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Fairchild Semiconductor FGH50N6S2

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August 2003

FGH50N6S2

600V, SMPS II Series N-Channel IGBT

General Description

The FGH50N6S2 is a Low Gate Charge, Low Plateau Voltage SMPS II IGBT combining the fast switching speed of the SMPS IGBTs along with lower gate charge, plateau voltage and avalanche capability (UIS). These LGC devices shorten delay times, and reduce the power requirement of the gate drive. These devices are ideally suited for high voltage switched mode power supply applications where low conduction loss, fast switching times and UIS capability are essential. SMPS II LGC devices have been specially designed for:

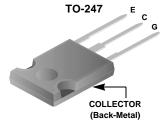
- Power Factor Correction (PFC) circuits
- · Full bridge topologies
- Half bridge topologies
- · Push-Pull circuits
- Uninterruptible power supplies
- · Zero voltage and zero current switching circuits

IGBT formerly Developmental Type TA49342

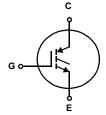
Features

- 100kHz Operation at 390V, 40A
- 200kHZ Operation at 390V, 25A
- 600V Switching SOA Capability
- Typical Fall Time. 90ns at TJ = 125°C
- Low Gate Charge 70nC at V_{GE} = 15V
- Low Plateau Voltage6.5V Typical
- · Low Conduction Loss

Package



Symbol



Device Maximum Ratings T_{C} = 25°C unless otherwise noted

Symbol	Parameter	Ratings	Units	
BV _{CES}	Collector to Emitter Breakdown Voltage	600	V	
I _{C25}	Collector Current Continuous, T _C = 25°C	75	Α	
I _{C110}	Collector Current Continuous, T _C = 110°C	60	Α	
I _{CM}	Collector Current Pulsed (Note 1)	240	Α	
V _{GES}	Gate to Emitter Voltage Continuous	±20	V	
V _{GEM}	Gate to Emitter Voltage Pulsed	±30	V	
SSOA	Switching Safe Operating Area at T _J = 150°C, Figure 2	150A at 600V		
E _{AS}	Pulsed Avalanche Energy, I _{CE} = 30A, L = 1mH, V _{DD} = 50V	480	mJ	
P _D	Power Dissipation Total T _C = 25°C	463	W	
	Power Dissipation Derating T _C > 25°C		W/°C	
TJ	Operating Junction Temperature Range	-55 to 150	°C	
T _{STG}	Storage Junction Temperature Range	-55 to 150	°C	

CAUTION: Stresses above those listed in "Device Maximum Ratings" may cause permanent damage to the device. This is a stress only rating and operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied.

NOTE:

Pulse width limited by maximum junction temperature.



Package Marking and Ordering Information

Device Marking	Device	Package	Reel Size	Tape Width	Quantity
50N6S2	FGH50N6S2	TO-247	Tube	N/A	30

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Electrical Characteristics T_J = 25°C unless otherwise noted

Symbol	Parameter	Test Conditions		Min	Тур	Max	Units
Off Stat	e Characteristics						
BV _{CES}	Collector to Emitter Breakdown Voltage	l _C = 250μA, V _{GE} = 0		600	-	-	V
BV _{ECS}	Emitter to Collector Breakdown Voltage			20	-	-	V
I _{CES}	Collector to Emitter Leakage Current	V _{CE} = 600V	$T_J = 25^{\circ}C$	-	-	250	μΑ
			$T_J = 125$ °C	-	-	2.8	mA
I _{GES}	Gate to Emitter Leakage Current	V _{GE} = ± 20V		-	-	±250	nA
On State	e Characteristics						
V _{CE(SAT)}	Collector to Emitter Saturation Voltage	$I_{\rm C} = 30A$,	T _J = 25°C	-	1.9	2.7	V
OL(OAI)		V _{GE} = 15V	T _{.I} = 125°C	-	1.7	2.2	V
V _{EC}	Diode Forward Voltage	I _{EC} = 30A		-	2.2	2.6	V
Dvnami	c Characteristics				•		
Q _{G(ON)}	Gate Charge	I _C = 30A,	V _{GE} = 15V	-	70	85	nC
- G(ON)		$V_{CE} = 300V$	$V_{GF} = 20V$	-	90	110	nC
V _{GE(TH)}	Gate to Emitter Threshold Voltage	$I_{C} = 250 \mu A, V_{CE} = V_{GE}$		3.5	4.3	5.0	V
V _{GEP}	Gate to Emitter Plateau Voltage	I _C = 30A, V _{CE} = 300V		-	6.5	8.0	V
	ng Characteristics						
SSOA	Switching SOA	$T_J = 150$ °C, $V_{GE} = 15$ V, $R_G = 3\Omega$		150	-	-	Α
		$L = 100 \mu H, V_{CE}$					
t _{d(ON)I}	Current Turn-On Delay Time	IGBT and Diode at $T_J = 25$ °C, $I_{CF} = 30A$,		-	13	-	ns
t _{rl}	Current Rise Time	$I_{CE} = 30A$, $V_{CE} = 390V$, $V_{GE} = 15V$, $R_{G} = 3\Omega$ $L = 200\mu H$ Test Circuit - Figure 26		-	15	-	ns
t _{d(OFF)I}	Current Turn-Off Delay Time			-	55	-	ns
t _{fl}	Current Fall Time			-	50	-	ns
E _{ON1}	Turn-On Energy (Note 2)			-	260	-	μJ
E _{ON2}	Turn-On Energy (Note 2)			-	330	-	μJ
E _{OFF}	Turn-Off Energy (Note 3)	IODT and Diada -+ T 40500		-	250	350	μJ
t _{d(ON)I}	Current Turn-On Delay Time	IGBT and Diode at T _J = 125°C			13	-	ns
t _{rl}	Current Rise Time	$\begin{aligned} &\text{I}_{CE} = 30\text{A}, \\ &\text{V}_{CE} = 390\text{V}, \\ &\text{V}_{GE} = 15\text{V}, \\ &\text{R}_{G} = 3\Omega \\ &\text{L} = 200\mu\text{H} \\ &\text{Test Circuit - Figure 26} \end{aligned}$		-	15	-	ns
t _{d(OFF)I}	Current Turn-Off Delay Time			-	92	150	ns
t _{fl}	Current Fall Time			-	88	100	ns
E _{ON1}	Turn-On Energy (Note 2)			-	260	-	μJ
E _{ON2}	Turn-On Energy (Note 2)			-	490	600	μJ
E _{OFF}	Turn-Off Energy (Note 3)			-	575	850	μJ

Thermal Characteristics

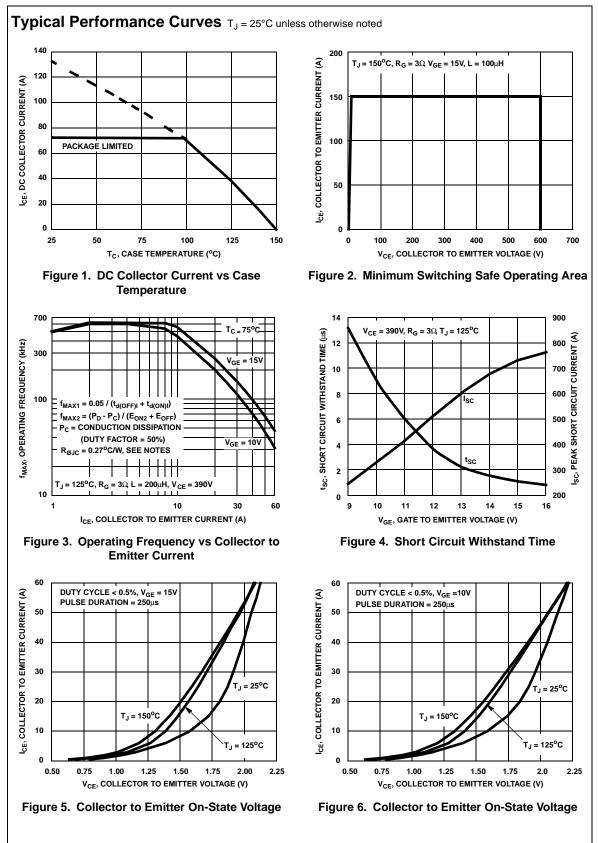
$R_{\theta JC}$	Thermal Resistance Junction-Case	IGBT	-	0.27	°C/W

NOTE

^{2.} Values for two Turn-On loss conditions are shown for the convenience of the circuit designer. E_{DN1} is the turn-on loss of the IGBT only. E_{ON2} is the turn-on loss when a typical diode is used in the test circuit and the diode is at the same T_J as the IGBT. The diode type is specified in figure 26.

^{3.} Turn-Off Energy Loss (E_{OFF}) is defined as the integral of the instantaneous power loss starting at the trailing edge of the input pulse and ending at the point where the collector current equals zero ($I_{CF} = 0A$). All devices were tested per JEDEC Standard No. 24-1 Method for Measurement of Power Device Turn-Off Switching Loss. This test method produces the true total Turn-Off Energy Loss.





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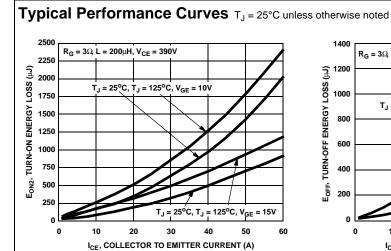


Figure 7. Turn-On Energy Loss vs Collector to Emitter Current

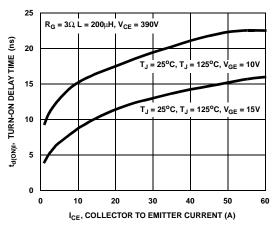


Figure 9. Turn-On Delay Time vs Collector to Emitter Current

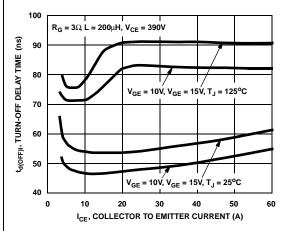


Figure 11. Turn-Off Delay Time vs Collector to Emitter Current

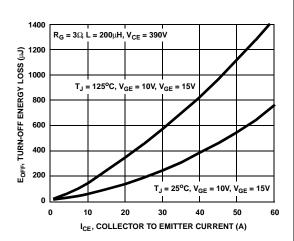


Figure 8. Turn-Off Energy Loss vs Collector to Emitter Current

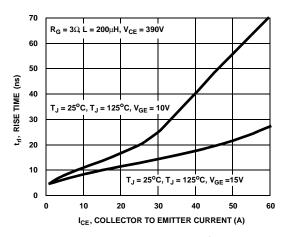


Figure 10. Turn-On Rise Time vs Collector to Emitter Current

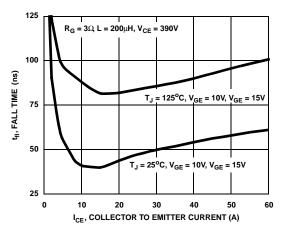
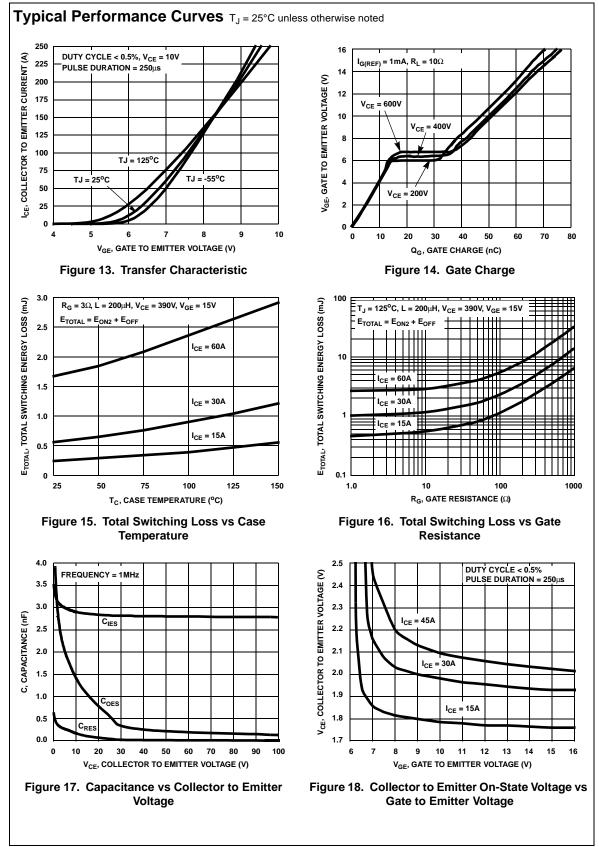


Figure 12. Fall Time vs Collector to Emitter

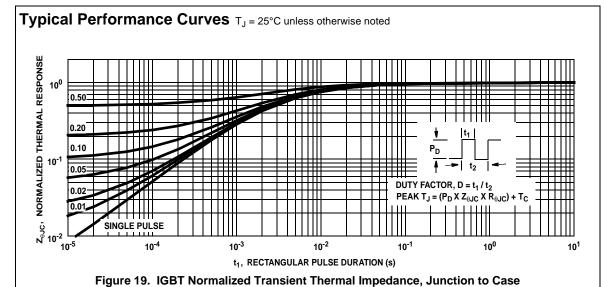
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Test Circuit and Waveforms

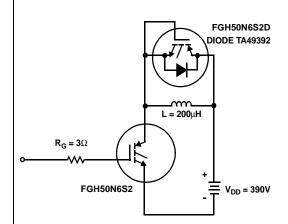


Figure 20. Inductive Switching Test Circuit

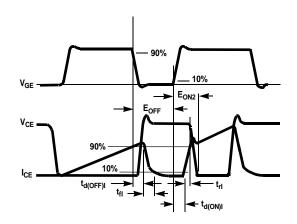


Figure 21. Switching Test Waveforms

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Handling Precautions for IGBTs

Insulated Gate Bipolar Transistors are susceptible to gate-insulation damage by the electrostatic discharge of energy through the devices. When handling these devices, care should be exercised to assure that the static charge built in the handler's body capacitance is not discharged through the device. With proper handling and application procedures, however, IGBTs are currently being extensively used in production by numerous equipment manufacturers in military, industrial and consumer applications, with virtually no damage problems due to electrostatic discharge. IGBTs can be handled safely if the following basic precautions are taken:

- 1. Prior to assembly into a circuit, all leads should be kept shorted together either by the use of metal shorting springs or by the insertion into conductive material such as "ECCOSORBD™ LD26" or equivalent.
- 2. When devices are removed by hand from their carriers, the hand being used should be grounded by any suitable means - for example, with a metallic wristband.
- 3. Tips of soldering irons should be grounded.
- 4. Devices should never be inserted into or removed from circuits with power on.
- 5. Gate Voltage Rating Never exceed the gatevoltage rating of V_{GEM}. Exceeding the rated V_{GE} can result in permanent damage to the oxide layer in the gate region.
- 6. **Gate Termination** The gates of these devices are essentially capacitors. Circuits that leave the gate open-circuited or floating should be avoided. These conditions can result in turn-on of the device due to voltage buildup on the input capacitor due to leakage currents or pickup.
- 7. Gate Protection These devices do not have an internal monolithic Zener diode from gate to emitter. If gate protection is required an external Zener is recommended.

Operating Frequency Information

Operating frequency information for a typical device (Figure 3) is presented as a guide for estimating device performance for a specific application. Other typical frequency vs collector current (I_{CE}) plots are possible using the information shown for a typical unit in Figures 5, 6, 7, 8, 9 and 11. The operating frequency plot (Figure 3) of a typical device shows f_{MAX1} or f_{MAX2}; whichever is smaller at each point. The information is based on measurements of a typical device and is bounded by the maximum rated junction temperature.

 f_{MAX1} is defined by $f_{MAX1} = 0.05/(t_{d(OFF)I} + t_{d(ON)I})$. Deadtime (the denominator) has been arbitrarily held to 10% of the on-state time for a 50% duty factor. Other definitions are possible. $t_{d(OFF)I}$ and $t_{d(ON)I}$ are defined in Figure 27. Device turn-off delay can establish an additional frequency limiting condition for an application other than $T_{JM}.\ t_{d(OFF)I}$ is important when controlling output ripple under a lightly loaded

 f_{MAX2} is defined by $f_{MAX2} = (P_D - P_C)/(E_{OFF} + E_{ON2})$. The allowable dissipation (P_D) is defined by $P_D = (T_{JM} - T_C)/R_{\theta JC}$. The sum of device switching and conduction losses must not exceed PD. A 50% duty factor was used (Figure 3) and the conduction losses (P_C) are approximated by $P_C = (V_{CE} \times I_{CE})/2$.

 $E_{\mbox{\scriptsize ON2}}$ and $E_{\mbox{\scriptsize OFF}}$ are defined in the switching waveforms shown in Figure 27. E_{ON2} is the integral of the instantaneous power loss ($I_{CE} \times V_{CE}$) during turnon and E_{OFF} is the integral of the instantaneous power loss (I_{CE} x V_{CE}) during turn-off. All tail losses are included in the calculation for E_{OFF}; i.e., the collector current equals zero $(I_{CE} = 0)$

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