is unrestricted if each clock phase is at least 200ns. Do not run the clock

Timing and Control

Conversion-start and data-read operations are controlled by the \overline{CS} and SCLK digital inputs. The timing diagrams of Figures 8 and 9 outline serial-interface operation.

while a conversion is in progress.

A $\overline{\text{CS}}$ falling edge initiates a conversion sequence: the T/H stage holds the input voltage, the ADC begins to convert, and DOUT changes from high impedance to logic low. SCLK must be kept low during the conversion. An internal register stores the data when the conversion is in progress.

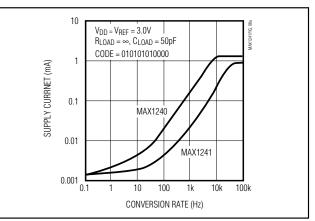


Figure 6. Average Supply Current vs. Conversion Rate

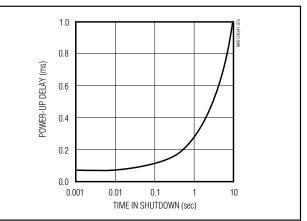


Figure 7. Typical Reference Power-Up Delay vs. Time in Shutdown

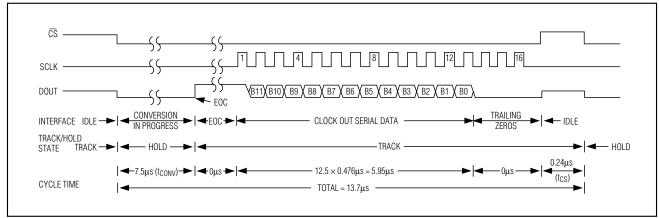


Figure 8. Interface Timing Sequence

M/IXI/M



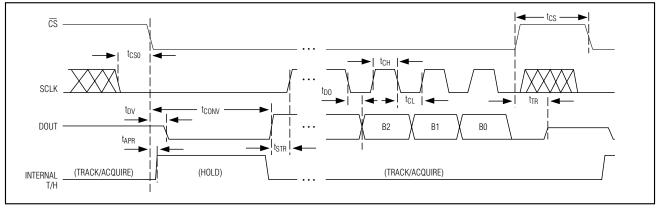


Figure 9. Detailed Serial-Interface Timing

End of conversion (EOC) is signaled by DOUT going high. DOUT's rising edge can be used as a framing signal. SCLK shifts the data out of this register any time after the conversion is complete. DOUT transitions on SCLK's falling edge. The next falling clock edge produces the MSB of the conversion at DOUT, followed by the remaining bits. Since there are 12 data bits and one leading high bit, at least 13 falling clock edges are needed to shift out these bits. Extra clock pulses occurring after the conversion result has been clocked out, and prior to a rising edge of \overline{CS} , produce trailing zeros at DOUT and have no effect on converter operation.

Minimum cycle time is accomplished by using DOUT's rising edge as the EOC signal. Clock out the data with 12.5 clock cycles at full speed. Pull $\overline{\text{CS}}$ high after reading

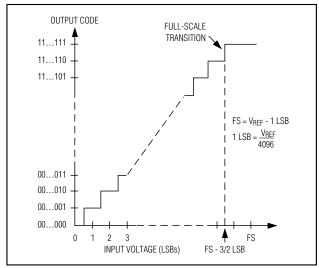


Figure 10. Unipolar Transfer Function, Full Scale (FS) = V_{REF} - 1 LSB, Zero Scale (ZS) = GND

the conversion's LSB. After the specified minimum time (tcs), $\overline{\text{CS}}$ can be pulled low again to initiate the next conversion.

Output Coding and Transfer Function

The data output from the MAX1240/MAX1241 is binary, and Figure 10 depicts the nominal transfer function. Code transitions occur halfway between successive-integer LSB values. If $V_{REF} = +2.500V$, then 1 LSB = 610µV or 2.500V/4096.

Applications Information

Connection to Standard Interfaces

The MAX1240/MAX1241 serial interface is fully compatible with SPI/QSPI and MICROWIRE standard serial interfaces (Figure 11).

If a serial interface is available, set the CPU's serial interface in master mode so the CPU generates the serial clock. Choose a clock frequency up to 2.1MHz.

- 1) Use a general-purpose I/O line on the CPU to pull CS low. Keep SCLK low.
- 2) Wait the for the maximum conversion time specified before activating SCLK. Alternatively, look for a DOUT rising edge to determine the end of conversion.
- 3) Activate SCLK for a minimum of 13 clock cycles. The first falling clock edge produces the MSB of the DOUT conversion. DOUT output data transitions on SCLK's falling edge and is available in MSB-first format. Observe the SCLK to DOUT valid timing characteristic. Data can be clocked into the μ P on SCLK's rising edge.
- 4) Pull CS high at or after the 13th falling clock edge. If CS remains low, trailing zeros are clocked out after the LSB.



5) With $\overline{\text{CS}}$ = high, wait the minimum specified time, t_{CS}, before initiating a new conversion by pulling $\overline{\text{CS}}$ low. If a conversion is aborted by pulling $\overline{\text{CS}}$ high before the conversion's end, wait for the minimum acquisition time, t_{ACQ}, before starting a new conversion.

CS must be held low until all data bits are clocked out. Data can be output in two bytes or continuously, as shown in Figure 8. The bytes contain the result of the conversion padded with one leading 1, and trailing 0s.

SPI and MICROWIRE

When using SPI or MICROWIRE, set <u>CPOL</u> = 0 and CPHA = 0. Conversion begins with a <u>CS</u> falling edge. DOUT goes low, indicating a conversion in progress. Wait until DOUT goes high or until the maximum specified 7.5µs conversion time elapses. Two consecutive 1-byte reads are required to get the full 12 bits from the ADC. DOUT output data transitions on SCLK's falling edge and is clocked into the µP on SCLK's rising edge.

The first byte contains a leading 1, and seven bits of conversion result. The second byte contains the remaining five bits and three trailing zeros. See Figure 11 for connections and Figure 12 for timing.

QSPI

Set CPOL = CPHA = 0. Unlike SPI, which requires two 1-byte reads to acquire the 12 bits of data from the ADC, QSPI allows the minimum number of clock cycles necessary to clock in the data. The MAX1240/MAX1241 requires 13 clock cycles from the μ P to clock out the 12 bits of data with no trailing zeros (Figure 13). The maximum clock frequency to ensure compatibility with QSPI is 2.097MHz.

Layout, Grounding, and Bypassing

For best performance, use printed circuit boards. Wirewrap boards are not recommended. Board layout should ensure that digital and analog signal lines are separated from each other. Do not run analog and digital (especially clock) lines parallel to one another, or digital lines underneath the ADC package.

Figure 14 shows the recommended system ground connections. Establish a single-point analog ground ("star" ground point) at GND, separate from the logic ground. Connect all other analog grounds and DGND to this star ground point for further noise reduction. No other digital

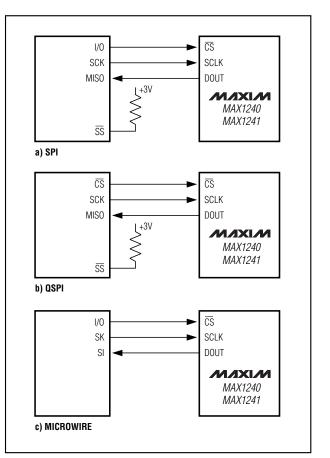


Figure 11. Common Serial-Interface Connections to the MAX1241

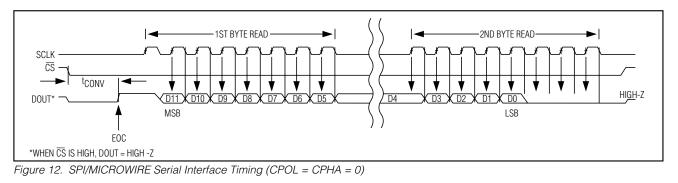
system ground should be connected to this single-point analog ground. The ground return to the power supply for this ground should be low impedance and as short as possible for noise-free operation.

High-frequency noise in the V_{DD} power supply may affect the ADC's high-speed comparator. Bypass this supply to the single-point analog ground with 0.1μ F and 4.7μ F bypass capacitors. Minimize capacitor lead lengths for best supply-noise rejection. If the power supply is very noisy, a 10Ω resistor can be connected as a lowpass filter to attenuate supply noise (Figure 14).

Why CPI DOI untii 7.5µ read DOI is cl The vers five

MAX1240/MAX124





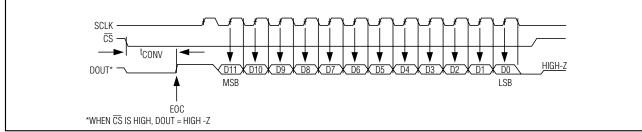


Figure 13. QSPI Serial Interface Timing (CPOL = CPHA = 0)

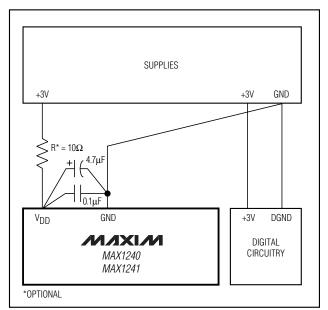


Figure 14. Power-Supply Grounding Condition



Ordering Information (continued)

PIN-INL PART TEMP RANGE PACKAGE (LSB) MAX1240AEPA+ -40°C to +85°C $\pm^{1/2}$ 8 PDIP MAX1240BEPA+ -40°C to +85°C 8 PDIP ±1 MAX1240CEPA+ -40°C to +85°C 8 PDIP ±1 MAX1240AESA+ -40°C to +85°C 8 SO $\pm^{1/2}$ MAX1240BESA+ -40°C to +85°C 8 SO ±1 MAX1240CESA+ -40°C to +85°C 8 SO ±1 MAX1240AMJA+ -55°C to +125°C $\pm^{1/2}$ 8 CERDIP[†] MAX1240BMJA+ -55°C to +125°C 8 CERDIP[†] ±1 MAX1240CMJA+ -55°C to +125°C 8 CERDIP[†] +1 MAX1241ACPA+ 0°C to +70°C 8 PDIP $\pm^{1/2}$ $0^{\circ}C$ to $+70^{\circ}C$ MAX1241BCPA+ 8 PDIP ±1 MAX1241CCPA+ 0°C to +70°C 8 PDIP ±1 MAX1241ACSA+ 0°C to +70°C 8 SO $\pm^{1/2}$ 0°C to +70°C MAX1241BCSA + 8 SO ± 1 0°C to +70°C 8 SO MAX1241CCSA+ ±1 0°C to +70°C MAX1241BC/D Dice* ±1 -40°C to +85°C 8 PDIP MAX1241AEPA+ $\pm^{1/2}$ 8 PDIP MAX1241BEPA+ -40°C to +85°C ±1 MAX1241CEPA+ -40°C to +85°C 8 PDIP ±1 $\pm^{1/2}$ MAX1241AESA+ -40°C to +85°C 8 SO MAX1241BESA+ -40°C to +85°C 8 SO +1 MAX1241CESA+ -40°C to +85°C 8 SO ±1 MAX1241AMJA+ -55°C to +125°C $\pm^{1/2}$ 8 CERDIP[†] -55°C to +125°C MAX1241BMJA+ ±1 8 CERDIP[†] MAX1241CMJA+ -55°C to +125°C 8 CERDIP[†] ±1

+Denotes lead(Pb)-free/RoHS-compliant package. *Dice are specified at $T_A = +25^{\circ}C$, DC parameters only.

[†]Contact factory for availability and processing to MIL-STD-883.

Chip Information

PROCESS: BICMOS SUBSTRATE CONNECTED TO GND

Package Information

For the latest package outline information and land patterns, go to **www.maxim-ic.com/packages**. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

PACKAGE TYPE	PACKAGE CODE	OUTLINE NO.	LAND PATTERN NO.
8 PDIP	P8+2	<u>21-0043</u>	—
8 SO	S8+5	<u>21-0041</u>	<u>90-0096</u>
8 CERDIP	J8+2	<u>21-0045</u>	—



Revision History

REVISION NUMBER	REVISION DATE	DESCRIPTION	PAGES CHANGED
3	3/10	Added automotive grade to data sheet	1, 2, 3, 7, 9, 14, 15, 16
4	6/10	Future product note removed from one part in the Ordering Information	1
5	8/10	Removed MAX1240BC/D and add MAX1240CC/D	1

Maxim cannot assume responsibility for use of any circuitry other than circuitry entirely embodied in a Maxim product. No circuit patent licenses are implied. Maxim reserves the right to change the circuitry and specifications without notice at any time.

_Maxim Integrated Products, 120 San Gabriel Drive, Sunnyvale, CA 94086 408-737-7600

© 2010 Maxim Integrated Products

15

Maxim is a registered trademark of Maxim Integrated Products, Inc.