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TMS470MF04207 TMS470MF03107

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# TMS470MF04207/TMS470MF03107 16/32-Bit RISC Flash Microcontroller

Check for Samples: TMS470MF04207, TMS470MF03107

#### **Features**

- **High-Performance Automotive Grade Microcontroller with Safety Features** 
  - Full Automotive Temperature Range
  - ECC on Flash and SRAM
  - CPU and Memory BIST (Built-In Self Test)
- ARM Cortex™-M3 32-Bit RISC CPU
  - Efficient 1.2 DMIPS/MHz
  - Optimized Thumb2 Instruction Set
  - Memory Protection Unit (MPU)
  - Open Architecture With Third-Party Support
  - Built-In Debug Module
- Operating Features
  - Up to 80MHz System Clock
  - Single 3.3V Supply Voltage
- Integrated Memory
  - 448KB Total Program Flash with ECC
  - Support for Flash EEPROM Emulation
  - 24K-Byte Static RAM (SRAM) with ECC
- Kev Peripherals
  - High-End Timer, MibADC, CAN, MibSPI
- Common TMS470M/570 Platform Architecture
  - Consistent Memory Map across the family
  - Real-Time Interrupt Timer (RTI)
  - Digital Watchdog
  - Vectored Interrupt Module (VIM)
  - Cyclic Redundancy Checker (CRC)
- · Frequency-Modulated Zero-Pin Phase-Locked Loop (FMzPLL)-Based Clock Module
  - Oscillator and PLL clock monitor
- Up to 49 Peripheral IO pins
  - 4 Dedicated GIO w/ External Interrupts

- Two External Clock Prescale (ECP) Modules
  - Programmable Low-Frequency External Clock (ECLK)
  - One Dedicated Pin and One Muxed **ECLK/HET** pin
- Communication Interfaces
  - Two CAN Controllers
    - One with 32 mailboxes, one with 16
    - Parity on mailbox RAM
  - Two Multi-buffered Serial Peripheral Interface (MibSPI)
    - 12 total chip selects
    - 64 buffers with parity on each
  - Two UART (SCI) interfaces
    - H/W Support for Local Interconnect Network (LIN 2.1 master mode)
- High-End Timer (HET)
  - Up to 16 Programmable I/O Channels
  - 128-Word High-End Timer RAM with Parity
- 16-Channel 10-Bit Multi-Buffered ADC (MibADC)
  - 64-Word FIFO Buffer with Parity
  - Single- or Continuous-Conversion Modes
  - 1.55 µs Minimum Sample/Conversion Time
  - Calibration Mode and Self-Test Features
- On-Chip Scan-Base Emulation Logic
  - IEEE Standard 1149.1 (JTAG) Test-Access Port and Boundary Scan
- Packages supported
  - 100-Pin Plastic Quad Flatpack (PZ Suffix)
  - Green/Lead-Free
- Development Tools Available
  - Development Boards
  - Code Composer Studio™ Integrated **Development Environment (IDE)**
  - HET Assembler and Simulator
  - nowFlash™ Flash Programming Tool
- Community Resources
  - TI E2E Community



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### 1.1 PZ Package Views

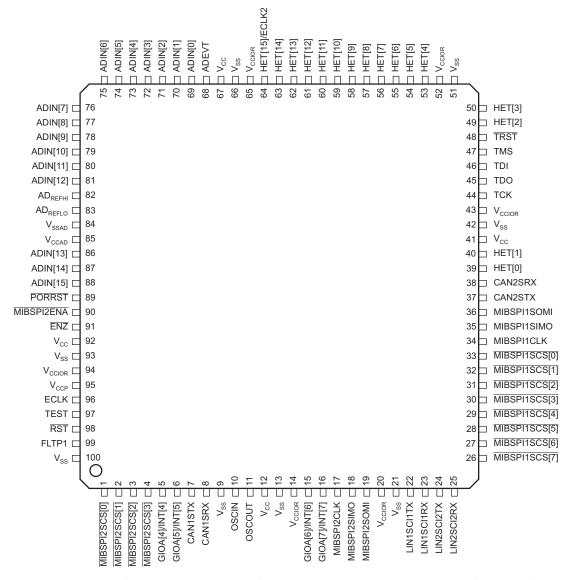


Figure 1-1. TMS470MF04207 and TMS470MF03107 100-Pin PZ Package (Top View)

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### 1.2 Description

The TMS470MF04207/03107 devices are members of the Texas Instruments TMS470M family of Automotive Grade 16/32-bit reduced instruction set computer (RISC) microcontrollers. The TMS470M microcontrollers offer high performance utilizing the high efficiency Cortex™-M3 16/32-bit RISC central processing unit (CPU), resulting in a high instruction throughput while maintaining greater code efficiency. The TMS470M devices utilize the big-endian format where the most-significant byte of a word is stored at the lowest numbered byte and the least-significant byte is stored at the highest numbered byte.

High-end embedded control applications demand more performance from their controllers while maintaining low costs. The TMS470M microcontroller architecture offers solutions to these performance and cost demands while maintaining low power consumption.

The TMS470MF04207/03107 device contains the following:

- 16/32-Bit RISC CPU Core
- TMS470MF04207 Up to 448K-Byte Program Flash with SECDED ECC
- TMS470MF03107 Up to 320K-Byte Program Flash with SECDED ECC
- 64K-Byte Flash with SECDED ECC for additional program space or EEPROM Emulation
- Up to 24K-Byte Static RAM (SRAM) with SECDED ECC
- Real-Time Interrupt Timer (RTI)
- Vectored Interrupt Module (VIM)
- · Hardware built-in self-test (BIST) checkers for SRAM (MBIST) and CPU (LBIST)
- 64-bit Cyclic Redundancy Checker (CRC)
- Frequency-Modulated Zero-Pin Phase-Locked Loop (FMzPLL)-Based Clock Module With Prescaler
- Two Multi-buffered Serial Peripheral Interfaces (MibSPI)
- · Two UARTs (SCI) with Local Interconnect Network Interfaces (LIN)
- Two CAN Controller (DCAN)
- High-End Timer (HET)
- External Clock Prescale (ECP) Module
- One 16-Channel 10-Bit Multi-Buffered ADC (MibADC)
- Error Signaling Module (ESM)
- Four Dedicated General-Purpose I/O (GIO) Pins and 45 Additional Peripheral I/Os (100-Pin Package)

The TMS470M memory includes general-purpose SRAM supporting single-cycle read/write accesses in byte, half-word, and word modes. The SRAM on the TMS470M devices can be protected by means of ECC. This feature utilizes a single error correction and double error detection circuit (SECDED circuit) to detect and optionally correct single bit errors as well as detect all dual bit and some multi-bit errors. This is achieved by maintaining an 8-bit ECC checksum/code for each 64-bit double-word of memory space in a separate ECC RAM memory space.

The flash memory on this device is a nonvolatile, electrically erasable and programmable memory. It is implemented with a 144-bit wide data word (128-bit without ECC) and a 64-bit wide flash module interface. The flash operates with a system clock frequency of up to 28 MHz. Pipeline mode, which allows linear prefetching of flash data, enables a system clock of up to 80 MHz.

The enhanced real-time interrupt (RTI) module on the TMS470M devices has the option to be driven by the oscillator clock. The digital watchdog (DWD) is a 25-bit resetable decrementing counter that provides a system reset when the watchdog counter expires.

The TMS470M devices have six communication interfaces: two LIN/SCIs, two DCANs, and two MibSPIs. The LIN is the Local Interconnect Network standard and also supports an SCI mode. SCI can be used in a full-duplex, serial I/O interface intended for asynchronous communication between the CPU and other peripherals using the standard non-return-to-zero (NRZ) format. The DCAN uses a serial, multimaster communication protocol that efficiently supports distributed real-time control with robust communication



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rates of up to 1 megabit per second (Mbps). The DCAN is ideal for applications operating in noisy and harsh environments (e.g., automotive and industrial fields) that require reliable serial communication or multiplexed wiring. The MibSPI provides a convenient method of serial interaction for high-speed communications between similar shift-register type devices. The MibSPI provides the standard SOMI, SIMO, and SPI clock interface as well as up to eight chip select lines.

The HET is an advanced intelligent timer that provides sophisticated timing functions for real-time applications. The timer is software-controlled, using a reduced instruction set, with a specialized timer micromachine and an attached I/O port. The HET can be used for compare, capture, or general-purpose I/O. It is especially well suited for applications requiring multiple sensor information and drive actuators with complex and accurate time pulses. The TMS470M HET peripheral contains the XOR-share feature. This feature allows two adjacent HET high- resolution channels to be XORed together, making it possible to output smaller pulses than a standard HET.

The TMS470M devices have one 10-bit-resolution, sample-and-hold MibADC. Each of the MibADC channels can be grouped by software for sequential conversion sequences. There are three separate groupings, all three of which can be triggered by an external event. Each sequence can be converted once when triggered or configured for continuous conversion mode.

The frequency-modulated zero-pin phase-locked loop (FMzPLL) clock module contains a phase-locked loop, a clock-monitor circuit, a clock-enable circuit, and a prescaler. The function of the FMzPLL is to multiply the external frequency reference to a higher frequency for internal use. The FMzPLL provides the input to the global clock module (GCM). The GCM module subsequently provides system clock (HCLK), real-time interrupt clock (RTICLK), CPU clock (GCLK), HET clock (VCLK2), DCAN clock (AVCLK1), and peripheral interface clock (VCLK) to all other TMS470M device modules.

The TMS470MF04207/TMS470MF03107 devices also have two external clock prescaler (ECP) modules that when enabled, output a continuous external clock (ECLK). The ECLK1 frequency is a user-programmable ratio of the peripheral interface clock (VCLK) frequency. The second ECLK output can be selected in place of HET15 output. It shares the same source clock as ECLK1 but can be independently programmed for a separate output frequency from ECLK1.

An error signaling module (ESM) provides a common location within the device for error reporting allowing efficient error checking and identification.



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### 1.3 Functional Block Diagram

Figure 1-2 shows the functional block diagram of the TMS470M devices.

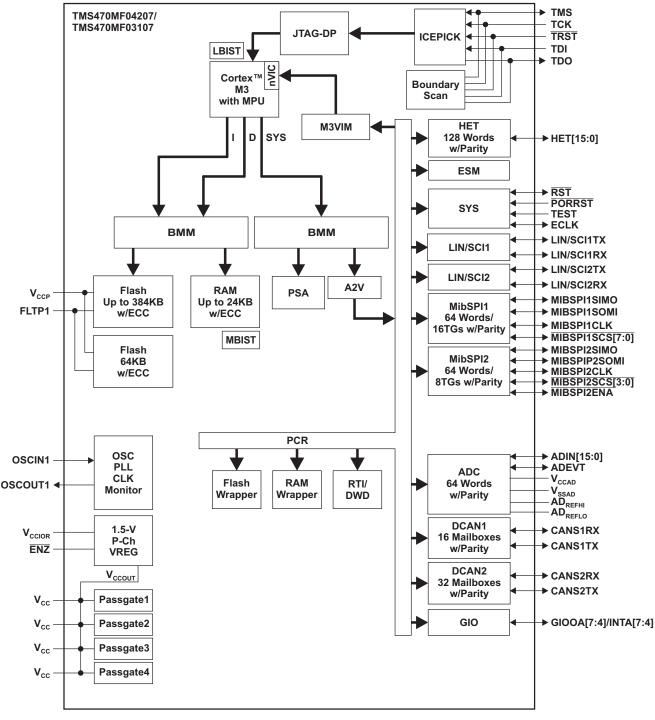


Figure 1-2. TMS470M Functional Block Diagram

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#### 1.4 **Terms and Acronyms**

### Table 1-1. Terms and Acronyms

Terms and Acronyms	Description	Comments
A2V	AHB to VBUSP Bridge	The A2V bridge provides the memory interface between the proprietary TI VBUSP and the ARM AHB bus in the TMS470 platform devices.
ADC	Analog To Digital Converter	
AHB	Advanced High-performance Bus	Part of the M3 core
ВММ	Bus Matrix Master	The BMM provides connectivity between different bus slave modules to different bus master modules. Accesses from different master modules are executed in parallel if no resource conflict occurs or if the master modules are kept in series through arbitration
CRC	Cyclic Redundancy Check Controller	
DAP	Debug Access Port	DAP is an implementation of an ARM Debug Interface.
DCAN	Controller Area Network	
DWD	Digital Watchdog	
ECC	Error Correction Code	
ESM	Error Signaling Module	
GIO	General-Purpose Input/Output	
HET	High-End Timer	
ICEPICK	In Circuit Emulation TAP (Test Access Port) Selection Module	ICEPick can connect or isolate a module level TAP to or from a higher level chip TAP. ICEPick was designed with both emulation and test requirements in mind.
JTAG	Joint Test Access Group	IEEE Committee responsible for Test Access Ports
JTAG-DP	JTAG Debug Port	JTAG-DP contains a debug port state machine (JTAG) that controls the JTAG-DP operation, including controlling the scan chain interface that provides the external physical interface to the JTAG-DP. It is based closely on the JTAG TAP State Machine, see IEEE Std 1149.1-2001.
LBIST	Logic Built-In Self Test	Test the integrity of M3 CPU
LIN	Local Interconnect Network	
M3VIM	Cortex-M3 Vectored Interrupt Manager	
MBIST	Memory Built-In Self Test	Test the integrity of SRAM
MibSPI	Multi-Buffered Serial Peripheral Interface	
MPU	Protection Unit	
NVIC	Nested Vectored Interrupt Controller	Part of the M3 core
OSC	Oscillator	
PCR	Peripheral Central Resource	
PLL	Phase-Locked Loop	
PSA	Parallel Signature Analysis	
RTI	Real-Time Interrupt	
SCI	Serial Communication Interface	
SECDED	Single Error Correction and Double Error Detection	
STC	Self Test Controller	
SYS	System Module	
VBUS	Virtual Bus	One of the protocols that comprises CBA (Common Bus Architecture)
VBUSP	Virtual Bus-Pipelined	One of the protocols that comprises CBA (Common Bus Architecture)
VREG	Voltage Regulator	



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#### **Device Overview**

The TMS470MF04207/03107 device is a TMS470M Platform Architecture implemented in F035 130-nm TI technology. Table 2-1 identifies all the characteristics of the TMS470MF04207/03107 device except the SYSTEM and CPU, which are generic.

**Table 2-1. Device Characteristics** 

CHARACTERISTICS	DEVICE DESCRIPTION TMS470MF04207/03107	COMMENTS FOR TMS470M
MEMORY		
INTERNAL MEMORY	Pipeline/Non-Pipeline 2 Banks with up to 448K-Byte Flash with ECC Up to 24K-Byte SRAM with ECC CRC, 1-channel	Flash is pipeline-capable
PERIPHERALS		
	riority configurations, see Table 3-4. and their peripheral selects, see Ta	
CLOCK	FMzPLL	Frequency-modulated zero-pin PLL has no external loop filter pins.
GENERAL-PURPOSE I/Os	4 I/O	The GIOA port has up to four (4) external pins with external interrupt capability.
LIN/SCI	2 LIN/SCI	
DCAN	2 DCAN	Each with 16/32 mailboxes, respectively.
MibSPI	2 MibSPI	One MibSPI with eight chip select pins, 16 transfer groups, and a 64 word buffer with parity. A second MibSPI with four chip select pins, 1 enable pin, 8 transfer groups, and a 64 word buffer with parity.
pins to share the next high HR sharing is independen externally. If an odd pin is odd pin can only be used		The high-resolution (HR) SHARE feature allows even-numbered HR pins to share the next higher odd-numbered HR pin structures. This HR sharing is independent of whether or not the odd pin is available externally. If an odd pin is available externally and shared, then the odd pin can only be used as a general-purpose I/O. HET RAM with parity checking capability.
HET RAM	128-Instruction Capacity	
MibADC	10-bit, 16-channel 64-word FIFO	MibADC RAM includes parity support.
CORE VOLTAGE	1.55 V	The core voltage is supplied and regulated by the device's internal voltage regulator. There is not need for an externally supplied core voltage.
I/O VOLTAGE	3.3 V	
PINS	100	Available in a 100-pin package.
PACKAGE	PZ (100 pin)	The 100-pin package designator is PZ.

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# 2.1 Memory Map Summary

#### 2.1.1 Memory Map

Figure 2-1 and Figure 2-2 show the TMS470MF04207 and TMS470MF03107 memory maps.

0xFFFFFFF 0xFFF80000	SYSTEM Module
0xFFF7FFF	Peripherals
0xFF000000 0xFEFFFFF	
0xFE000000	PSA
0x08405FFF	
0x08400000	RAM - ECC
0x08105FFF	RAM - CLR Space <sup>(A)</sup> (24KB)
0x08100000	
0x08085FFF	(A) (A ((A))
0x08080000	RAM - SET Space <sup>(A)</sup> (24KB)
0x08005FFF	
0x08003FFF	RAM (24KB)
0x0047FFFF	FLASH - ECC (Bank 1)
0x00440000	FLASH - ECC (Balik 1)
0x0042FFFF 0x00400000	FLASH - ECC (Bank 0)
0x0008FFFF	FLASH (64KB - Bank 1)
0x00080000 0x0005FFFF	TEACH (CHAS Bank I)
UXUUUSFFFF	
	FLASH (384KB - Bank 0)
0x00000000	

A. The RAM supports bit access operation which allows set/clear to dedicated bits without disturbing the other bits; for detailed description, see the Architecture Specification.

Figure 2-1. TMS470MF04207 Memory Map

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0xFFFFFFF 0xFFF80000	SYSTEM Module
0xFFF7FFF	Peripherals
0xFF000000 0xFEFFFFF	PSA
0xFE000000	
0x08403FFF 0x08400000	RAM - ECC
0,000,00000	
0x08103FFF 0x08100000	RAM - CLR Space <sup>(A)</sup> (16KB)
0x08083FFF 0x08080000	RAM - SET Space <sup>(A)</sup> (16KB)
0x08003FFF	RAM (16KB)
0x08000000	KAWI (TOND)
0x00447FFF 0x00440000	FLASH - ECC (Bank 1)
0x0041FFFF 0x00400000	FLASH - ECC (Bank 0)
0x0008FFFF 0x00080000	FLASH (64KB - Bank 1)
0x0003FFFF	
	FLASH (256KB - Bank 0)
0x00000000	

A. The RAM supports bit access operation which allows set/clear to dedicated bits without disturbing the other bits; for detailed description, see the Architecture Specification.

Figure 2-2. TMS470MF03107 Memory Map

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### 2.1.2 Memory Selects

Memories in the TMS470M devices are located at fixed addresses. Table 2-2 through Table 2-7 detail the mapping of the memory regions.

Table 2-2. TMS470MF04207-Specific Memory Frame Assignment

MEMORY FRAME NAME	START ADDRESS	ENDING ADDRESS	MEMORY TYPE	ACTUAL MEMORY
nCS0 <sup>(1)</sup>	0x0000 0000	0x0005 FFFF	Flash	384K Bytes
ncso v	0x0008 0000	0x0008 FFFF	Flash	64K Bytes
RAM-CLR	0x0810 0000	0x0810 5FFF	Internal RAM	24K Bytes
RAM-SET	0x0808 0000	0x0808 5FFF	Internal RAM	24K Bytes
CSRAMO <sup>(1)</sup>	0x0800 0000	0x0800 5FFF	Internal RAM	24K Bytes
CSRAMO	0x0840 0000	0x0840 5FFF	Internal RAM-ECC	24K Bytes

Additional address mirroring could be present resulting in invalid but addressable locations beyond those listed above. TI recommends the use of the MPU for protecting access to addresses outside the intended range of use.

Table 2-3. TMS470MF03107-Specific Memory Frame Assignment

MEMORY FRAME NAME	START ADDRESS	ENDING ADDRESS	MEMORY TYPE	ACTUAL MEMORY
nCS0 <sup>(1)</sup>	0x0000 0000	0x0003 FFFF	Flash	256K Bytes
IIC30 V	0x0008 0000	0x0008 FFFF	Flash	64K Bytes
RAM-CLR	0x0810 0000	0x0810 3FFF	Internal RAM	16K Bytes
RAM-SET	0x0808 0000	0x0808 3FFF	Internal RAM	16K Bytes
CSRAM0 <sup>(1)</sup>	0x0800 0000	0x0800 3FFF	Internal RAM	16K Bytes
CSRAMO	0x0840 0000	0x0840 3FFF	Internal RAM-ECC	16K Bytes

Additional address mirroring could be present resulting in invalid but addressable locations beyond those listed above. TI recommends the use of the MPU for protecting access to addresses outside the intended range of use.

Table 2-4. Memory Initialization and MBIST

CONNECTING MODULE	ADDRESS RANGE		MEMORY INITIALIZATION	MBIST CONTROLLER
CONNECTING MODULE	BASE ADDRESS	ENDING ADDRESS	CHANNEL	ENABLE CHANNEL
System RAM (TMS470MF04207)	0x0800 0000	0x0800 5FFF	0	0
System RAM (TMS470MF03107)	0x0800 0000	0x0800 3FFF	0	0
MibSPI1 RAM	0xFF0E 0000	0xFF0F FFFF	1	1 or 2 <sup>(1)</sup>
MibSPI2 RAM	0xFF0C 0000	0xFF0D FFFF	2	1 or 2 <sup>(1)</sup>
DCAN1 RAM	0xFF1E 0000	0xFF1F FFFF	3	3 or 4 <sup>(1)</sup>
DCAN2 RAM	0xFF1C 0000	0xFF1D FFFF	4	3 01 417
ADC RAM	0xFF3E 0000	0xFF3F FFFF	5	5
HET RAM	0xFF46 0000	0xFF47 FFFF	Not Available	6
STC ROM	Not Applicable	Not Applicable	Not Applicable	7

<sup>(1)</sup> There are single MBIST controllers for both MibSPI RAMs and both DCAN RAMs. The MBIST controller for both MibSPI RAMs is mapped to channels 1 and 2 and the MBIST controller for both DCAN RAMs is mapped to channels 3 and 4. MBIST on these modules can be initiated by selecting one of the 2 channels or both.

Table 2-5. Peripheral Memory Chip Select Assignment

CONNECTING MODULE	ADDRES	PERIPHERAL	
	BASE ADDRESS	ENDING ADDRESS	SELECTS
MibSPI1 RAM	0xFF0E 0000	0xFF0F FFFF	PCS[7]
MibSPI2 RAM	0xFF0C 0000	0xFF0D FFFF	PCS[6]
DCAN1 RAM	0xFF1E 0000	0xFF1F FFFF	PCS[14]
DCAN2 RAM	0xFF1C 0000	0xFF1D FFFF	PCS[15]

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#### Table 2-5. Peripheral Memory Chip Select Assignment (continued)

CONNECTING MODULE	ADDRES	PERIPHERAL	
CONNECTING MODULE	BASE ADDRESS	ENDING ADDRESS	SELECTS
ADC RAM	0xFF3E 0000	0xFF3F FFFF	PCS[31]
HET RAM	0xFF46 0000	0xFF47 FFFF	PCS[35]

#### **NOTE**

All used peripheral memory chip selects should decode down to the smallest possible address for this particular peripheral configuration, starting from 4kB upwards. Unused addresses should generate an illegal address error when accessed.

**Table 2-6. System Peripheral Registers** 

FRAME NAME	ADDRESS RANGE		
FRAME NAME	FRAME START ADDRESS	FRAME ENDING ADDRESS	
PSA	0xFE00 0000	0xFEFF FFFF	
Flash Wrapper Registers	0xFFF8 7000	0xFFF8 7FFF	
PCR Register	0xFFFF E000	0xFFFF E0FF	
System Frame 2 Registers	0xFFFF E100	0xFFFF E1FF	
CPU STC (LBIST)	0xFFFF E400	0xFFFF E4FF	
ESM Register	0xFFFF F500	0xFFFF F5FF	
RAM ECC Register	0xFFFF F900	0xFFFF F9FF	
RTI Register	0xFFFF FC00	0xFFFF FCFF	
VIM Register	0xFFFF FE00	0xFFFF FEFF	
System Registers	0xFFFF FF00	0xFFFF FFFF	

Table 2-7. Peripheral Select Map with Address Range

CONNECTING MODULE	BASE ADDRESS	END ADDRESS	PERIPHERAL SELECTS
MibSPI2	0xFFF7 F600	0xFFF7 F7FF	Delai
MibSPI1	0xFFF7 F400	0xFFF7 F5FF	PS[2]
LIN/SCI1	0xFFF7 E500	0xFFF7 E5FF	DOIGI
LIN/SCI2	0xFFF7 E400	0xFFF7 E4FF	PS[6]
DCAN2	0xFFF7 DE00	0xFFF7 DFFF	DOIOI
DCAN1	0xFFF7 DC00	0xFFF7 DDFF	PS[8]
ADC	0xFFF7 C000	0xFFF7 C1FF	PS[15]
GIO	0xFFF7 BC00	0xFFF7 BCFF	PS[16]
HET	0xFFF7 B800	0xFFF7 B8FF	PS[17]

#### 2.1.3 Flash Memory

When in pipeline mode, the Flash operates with a system clock frequency of up to 80 MHz (versus a system clock in non-pipeline mode of up to 28 MHz). Flash in pipeline mode is capable of accessing 128-bit words and provides four 32-bit pipelined words to the CPU.

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#### **NOTE**

- After a system reset, pipeline mode is disabled [FRDCNTL[2:0] is 000b, see the Flash chapter in the TMS470M Series Technical Reference Manual (literature number SPNU495)]. In other words, the device powers up and comes out of reset in non-pipeline mode.
- The flash external pump voltage (V<sub>CCP</sub>) is required for all operations (program, erase, and read).

### 2.1.4 Flash Program and Erase

The TMS470MF04207/TMS470MF03107 devices flash contain one 384/256K-byte memory array (or bank) and one 64K-byte bank for a total of up to 12 sectors. Table 2-8 and Table 2-9 show the TMS470MF04207 and TMS470MF03107 flash memory banks and sectors.

The minimum size for an erase operation is one sector. The maximum size for a program operation is one 32-bit word.

Table 2-8. TMS470MF04207 Flash Memory Banks and Sectors

SECTOR NO.	SEGMENT	LOW ADDRESS	HIGH ADDRESS	MEMORY ARRAYS (OR BANKS)
0	16k	0x0000 0000	0x0000 3FFF	
1	16k	0x0000 4000	0x0000 7FFF	
2	32k	0x0000 8000	0x0000 FFFF	
3	64k	0x0001 0000	0x0001 FFFF	BANK 0
4	64k	0x0002 0000	0x0002 FFFF	(384K Bytes)
5	64k	0x0003 0000	0x0003 FFFF	
6	64k	0x0004 0000	0x0004 FFFF	
7	64k	0x0005 0000	0x0005 FFFF	
0	16k	0x0008 0000	0x0008 3FFF	
1	16k	0x0008 4000	0x0008 7FFF	BANK 1 <sup>(1)</sup>
2	16k	0x0008 8000	0x0008 BFFF	(64K Bytes)
3	16k	0x0008 C000	0x0008 FFFF	

<sup>(1)</sup> Bank 1 can be used as either EEPROM emulation space or as program space.

#### Table 2-9. TMS470MF03107 Flash Memory Banks and Sectors

SECTOR NO.	SEGMENT	LOW ADDRESS	HIGH ADDRESS	MEMORY ARRAYS (OR BANKS)
0	16k	0x0000 0000	0x0000 3FFF	
1	16k	0x0000 4000	0x0000 7FFF	
2	32k	0x0000 8000	0x0000 FFFF	BANK 0
3	64k	0x0001 0000	0x0001 FFFF	(256K Bytes)
4	64k	0x0002 0000	0x0002 FFFF	
5	64k	0x0003 0000	0x0003 FFFF	
0	16k	0x0008 0000	0x0008 3FFF	
1	16k	0x0008 4000	0x0008 7FFF	BANK 1 <sup>(1)</sup>
2	16k	0x0008 8000	0x0008 BFFF	(64K Bytes)
3	16k	0x0008 C000	0x0008 FFFF	

<sup>(1)</sup> Bank 1 can be used as either EEPROM emulation space or as program space.

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#### 2.2 **Terminal Functions**

The terminal functions table (Table 2-10) identifies the pin names, the associated pin numbers, input voltage, output voltage, whether the pin has any internal pullup/pulldown resistors and a functional pin description. The TMS470MF04207 and TMS470MF03107 devices have the same pin out.

**Table 2-10. Terminal Functions** 

TERMINAL		INPUT		(4)			
NAME	100 PIN	VOLTAGE <sup>(1)</sup>	CURRENT <sup>(3)</sup>	IPU/IPD <sup>(4)</sup>	DESCRIPTION		
HIGH-END TIMER (HET)							
HET[0]	39						
HET[1]	40				Timer input capture or output compare. The HET[15:0] applicable pins can be programmed as		
HET[2]	49				general-purpose input/output (GIO) pins.		
HET[3]	50				The high-resolution (HR) SHARE feature allows		
HET[4]	53				even HR pins to share the next higher odd HR pin		
HET[5]	54				structure. The next higher odd HR pin structure is always implemented, even if the next higher odd		
HET[6]	55				HR pad and/or pin itself is not.		
HET[7]	56	3.3-V I/O	Adaptive impedance 4	Programmable	Note: HET[15] is muxed with ECLK2 output. If		
HET[8]	57	3.3-7 1/0	mA	IPĎ (100 μA)	ECLK2 output is enabled (through SYSPC1 registe at 0xFFFFFF00), ECLK2 is output on this pin and		
HET[9]	58			H N th in	HET[15] becomes an internal only HET channel.		
HET[10]	59				Note: ECLK2 source select must be programmed		
HET[11]	60				the same as ECLK1 due to device specific		
HET[12]	61				implementation details.		
HET[13]	62				<b>Note:</b> ECLK2 is enabled and ECLK2 divider is programmed through ECP control register 1 in		
HET[14]	63				System Frame 2 Registers (0xFFFFE128).		
HET[15]/ECLK2	64						
	1	T	CAN CONTR	OLLER 1 (DCAN1)			
CAN1STX	7	227/1/0	Adaptive	Programmable	DCAN1 transmit pin or GIO pin.		
CAN1SRX	8	3.3-V I/O	impedance 4 mA	IPU (100 μA)	DCAN1 receive pin or GIO pin.		
	•		CAN CONTR	OLLER 2 (DCAN2)			
CAN2STX	37		Adaptive	Programmable	DCAN2 transmit pin or GIO pin		
CAN2SRX	38	3.3-V I/O	impedance 4 mA	IPU (100 μA)	DCAN2 receive pin or GIO pin		
	*	•	GENERAL-P	URPOSE I/O (GIO)			
GIOA[4]/INT[4]	5				General-purpose input/output pins.		
GIOA[5]/INT[5]	6	3.3-V I/O	Adaptive	Programmable	They are interrupt-capable pins.		
GIOA[6]/INT[6]	15	3.3-V I/O	impedance 4 mA	IPĎ (100 μA)			
GIOA[7]/INT[7]	16						

(1) PWR = power, GND = ground, REF = reference voltage, NC = no connect
 (2) All I/O pins, except RST, are configured as inputs while PORRST is low and immediately after PORRST goes high.

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IPD = internal pulldown, IPU = internal pullup (all internal pullups and pulldowns are inactive on input pins when PORRST is asserted)

The TMS470M device utilizes adaptive impedance 4 mA buffers that default to an adaptive impedance mode of operation. As a fail-safe, the adaptive impedance features of the buffer may be disabled and revert the buffer to a standard buffer mode.

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### Table 2-10. Terminal Functions (continued)

TERMINAL		INPUT	o. = 4 o = (1)   OUIPUI  (10 p. (4)																
NAME	100 PIN	VOLTAGE <sup>(1)</sup>	CURRENT <sup>(3)</sup>	IPU/IPD <sup>(4)</sup>	DESCRIPTION														
		MULTI-BUFFER	ED SERIAL PE	RIPHERAL INTER	FACE 1 (MIBSPI1)														
MIBSPI1CLK	34				MIBSPI1 clock. MIBSPI1CLK can be programmed as a GIO pin.														
MIBSPI1SCS[0]	33																		
MIBSPI1SCS[1]	32																		
MIBSPI1SCS[2]	31																		
MIBSPI1SCS[3]	30				MIBSPI1 slave chip select. MIBSPI1SCS[7:0] can														
MIBSPI1SCS[4]	29	3.3-V I/O	Adaptive impedance 4	Programmable	be programmed as a GIO pins.														
MIBSPI1SCS[5]	28	0.0 1 1/0	mA	IPŪ (100 μA)															
MIBSPI1SCS[6]	27																		
MIBSPI1SCS[7]	26																		
MIBSPI1SIMO	35				MIBSPI1 data stream. Slave in/master out. MIBSPI1SIMO can be programmed as a GIO pin.														
MIBSPI1SOMI	36				MIBSPI1 data stream. Slave out/master in. MIBSPI1SOMI can be programmed as a GIO pin.														
		MULTI-BUFFER	ED SERIAL PE	RIPHERAL INTER	FACE 2 (MibSPI2)														
MibSPI2CLK	17			Adaptive Programmable	MibSPI2 clock. MibSPI2CLK can be programmed as a GIO pin.														
MibSPI2SCS[0]	1																		
MibSPI2SCS[1]	2				MibSPI2 slave chip select MibSPI2SCS[3:0] can be														
MibSPI2SCS[2]	3				programmed as GIO pins.														
MibSPI2SCS[3]	4		Adaptive																
MibSPI2ENA	90	3.3-V I/O	impedance 4		MibSPI2 enable pin. MibSPI2ENA can be programmed as a GIO pin.														
MibSPI2SIMO[0]	18																		MibSPI2 data stream. Slave in/master out. MibSPI2SIMO pins can be programmed as a GIO pins.
MibSPI2SOMI[0]	19				MibSPI2 data stream. Slave out/master in. MibSPI2SOMI pins can be programmed as GIO pins.														
	LOCAL INT	TERCONNECT N	NETWORK/SER	IAL COMMUNICA	TIONS INTERFACE (LIN/SCI)														
LIN/SCI1RX	23	3.3-V I/O	Adaptive	Programmable	LIN/SCI1 data receive. Can be programmed as a GIO pin.														
LIN/SCI1TX	22	3.3-7 1/0	impedance 4 mA	ΙΡŬ (100 μA)	LIN/SCI1 data transmit. Can be programmed as a GIO pin.														
LIN/SCI2RX	25	227/1/0	Adaptive	Programmable	LIN/SCI2 data receive. Can be programmed as a GIO pin.														
LIN/SCI2TX	24	3.3-V I/O	impedance 4 mA	ΙΡŬ (100 μΑ)	LIN/SCI2 data transmit. Can be programmed as a GIO pin.														
		MULTI-BUFFER	ED ANALOG-T	O-DIGITAL CONV	ERTER (MIBADC)														
ADEVT	68	3.3-V I/O	Adaptive impedance 4 mA	Programmable IPD (100 μA)	MibADC event input. Can be programmed as a GIO pin.														



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#### Table 2-10. Terminal Functions (continued)

TERMINAL		INPUT	OUTPUT	(4)			
NAME	100 PIN	VOLTAGE <sup>(1)</sup>	CURRENT <sup>(3)</sup>	IPU/IPD <sup>(4)</sup>	DESCRIPTION		
ADIN[0]	69						
ADIN[1]	70						
ADIN[2]	71						
ADIN[3]	72						
ADIN[4]	73						
ADIN[5]	74						
ADIN[6]	75						
ADIN[7]	76	0.014					
ADIN[8]	77	3.3 V			MibADC analog input pins.		
ADIN[9]	78						
ADIN[10]	79						
ADIN[11]	80						
ADIN[12]	81						
ADIN[13]	86						
ADIN[14]	87						
ADIN[15]	88						
AD <sub>REFHI</sub>	82	3.3-V REF			MibADC module high-voltage reference input.		
AD <sub>REFLO</sub>	83	GND REF			MibADC module low-voltage reference input.		
V <sub>CCAD</sub>	85	3.3-V PWR			MibADC analog supply voltage.		
V <sub>SSAD</sub>	84	GND			MibADC analog ground reference.		
		+	OSCILL	ATOR (OSC)			
OSCIN	10	1.55-V I			Crystal connection pin or external clock input.		
OSCOUT	11	1.55-V O			External crystal connection pin.		
	<del></del>		SYSTEM	MODULE (SYS)			
PORRST	89	3.3-V I		IPD (100 μA)	Input master chip power-up reset. External V <sub>CC</sub> monitor circuitry must assert a power-on reset.		
RST	98	3.3-V I/O	Adaptive impedance 4 mA	IPU (100 μA)	Bidirectional reset. The internal circuitry can assert a reset, and an external system reset can assert a device reset.  On this pin, the output buffer is implemented as an open drain (drives low only).  To ensure an external reset is not arbitrarily generated, TI recommends that an external pullup resistor be connected to this pin.		
ECLK	96	3.3-V I/O	Adaptive impedance 4 mA	Programmable IPD (100 μA)	Bidirectional pin. ECLK can be programmed as a GIO pin.		
			TEST/E	DEBUG (T/D)			
TCK	44	3.3-V I		IPD (100 μA)	Test clock. TCK controls the test hardware (JTAG).		
TDI	46			IPU (100 μA)	Test data in pin. TDI inputs serial data to the test instruction register, test data register, and programmable test address (JTAG).		
TDO	45	3.3-V I/O	Adaptive impedance 4 mA	IPD (100 μA)	Test data out pin. TDO outputs serial data from the test instruction register, test data register, identification register, and programmable test address (JTAG).		
TMS	47			IPU (100 μA)	Serial input pin for controlling the state of the CPU test access port (TAP) controller (JTAG).		
TRST	48	3.3-V I		IPD (100 μA)	Test hardware reset to TAP. IEEE Standard 1149-1 (JTAG) Boundary-Scan Logic.		

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### Table 2-10. Terminal Functions (continued)

TERMII	NAL	INPUT	OUTPUT	(4)	
NAME	100 PIN	VOLTAGE <sup>(1)</sup>	CURRENT <sup>(3)</sup>	IPU/IPD <sup>(4)</sup>	DESCRIPTION
TEST	97	3.3-V I		IPD (100 μA)	Test enable. Reserved for internal use only. TI recommends that this pin be connected to ground or pulled down to ground by an external resistor.
ENZ	91				Enables/disables the internal voltage regulator.
					0V - Enables internal voltage regulator.
		3.3-V I		IPD (100 μA)	3.3V-Disables internal voltage regulator.
					Note: The ENZ pin is provided to facilitate testing across the core voltage range and is not intended for disabling the on chip voltage regulator during application use.
		_	F	FLASH	
FLTP1	99				Flash Test Pad 1 pin. For proper operation, this pin must connect only to a test pad or not be connected at all [no connect (NC)]. The test pad must not be exposed in the final product where it might be subjected to an ESD event.
V <sub>CCP1</sub>	95				Flash external pump voltage (3.3 V). This pin is
V <sub>CCP2</sub>	95	3.3-V PWR			required for both Flash read and Flash program ar erase operations. V <sub>CCP1</sub> and V <sub>CCP2</sub> are double bonded to the same pin.
			SUPPLY VOLT	TAGE CORE (1.55	V)
V <sub>CC</sub>	12				Vreg output voltage when Vreg is enabled. V <sub>CC</sub>
	41	1.55-V PWR			input when Vreg is disabled.
	67	1.00 V 1 VVIC			
	92				
		SUPPLY VO	DLTAGE DIGIT	AL I/O AND REGU	LATOR (3.3 V)
V <sub>CCIOR</sub>	14	1			
	20	_			
	43 52	3.3-V PWR			Digital I/O and internal regulator supply voltage.
	65	-			
	94	_			
	_ · · · · ·	1	SUPPI	LY GROUND	1
V <sub>SS</sub>	9				
	13	1			
	21	=			
	42	GND			District I/O and care county are a set-
	51				Digital I/O and core supply ground reference.
	66				
	93				
	100				

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#### 2.3 Device Support

#### 2.3.1 Device and Development-Support Tool Nomenclature

To designate the stages in the product development cycle, TI assigns prefixes to the part numbers of all devices and support tools. Each commercial family member has one of three prefixes: TMX, TMP, or TMS (e.g.,TMS470MF04207). Texas Instruments recommends two of three possible prefix designators for its support tools: TMDX and TMDS. These prefixes represent evolutionary stages of product development from engineering prototypes (TMX/TMDX) through fully qualified production devices/tools (TMS/TMDS).

Device development evolutionary flow:

TMX Experimental device that is not necessarily representative of the final device's electrical

specifications.

TMP Final silicon die that conforms to the device's electrical specifications but has not completed

quality and reliability verification.

**TMS** Fully-qualified production device.

Support tool development evolutionary flow:

TMDX Development-support product that has not yet completed Texas Instruments internal

qualification testing.

**TMDS** Fully qualified development-support product.

TMX and TMP devices and TMDX development-support tools are shipped against the following disclaimer:

"Developmental product is intended for internal evaluation purposes."

TMS devices and TMDS development-support tools have been characterized fully, and the quality and reliability of the device have been demonstrated fully. Tl's standard warranty applies.

Predictions show that prototype devices (TMX or TMP) have a greater failure rate than the standard production devices. Texas Instruments recommends that these devices not be used in any production system because their expected end-use failure rate still is undefined. Only qualified production devices are to be used.

TI device nomenclature also includes a suffix with the device family name. This suffix indicates the package type (for example, PZ), the temperature range (for example, "Blank" is the commercial temperature range), and the device speed range in megahertz.

Figure 2-3 illustrates the numbering and symbol nomenclature for the TMS470M family.

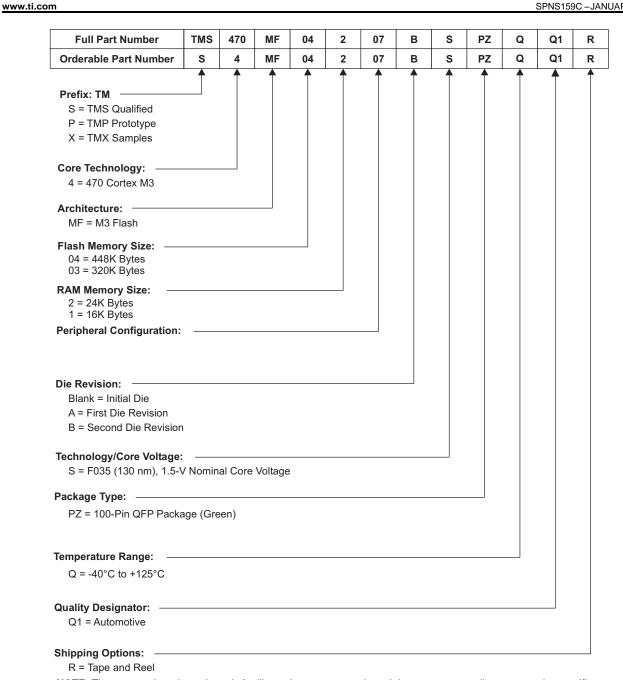
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NOTE: The part number given above is for illustrative purposes only and does not necessarily represent the specific part number or silicon revision to which this document applies.

Figure 2-3. TMS470M Device Numbering Conventions



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### 3 Device Configurations

#### 3.1 Reset/Abort Sources

Resets/aborts are handled as shown in Table 3-1.

#### Table 3-1. Reset/Abort Sources

ERROR SOURCE	SYSTEM MODE	ERROR RESPONSE	ESM HOOKUP, GROUP.CHANNEL
1) CPU TRANSACTIONS			
Precise write error (NCNB/Strongly Ordered)	User/Privilege	Precise Abort (CPU)	n/a
Precise read error (NCB/Device or Normal)	User/Privilege	Precise Abort (CPU)	n/a
Imprecise write error (NCB/Device or Normal)	User/Privilege	Imprecise Abort (CPU)	n/a
External imprecise error (Illegal transaction with ok response)	User/Privilege	ESM	2.17
Illegal instruction	User/Privilege	Undefined Instruction Trap (CPU) <sup>(1)</sup>	n/a
M3 Lockup	User/Privilege	ESM => NMI	2.16
MPU access violation	User/Privilege	Abort (CPU)	n/a
2) SRAM			
ECC single error (correctable)	User/Privilege	ESM	1.26
ECC double error (uncorrectable)	User/Privilege	ESM => NMI	2.6
3) FLASH WITH ECC			
ECC single error (correctable)	User/Privilege	ESM	1.6
ECC double error (uncorrectable)	User/Privilege	ESM => NMI	2.4
8) HET			
HET Memory parity error	User/Privilege	ESM	1.7
9) MIBSPI			
MibSPI1 memory parity error	User/Privilege	ESM	1.17
MibSPI2 memory parity error	User/Privilege	ESM	1.18
10) MIBADC			
Memory parity error	User/Privilege	ESM	1.19
11) DCAN/CAN			
DCAN1 memory parity error	User/Privilege	ESM	1.21
DCAN2 memory parity error	User/Privilege	ESM	1.23
13) PLL			
PLL slip error	User/Privilege	ESM	1.10
14) CLOCK MONITOR			
Clock monitor interrupt	User/Privilege	ESM	1.11
19) VOLTAGE REGULATOR	<u> </u>	<u>-</u>	
Vcc out of range	n/a	Reset	n/a
20) CPU SELFTEST (LBIST)			I
CPU Selftest (LogicBIST) error	User/Privilege	ESM	1.27
21) ERRORS REFLECTED IN THE SYSESR REGIS	TER		
Power-Up Reset/Vreg out of voltage <sup>(2)</sup>	n/a	Reset	n/a

<sup>(1)</sup> The undefined instruction trap is NOT detected outside of the CPU. The trap is taken only if the code reaches the execute stage of the CPU.

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<sup>(2)</sup> Both a power-on reset and Vreg out-of-range reset are indicated by the PORST bit in the SYSESR register.

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#### Table 3-1. Reset/Abort Sources (continued)

·			
ERROR SOURCE	SYSTEM MODE	ERROR RESPONSE	ESM HOOKUP, GROUP.CHANNEL
Oscillator fail / PLL slip (3)	n/a	Reset	n/a
M3 Lockup/LRM	n/a	Reset	n/a
Watchdog time limit exceeded	n/a	Reset	n/a
CPU Reset	n/a	Reset	n/a
Software Reset	n/a	Reset	n/a
External Reset	n/a	Reset	n/a

<sup>(3)</sup> Oscillator fail/PLL slip can be configured in the system register (SYS.PLLCTL1) to generate a reset.

#### 3.2 Lockup Reset Module

The lockup reset module (LRM) is implemented to communicate a lockup condition by the core. The LRM provides a small watchdog timer which can generate a system reset in case a lockup condition that is identified by the core cannot be cleared by software.

### 3.3 ESM Assignments

The ESM module is intended for the communication critical system failures in a central location. The error indication is by an error interrupt when the failure is recognized from any detection unit. The ESM module consist of three error groups with 32 inputs each. The generation of the interrupts is shown in Table 3-2. ESM assignments are listed in Table 3-3.

Table 3-2. ESM Groups

ERROR GROUP	INTERRUPT, LEVEL
Group1	maskable, low/high
Group2	non-maskable, high
Group3	Not Used

Table 3-3. ESM Assignments

ERROR SOURCES	CHANNEL					
GROUP 1						
Reserved	0 - 5					
Flash - ECC Single Bit	6					
HET memory parity error	7					
Reserved	8-9					
PLL Slip Error	10					
Clock Monitor interrupt	11					
Reserved	12-16					
MibSPI1 memory parity error	17					
MibSPI2 memory parity error	18					
MibADC memory parity error	19					
Reserved	20					
DCAN1 memory parity error	21					
Reserved	22					
DCAN2 memory parity error	23					
Reserved	24-25					
SRAM - single bit	26					
CPU LBIST - selftest error	27					

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Table 3-3. ESM Assignments (continued)

ERROR SOURCES	CHANNEL
Reserved	28-31
GROUP 2	
Reserved	0-3
Flash - Double-Bit Error (uncorrectable)	4
Reserved	5
SRAM - Double-Bit Error (uncorrectable)	6
Reserved	7-15
M3 Lockup	16
M3 External Imprecise Abort	17
Reserved	18-31

### 3.4 Interrupt Priority (M3VIM)

The TMS470M platform interrupt architecture includes a vectored interrupt manager (M3VIM) that provides hardware assistance for prioritizing and controlling the many interrupt sources present on a device. Table 3-4 communicates the default interrupt request assignments.

**Table 3-4. Interrupt Request Assignments** 

MODULES	INTERRUPT SOURCES	DEFAULT VIM INTERRUPT REQUEST
ESM	ESM High level interrupt (NMI)	0
Reserved	(NMI)	1
ESM	ESM Low level interrupt	2
SYSTEM	Software interrupt (SSI)	3
RTI	RTI compare interrupt 0	4
RTI	RTI compare interrupt 1	5
RTI	RTI compare interrupt 2	6
RTI	RTI compare interrupt 3	7
RTI	RTI overflow interrupt 0	8
RTI	RTI overflow interrupt 1	9
Reserved	Reserved	10
GIO	GIO Interrupt A	11
GIO	GIO Interrupt B	12
HET	HET level 0 interrupt	13
HET	HET level 1 interrupt	14
MibSPI1	MibSPI1 level 0 interrupt	15
MibSPI1	MibSPI1 level 1 interrupt	16
Reserved	Reserved	17
LIN/SCI2	LIN/SCI2 level 0 interrupt	18
LIN/SCI2	LIN/SCI2 level 1 Interrupt	19
LIN/SCI1	LIN/SCI1 level 0 interrupt	20
LIN/SCI1	LIN/SCI1 level 1 Interrupt	21
DCAN1	DCAN1 level 0 Interrupt	22
DCAN1	DCAN1 level 1 Interrupt	23
ADC	ADC event group interrupt	24
ADC	ADC sw group 1 interrupt	25
ADC	ADC sw group 2 interrupt	26

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#### Table 3-4. Interrupt Request Assignments (continued)

MODULES	INTERRUPT SOURCES	DEFAULT VIM INTERRUPT REQUEST
MibSPI2	MibSPI2 level 0 interrupt	27
MibSPI2	MibSPI2 level 1 interrupt	28
DCAN2	DCAN2 level 0 interrupt	29
DCAN2	DCAN2 level 1 interrupt	30
ADC	ADC magnitude threshold interrupt	31
Reserved	Reserved	32
Reserved	Reserved	33
DCAN1	DCAN1 IF3 interrupt	34
DCAN2	DCAN2 IF3 interrupt	35
Reserved	Reserved	36-47

#### 3.5 MibADC

The multi-buffered analog-to-digital converter (MibADC) accepts an analog signal and converts the signal to a 10-bit digital value.

The TMS470M MibADC module stores its digital results in one of three FIFO buffers. There is one FIFO buffer for each conversion group [event, group1 (G1), and group2 (G2)], and the total MibADC FIFO on the device is divided amongst these three regions. The size of the individual group buffers are software programmable. MibADC buffers can be serviced by interrupts.

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#### 3.5.1 MibADC Event Triggers

All three conversion groups can be configured for event-triggered operation, providing up to three event-triggered groups.

The trigger source and polarity can be selected individually for group 1, group 2 and the event group from the options identified in Table 3-5.

Table 3-5. MibADC Event Hookup Configuration

EVENT NO.	SOURCE SELECT BITS for G1 or EVENT (G1SRC[2:0] or EVSRC[2:0])	SIGNAL PIN NAME
1	000	ADEVT
2	001	HET[1]
3	010	HET[3]
4	011	HET[16] <sup>(1)</sup>
5	100	HET[18] <sup>(1)</sup>
6	101	HET[24] <sup>(1)</sup>
7	110	HET[26] <sup>(1)</sup>
8	111	HET[28] <sup>(1)</sup>

<sup>(1)</sup> These channels are available as internal signals even if they are not included as pins (Section 1.1).

#### 3.6 MibSPI

The multi-buffered serial peripheral interface module allows CPU independent SPI communications with system peripherals.

The MibSPI1 module can support up to 16 transfer groups and 8 chip selects. In addition, up to 4 data formats can be supported allowing assignment of various formats to each transfer group.

The MibSPI2 module can support up to 8 transfer groups, 4 chip selects, and up to 4 data formats.

#### 3.6.1 MibSPI Event Trigger

The MibSPI module has the ability to automatically trigger SPI events based on internal and external event triggers.

The trigger sources can be selected individually for each transfer group from the options identified in Table 3-6.

Table 3-6. MibSPI1 and MibSPI2 Event Hookup Configuration

EVENT NO.	SOURCE SELECT BITS FOR MIBSPI EVENTS TGXCTRL TRIGSRC[3:0]	SIGNAL PIN NAME
Disabled	0000	No trigger source
EVENT0	0001	GIOA[0] <sup>(1)</sup>
EVENT1	0010	GIOA[1] <sup>(1)</sup>
EVENT2	0011	GIOA[2] <sup>(1)</sup>
EVENT3	0100	GIOA[3] <sup>(1)</sup>
EVENT4	0101	GIOA[4]
EVENT5	0110	GIOA[5]
EVENT6	0111	HET[20] <sup>(1)</sup>
EVENT7	1000	HET[21] <sup>(1)</sup>
EVENT8	1001	HET[22] <sup>(1)</sup>

<sup>(1)</sup> These channels are available as internal signals even if they are not included as pins (Section 1.1).

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Table 3-6. MibSPI1 and MibSPI2 Event Hookup Configuration (continued)

EVENT NO.	SOURCE SELECT BITS FOR MIBSPI EVENTS TGXCTRL TRIGSRC[3:0]	SIGNAL PIN NAME
EVENT9	1010	HET[23] <sup>(1)</sup>
EVENT10	1011	HET[28] <sup>(1)</sup>
EVENT11	1100	HET[29] <sup>(1)</sup>
EVENT12	1101	HET[30] <sup>(1)</sup>
EVENT13	1110	HET[31] <sup>(1)</sup>
EVENT14	1111	Internal Tick Counter

#### **JTAG ID** 3.7

The 32-bit JTAG ID code for this device is 0x0B8D802F.

#### **Scan Chains** 3.8

The device contains an ICEPICK module to access the debug scan chains; see Figure 3-1. Debug scan chain #0 handles the access to the CPU. The ICEPICK scan ID is 0x00366D05, which is the same as the device ID.

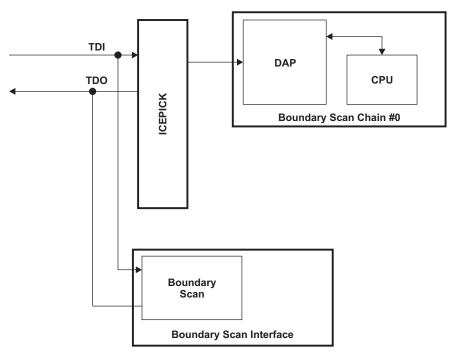


Figure 3-1. Debug Scan Chains

#### 3.9 Adaptive Impedance 4 mA IO Buffer

The adaptive impedance 4 mA buffer is a buffer that has been explicitly designed to address the issue of decoupling EMI sources from the pins which they drive. This is accomplished by adaptively controlling the impedance of the output buffer and should be particularly effective with capacitive loads.

The adaptive impedance 4 mA buffer features two modes of operation: Impedance Control Mode, and Low-Power Mode/Standard Buffer Mode as defined below:

Impedance Control Mode is enabled in the design by default. This mode adaptively controls the impedance of the output buffer.



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Standard Buffer Mode is used to configure the buffer back into a generic configuration. This buffer
mode is used when it is necessary to drive the output at very high speeds, or when EMI reduction is
not a concern.

Table 3-7. Adaptive Impedance 4 mA Buffer Mode Availability

MODULE OR PIN NAME	STANDARD BUFFER ENABLE (SBEN) <sup>(1)</sup>
SYS.ECLK	GPREG1.0
SYS.nRST	GPREG1.1
SYS.TDI/TDO	Standard Buffer Enabled
SYS.TMSC	Standard Buffer Enabled
HET	GPREG1.2
SCI1	GPREG1.3
LIN/SCI2	GPREG1.4
MIBSPI1	GPREG1.5
MibSPI2	GPREG1.6
Reserved	GPREG1.7
MIBADC.ADEVT	GPREG1.8
DCAN1	GPREG1.9
DCAN2	GPREG1.10
GIOA	GPREG1.11

<sup>(1)</sup> SBEN configuration can be achieved using the GPREG register within the system frame(0xFFFFFA0).

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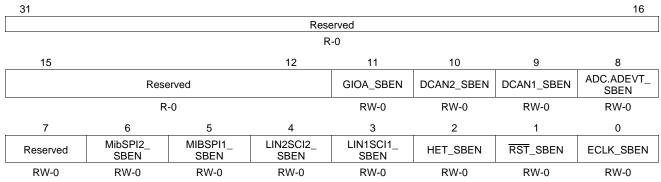
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### 3.9.1 Standard Buffer Enable Register (GPREG1)

A general purpose register with the system frame has been utilized to control the enabling of standard buffer mode. This register is shown in Figure 3-2 and described in Table 3-8

#### NOTE

In general, all device registers are defined within the TRM (SPNU450); however, in cases where the register definition is device specific, the register is defined within the device specific datasheet.



LEGEND: R/W = Read/Write; R = Read only; -n = value after reset

Figure 3-2. General-Purpose Register 1 (GPREG1)

#### Table 3-8. General-Purpose Register 1 (GPREG1) Field Descriptions

Bit	Field	Value	Description
31-12	Reserved		These bits are reserved. Reads return 0 and writes have no effect.
11	1 GIOA_SBEN		GIOA port standard buffer enable bit. This bit enables/disables standard buffer mode for all GIOA pins
		0	Standard buffer mode is not enabled.
		1	Standard buffer mode is enabled for all associated module pins.
10	DCAN2_SBEN		DCAN2 standard buffer enable bit. This bit enables/disables standard buffer mode for all DCAN2 pins.
		0	Standard buffer mode is not enabled.
		1	Standard buffer mode is enabled for all associated module pins.
9	DCAN1_SBEN		DCAN1 standard buffer enable bit. This bit enables/disables standard buffer mode for all DCAN1 pins.
		0	Standard buffer mode is not enabled.
		1	Standard buffer mode is enabled for all associated module pins.
8	ADC.ADEVT_SBEN		ADC.ADEVT standard buffer enable bit. This bit enables/disables standard buffer mode for the ADC.ADEVT pin.
		0	Standard buffer mode is not enabled.
		1	Standard buffer mode is enabled for the ADEVT pin.
7	Reserved		These bits are reserved. Reads return 0 and writes have no effect.
6	MibSPI2_SBEN		MibSPI2 standard buffer enable bit. This bit enables/disables standard buffer mode for all MibSPI2 pins.
		0	Standard buffer mode is not enabled.
		1	Standard buffer mode is enabled for all associated module pins.
5	MIBSPI1		MIBSPI1 standard buffer enable bit. This bit enables/disables standard buffer mode for all MIBSPI1 pins.
		0	Standard buffer mode is not enabled.
		1	Standard buffer mode is enabled for all associated module pins.

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Table 3-8. General-Purpose Register 1 (GPREG1) Field Descriptions (continued)

Bit	Field	Value	Description
4	LIN2SCI2_SBEN		LIN/SCI2 standard buffer enable bit. This bit enables/disables standard buffer mode for all LIN/SCI2 pins.
		0	Standard buffer mode is not enabled.
		1	Standard buffer mode is enabled for all associated module pins.
3	LIN1SCI1_SBEN		LIN/SCI1 standard buffer enable bit. This bit enables/disables standard buffer mode for all LIN/SCI1 pins.
		0	Standard buffer mode is not enabled.
		1	Standard buffer mode is enabled for all associated module pins.
2	HET_SBEN		HET standard buffer enable bit. This bit enables/disables standard buffer mode for all HET pins.
		0	Standard buffer mode is not enabled.
		1	Standard buffer mode is enabled for all associated module pins.
1	RST_SBEN		RST standard buffer enable bit. This bit enables/disables standard buffer mode for the RST pin.
		0	Standard buffer mode is not enabled.
		1	Standard buffer mode is enabled for the RST pin.
0	ECLK_SBEN		ECLK standard buffer enable bit. This bit enables/disables standard buffer mode for the ECLK pin.
		0	Standard buffer mode is not enabled.
		1	Standard buffer mode is enabled for the ECLK pin.

### 3.9.2 Coresight Components/Debug ROM

Coresight registers are memory-mapped and accessible via the CPU and JTAG.

Table 3-9. Debug Component Memory Map

COMPONENT	FRAME START ADDRESS	FRAME END ADDRESS	FRAME SIZE	MEMORY TYPE
	N	13 INTEGRATION F	RAME	
DWT	0xE000_1000	0xE000_1FFF	4K	
FPB	0xE000_2000	0xE000_2FFF	4K	Control Registers for
NVIC	0xE000_E000	0xE000_EFFF	4K	debug and trace modules
Debug ROM 1	0xE00F_F000	0xE00F_FFFF	4K	

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#### Table 3-10. Debug ROM contents for Debug ROM 1 (M3 ROM)

ADDRESS OFFSET see Table 3-9	DESCRIPTION	VALUE
0x000	NVIC	0xFFF0_F003
0x004	DWT	0xFFF0_2003
0x008	FPB	0xFFF0_3003
0x00C	ITM	0xFFF0_1003
0x010	TPIU <sup>(1)</sup>	0xFFF4_1002
0x014	ETM <sup>(1)</sup>	0xFFF4_2002
0x018	End of Table	0x0000_0000

<sup>(1)</sup> Cortex™-M3 debug ROM always will have entries for optional components TPIU and ETM. Whether or not these components are present is determined by bit number 0 of the entry value.

### 3.10 Built-In Self Test (BIST) Features

#### 3.10.1 STC/LBIST

The TMS470M family supports a logic built-in self test (LBIST or CPUBIST) of the M3 CPU.

LBIST testing can be performed in two modes of operation:

- Full Execution. In this mode, the full suite of test patterns is run without interruption. This test is started via CPU control and is well suited for use at device start up.
- Cyclic Execution. During cyclic execution, a small percentage of time will be dedicated to running a subset of the self-test (STC Intervals). This mode is well suited for executing on a periodic basis to minimize the bandwidth use. After all STC intervals are executed, all test patterns will have been run.

#### NOTE

- 1. The application will need to disable peripherals and or interrupts to avoid missing interrupts.
- 2. No debugger interaction is possible with the CPU during self test. This includes access to memory and registers since access is through the CPU.

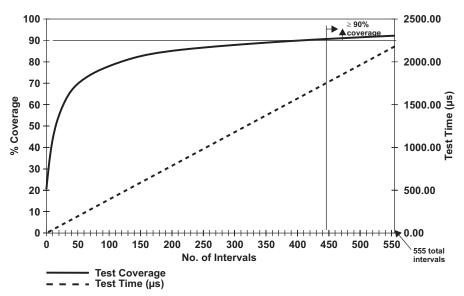
The default value of the LBIST clock prescaler (STCDIV) is divide-by-1 and the device will support STC frequencies up to and including HCLK frequency. In order to minimize the current consumption during LBIST execution, the LBIST clock prescalar (STCDIV) may be configured to reduce the LBIST frequency.

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- A. A single LBIST interval is 158 STC CLK cycles in duration, excluding clock transition timing of 20 cycles.
- B. This device has 555 total intervals.

Figure 3-3. CPU BIST Intervals vs Coverage

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#### 3.10.2 MBIST

The TMS470M supports memory built-in self test (MBIST) of the SRAM. The MBIST is accessible via the application in order to facilitate memory self test by the application by enabling the MBIST controllers associated with the specific RAMs to be tested. (For device-specific MBIST controller assignments, see Table 2-4.)

#### The MBIST controller:

- · Supports testing of all system and peripheral RAM.
- · Captures the MBIST results in the MBIST status register (MSTFAIL).
- Supports execution of each Memory BIST controller in parallel (MSINENA).
   For MSIENA bit assignments, see Table 2-4
- Supports execution of each Memory BIST controller individually (MSINENA).
   For MSIENA bit assignments, see Table 2-4

The MBIST controller selection is mapped to the MBIST controller/memory initialization enable register (MSIENA) within the SYS register frame. Each MBIST controller is enabled by setting the corresponding bit within this register and then enabling memory self-test via the memory self-test global enable within the global control register (MSTGCR.MSTGENA[3:0]).

The MBIST controllers support execution of the following tests:

Table 3-11. MBIST Algorithms and Cycle Counts<sup>(1)</sup>

Module	Algorithm (Cycle Counts)						
	Checker Board	March13N Background 0	March11N Background A	March13N Background 3/0F/69	PMOS Open Address Decode	ROM2	
ADC RAM	1427	1555	1089	4033	4225	-	
DCAN RAM	1503	1503	1057	3745	3265	-	
SRAM	26835	26835	22529	79873	147457	-	
HET RAM	7539	8307	6529	24193	29185	-	
MibSPI RAM	3583	3583	2817	9985	10753	-	
STC ROM	-	-	-	-	-	18433	

<sup>(1)</sup> Cycle times provided are for the execution of the specific algorithms and do not include overhead from the BIST statemachine.

#### **NOTE**

The algorithm to be applied is selectable via the memory self-test global control register algo selection field (MSTGCR.MBIST\_ALGSEL[7:0]).

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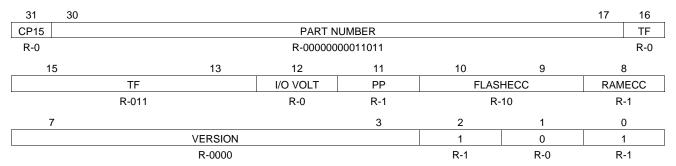
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### 3.11 Device Identification Code Register

The device identification code register identifies the coprocessor status, an assigned device-specific part number, the technology family (TF), the I/O voltage, whether or not parity is supported, the levels of flash and RAM error detection, and the device version. The TMS470M device identification code base register value is 0X00366D05 and is subject to change based on the silicon version.



LEGEND: R = Read only; -n = value after reset

Figure 3-4. TMS470 Device ID Bit Allocation Register

#### Table 3-12. TMS470 Device ID Bit Allocation Register Field Descriptions

Bit	Field	Value	Description
31	CP15		This bit indicates the presence of coprocessor (CP15).
		0	No coprocessor present in the device.
		1	Coprocessor present in the device.
30-17	PART NUMBER		These bits indicate the assigned device-specific part number. The assigned device-specific part number for the TMS470M device is 0000000011011.
16-13	TF		Technology family bit. These bits indicate the technology family (C05, F05, F035, C035).
		0011	F035
12	I/O VOLT		I/O voltage bit. This bit identifies the I/O power supply.
		0	3.3 V
		1	5 V
11	PP		Peripheral parity bit. This bit indicates whether parity is supported.
		0	No parity on peripheral.
		1	Parity on peripheral.
10	FLASHECC		Flash ECC bits. These bits indicate the level of error detection and correction on the flash memory.
		00	No error detection/correction.
		01	Program memory with parity.
		10	Program memory with ECC.
		11	Reserved
8	RAMECC		RAM ECC bits. This bit indicates the presence of error detection and correction on the CPU RAM.
		0	RAM ECC not present.
		1	RAM ECC present.
7-3	VERSION		These bits identify the silicon version of the device.
2-0	101		Bits 2:0 are set to 101 by default to indicate a platform device.

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#### 3.12 Device Part Numbers

Table 3-13 lists all the available TMS470MF04207/TMS470MF03107 device configurations.

#### **Table 3-13. Device Part Numbers**

DEVICE PART NUMBER	SAP PART NUMBER	PROGRAM MEMORY	PACKAGE TYPE	TEMPERATURE RANGE	PbFREE/
	SAF FART NUMBER	FLASH EEPROM	100-PIN LQFP	-40°C to 125°C	GREEN <sup>(1)</sup>
TMS470MF04207PZQ	S4MF04207SPZQQ1	Х	X	Х	Х
TMS470MF03107PZQ	S4MF03107SPZQQ1	Х	X	X	Х

<sup>(1)</sup> RoHS compliant products are compatible with the current RoHS requirements for all six substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials, unless exempt. Pb-Free products are RoHS Compliant, plus suitable for use in higher temperature lead-free solder processes (typically 245 to 260°C). Green products are RoHS and Pb-Free, plus also free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material).

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### 4 Device Operating Conditions

### 4.1 Absolute Maximum Ratings Over Operating Free-Air Temperature Range, Q Version<sup>(1)</sup>

Supply voltage range:	V <sub>CC</sub> <sup>(2)</sup>	-0.5 V to 2.1 V
	V <sub>CCIOR</sub> , V <sub>CCAD</sub> , V <sub>CC</sub> (Flash pump) <sup>(2)</sup>	-0.5 V to 4.1 V
Input voltage range:	All input pins	-0.5 V to 4.1 V
Input clamp current:	I <sub>IK</sub> (V <sub>I</sub> < 0 or V <sub>I</sub> > V <sub>CCIOR</sub> ) All pins, except ADIN[0:15]	±20 mA
	$I_{IK}$ (V <sub>I</sub> < 0 or V <sub>I</sub> > V <sub>CCIOR</sub> ) ADIN[0:15]	±10 mA
Operating free-air temperature range, T <sub>A</sub> :	Q version	-40°C to 125°C
Operating junction temperature range, T <sub>J</sub> :	Standard	-40°C to 150°C
Storage temperature range, T <sub>stg</sub>		-65°C to 150°C

<sup>(1)</sup> Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only. Functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

### 4.2 Device Recommended Operating Conditions<sup>(1)</sup>

			MIN	NOM	MAX	UNIT
V <sub>CCIOR</sub>	Digital I/O and internal regulator supply	Digital I/O and internal regulator supply voltage		3.3	3.6	V
V <sub>CC</sub>	Voltage regulator output voltage	Voltage regulator output voltage		1.55	1.70	V
$V_{CCAD}$	MibADC supply voltage	MibADC supply voltage		3.3	3.6	V
V <sub>CCP</sub>	Flash pump supply voltage	Flash pump supply voltage		3.3	3.6	V
V <sub>SS</sub>	Digital logic supply ground	Digital logic supply ground		0		V
V <sub>SSAD</sub>	MibADC supply ground		-0.1		0.1	V
T <sub>A</sub>	Operating free-air temperature	Q version	-40		125	°C
TJ	Operating junction temperature	Operating junction temperature			150	°C

<sup>(1)</sup> All voltages are with respect to  $V_{SS}$ , except  $V_{CCAD}$ , which is with respect to  $V_{SSAD}$ .

<sup>(2)</sup> All voltage values are with respect to their associated grounds.

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#### 4.3 **Electrical Characteristics Over Recommended Operating Free-Air Temperature** Range, Q Version<sup>(1)(2)</sup>

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT		
V <sub>hys</sub>	Input hysteresis			150			mV	
V <sub>IL</sub>	Laurelian et caleana	All inputs <sup>(3)</sup>		-0.3		0.8	V	
	Low-level input voltage	OSCIN				0.2 V <sub>CC</sub>	V	
14	IP-1 1 IP C IC	All inputs <sup>(3)</sup>		2		V <sub>CCIOR</sub> + 0.3	V	
$V_{IH}$	High-level input voltage	OSCIN		0.8 V <sub>CC</sub>				
			$I_{OL} = I_{OL} MAX$			0.2 V <sub>CCIOR</sub>		
V <sub>OL</sub>	V <sub>OI</sub> Low-level output voltage		I <sub>OL</sub> = 50 μA Standard mode			0.2	V	
vol Low-level output voltage			I <sub>OL</sub> = 50 μA Impedance Control mode			0.2 V <sub>CCIOR</sub>	, <b>v</b>	
V <sub>OH</sub> High-level output voltage			$I_{OH} = I_{OH} MAX$	0.8 V <sub>CCIOR</sub>				
		I <sub>OH</sub> = 50 μA Standard mode	V <sub>CCIOR</sub> -0.2			V		
VOH	V <sub>OH</sub> High-level output voltage		I <sub>OH</sub> = 50 μA Impedance Control mode	0.8 V <sub>CCIOR</sub>				
I <sub>IC</sub>	Input clamp current (I/O pins) <sup>(4)</sup>		$V_I < V_{SSIO}$ - 0.3 or $V_I$ > $V_{CCIOR}$ + 0.3	-2		2	mA	
		I <sub>IH</sub> Pulldown	$V_I = V_{CCIOR}$	40		190	μΑ	
I <sub>I</sub>	Input current (I/O pins)	I <sub>IL</sub> Pullup	$V_I = V_{SS}$	-190		-40		
"	par oanom (# o pino)	All other pins	No pullup or pulldown	-1		1		
I <sub>OL</sub>	Low-level output tcurrent	Adaptive impedance 4 mA Buffer	V <sub>OL</sub> = V <sub>OL</sub> MAX			4	mA	
I <sub>OH</sub>	High-level output current	Adaptive impedance 4 mA Buffer	V <sub>OH</sub> = V <sub>OH</sub> MIN	-4			mA	
I <sub>CC</sub>	V <sub>CC</sub> digital supply current (operating mode, internal regulator disabled)		HCLK = 80 MHz, VCLK = 80 MHz, V <sub>CC</sub> = 1.70 V <sup>(5)</sup>			110	mA	
	V <sub>CCIOR</sub> IO and digital supply current (operating mode, internal regulator enabled)		HCLK = 80 MHz, VCLK = 80 MHz, No DC load, V <sub>CCIOR</sub> = 3.6 V <sup>(5)(6)</sup>			115		
I <sub>CCIOR</sub>	V <sub>CCIOR</sub> IO and digital supply execution, internal regulator	current (LBIST enabled) <sup>(7)</sup>	$\label{eq:hclk} \begin{array}{l} \text{HCLK} = 80 \text{ MHz},\\ \text{VCLK} = 80 \text{ MHz},\\ \text{STCCLK} = 80 \text{ MHz},\\ \text{No DC load},\\ \text{V}_{\text{CCIOR}} = 3.6 \text{ V}^{(6)} \end{array}$			155	mA	
V <sub>CCIOR</sub> IO and digital supply execution, internal regulator of		current (MBIST enabled) <sup>(8)</sup>	$\label{eq:CLK} \begin{aligned} & \text{HCLK} = 80 \text{ MHz}, \\ & \text{VCLK} = 80 \text{ MHz}, \\ & \text{No DC load, V}_{\text{CCIOR}} \\ & = 3.6 \text{ V}^{(6)} \end{aligned}$			130		

- Source currents (out of the device) are negative while sink currents (into the device) are positive.
- "All frequencies" will include all specified device configuration frequencies.

  The V<sub>IL</sub> here does <u>not apply</u> to the OSCIN, OSCOUT and PORRST pins; the V<sub>IH</sub> here does not apply to the OSCIN, OSCOUT and RST pins; For RST and PORRST exceptions, see Section 5.1.
- Parameter does not apply to input-only or output-only pins.
- Maximum currents are measured using a system-level test case. This test case exercises all of the device peripherals concurrently (excluding MBIST and STC LBIST).
- I/O pins configured as inputs or outputs with no load. All pulldown inputs ≤ 0.2 V. All pullup inputs ≥ V<sub>CCIO</sub> 0.2 V. ECLK output ≤ (6)2 MHz.
- LBIST current specified is peak current for the maximum supported operating clock (HCLK = 80 MHz) and STC CLK = HCLK. Lower current consumption can be achieved by configuring a slower STC Clock frequency. The current peak duration can last for the duration of 1 LBIST test interval.
- MBIST currents specified are for execution of MBIST on all RAMs in parallel. Lower current consumption can be achieved by sequenced execution of MBIST on each of the RAM spaces available.

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#### Electrical Characteristics Over Recommended Operating Free-Air Temperature Range, Q Version<sup>(1)(2)</sup> (continued)

	PARAMETER		MIN	TYP	MAX	UNIT
I <sub>CCAD</sub>	V <sub>CCAD</sub> supply current (operating mode)	All frequencies, V <sub>CCAD</sub> = 3.6 V			8	mA
		V <sub>CCP</sub> = 3.6 V read operation <sup>(5)</sup>			10	
I <sub>CCP</sub> V <sub>CCP</sub> pump supply current	V <sub>CCP</sub> pump supply current	V <sub>CCP</sub> = 3.6 V program <sup>(9)</sup>			75	mA
		V <sub>CCP</sub> = 3.6 V erase			75	
I <sub>CCTOTAL</sub> (10)	$V_{\rm CCIOR}$ + $V_{\rm CCAD}$ + $V_{\rm CCP}$ total digital supply current (operating mode, internal regulator enabled)	$\label{eq:hatconstraints} \begin{split} & \text{HCLK} = 80 \text{ MHz}, \\ & \text{VCLK} = 80 \text{ MHz}, \\ & \text{No DC load}, \\ & \text{V}_{\text{CCIOR}} = 3.6 \text{ V}^{(5)(6)} \end{split}$			125	mA
C <sub>I</sub>	Input capacitance		6			pF
Co	Output capacitance		7			pF

<sup>(9)</sup> Assumes reading from one bank while programming a different bank.

# **Peripheral Information and Electrical Specifications**

#### RST and PORRST Timings 5.1

Table 5-1. Timing Requirements for PORRST<sup>(1)</sup>

(see Figure 5-1)

NO.			MIN	MAX	UNIT
	V <sub>CCPORL</sub>	V <sub>CC</sub> low supply level when $\overline{\text{RST}}$ becomes active	1.30		V
	V <sub>CCPORH</sub>	V <sub>CC</sub> high supply level when RST becomes active		1.80	V
	V <sub>CCIOPORL</sub>	V <sub>CCIO</sub> low supply level when PORRST must be active during power up		1.1	V
	V <sub>CCIOPORH</sub>	V <sub>CCIO</sub> high supply level when PORRST must remain active during power up and become active during power down	3.0		V
	V <sub>IL</sub> <sup>(2)</sup>	Low-level input voltage after V <sub>CCIOR</sub> > V <sub>CCIOPORH</sub>		0.2 V <sub>CCIOR</sub>	V
	V <sub>OH</sub> <sup>(3)</sup>	High-level output voltage after V <sub>CCIOR</sub> > V <sub>CCIOPORH</sub>	0.8 V <sub>CCIOR</sub>		V
	V <sub>IL(PORRST)</sub>	Low-level input voltage of PORRST before V <sub>CCIOP</sub> > V <sub>CCIOPORL</sub>		0.5	V
3	t <sub>su(PORRST)r</sub>	Setup time, PORRST active before V <sub>CCIOR</sub> > V <sub>CCIOPORL</sub> during power up	0		ms
5	t <sub>su(VCCIOR)r</sub>	Setup time, $V_{CCIOP} > V_{CCIOPORL}$ before $V_{CC} > V_{CCPORL}$	0		ms
6	t <sub>h(PORRST)r</sub>	Hold time, $\overline{PORRST}$ active after $V_{CC} > V_{CCPORH}$	1		ms
7	t <sub>su(PORRST)f</sub>	Setup time, $\overline{\text{PORRST}}$ active before $V_{CC} \le V_{CCPORH}$ during power down	8		μs
8	t <sub>h(PORRST)rio</sub>	Hold time, PORRST active after V <sub>CCIOP</sub> > V <sub>CCIOPORH</sub>	1		ms
9	t <sub>h(PORRST)d</sub>	Hold time, PORRST active after V <sub>CCIOR</sub> < V <sub>CCIORPORL</sub>	0		ms
10	t <sub>su(PORRST)fio</sub>	Setup time, PORRST active before V <sub>CC</sub> ≤ V <sub>CCIOPORH</sub> during power down	0		ns
11	t <sub>su(VCCIO)f</sub>	Setup time, V <sub>CC</sub> < <sub>VCCPORE</sub> before V <sub>CCIO</sub> < V <sub>CCIOPORL</sub>	0		ns
	$t_{f(PORRST)}$	Filter time PORRST, pulses less than MIN get filtered out; pulses greater than MAX generate a reset.	30	185	ns
	t <sub>f(RST)</sub>	Filter time RST, pulses less than MIN get filtered out; pulses greater than MAX generate a reset.	40	150	ns

When the  $V_{CC}$  timing requirements for  $\overline{PORRST}$  are satisfied, there are no timing requirements for  $V_{CCP}$ . Corresponds to  $\overline{PORRST}$ . Corresponds to  $\overline{RST}$ .

<sup>(10)</sup> Total device operating current is derived from the total |<sub>CCIOR</sub>, |<sub>CCAD</sub>, and |<sub>CCP</sub> in normal operating mode excluding MBIST and LBIST execution. It is expected that the total will be less than the sums of the values of the individual components due to statistical calculations involved in producing the specification values.

<sup>(2)</sup> 

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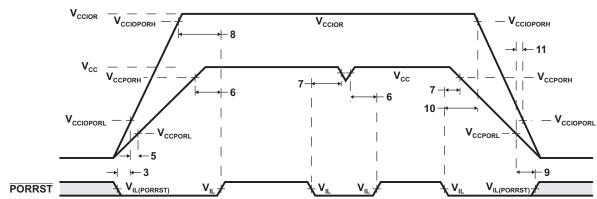
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V<sub>cc</sub> (1.55 V)  $V_{CCP}/V_{CCIOR}$  (3.3 V)

Note: V<sub>CC</sub> is provided by the on-chip voltage regulator during normal application run time. It is not recommended to use the device in an application with the Vreg disabled due to potential glitching issues; however, if used in this mode, the application should ensure that the specified voltage ranges for V<sub>CC</sub> are maintained.

Figure 5-1. PORRST Timing Diagram

Table 5-2. Switching Characteristics Over Recommended Operating Conditions for RST and PORRST<sup>(1)</sup>

	PARAMETER			UNIT
	Valid time, RST active after PORRST inactive	1024t <sub>c(OSC)</sub>		
t <sub>v(RST)</sub>	Valid time, RST active (all others)	8t <sub>c(VCLK)</sub>		ns
V <sub>CCIOPORL</sub>	Vccio low supply level when PORRST must be active during power-up and power-down		1.1	V

(1) Specified values do NOT include rise/fall times. For rise and fall timings, see Table 5-13.

#### Table 5-3. Internal Voltage Regulator Specifications

· · · · · · · · · · · · · · · · · · ·						
	PARAMETER					
t <sub>D(VCCIOR)0-3</sub>	Delay time, input supply to ramp from 0 V to 3 V	12	1000	μs		
t <sub>V(PORRST)L</sub>	Valid time, PORRST active after input supply becomes ≥ 3.0 V	1		ms		
$V_{CCIORmin(PORRST)f}$	Minimum input voltage, when PORRST must be made active during power down or brown out	3.0		V		
C <sub>(VCC)core</sub>	Capacitance distributed over core V <sub>CC</sub> pins for voltage regulator stability	1.2	6.0	μF		
ESR <sub>(max)core</sub>	Total combined ESR of stabilization capacitors on core V <sub>CC</sub> pins	0	0.75	Ω		

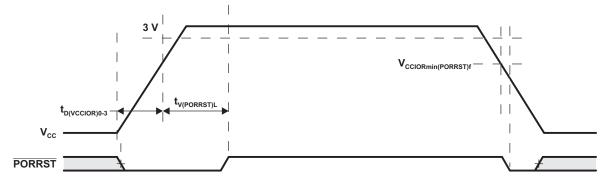


Figure 5-2. PORRST Timing Requirements



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### Table 5-4. VREG Recommended Operation Conditions

	PARAMETER	CONDITIONS	MIN	MAX	UNIT
I <sub>CC</sub> V <sub>CC</sub> Load Rating	Normal mode, regulator active	0	200	mA	
		Off, enable forced off	-	-	μA

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#### 5.2 **PLL and Clock Specifications**

Table 5-5. Timing Requirements for PLL Circuits Enabled or Disabled

PARAMETER		MIN	MAX	UNIT
f <sub>(OSC)</sub>	Input clock frequency	5	20	MHz
t <sub>c(OSC)</sub>	Cycle time, OSCIN	50		ns
t <sub>w(OSCIL)</sub>	Pulse duration, OSCIN low	15		ns
t <sub>w(OSCIH)</sub>	Pulse duration, OSCIN high	15		ns

#### 5.2.1 External Reference Resonator/Crystal Oscillator Clock Option

The oscillator is enabled by connecting the appropriate fundamental 5-20 MHz resonator/crystal and load capacitors across the external OSCIN and OSCOUT pins as shown in Figure 5-3(a). The oscillator is a single stage inverter held in bias by an integrated bias resistor.

TI strongly encourages each customer to submit samples of the device to the resonator/crystal vendors for validation. Vendors are equipped to determine which load capacitors will best tune their resonator/crystal to the microcontroller device for optimum start-up and operation over temperature/voltage extremes.

#### 5.2.2 External Clock Source

An external oscillator source can be used by connecting a 1.55-V clock signal to the OSCIN pin and leaving the OSCOUT pin unconnected (open) as shown in Figure 5-3(b).



The values of C1 and C2 should be provided by the resonator/crystal vendor.

Figure 5-3. Recommended Crystal/Clock Connection



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#### 5.2.3 Validated FMPLL Settings

The following table includes the validated FMPLL settings.

#### Table 5-6. Validated FMPLL Settings

Mode	OSCIN Frequency (MHz)	PLLCTL1 <sup>(1)</sup>	PLLCTL2 <sup>(1)</sup>	FMPLL Output Frequency (MHz)	Modulation Bandwidth (KHz)	Modulation Depth
	10	0x20048B00	0x00007800			
	12	0x20026100	0x00007C00			
	12	0x20058B00	0x00007800			
Non-Modulated (2)	14	0x20055F00	0x00007600	50		
Non-wodulated V	14	0x20068B00	0x00007800	56	-	-
	16	0x20036100	0x00007C00			
	16	0x20078B00	0x00007800			
	20	0x20098B00	0x00007800			
	10	0x20049F00	0x00007800	64		
	12	0x20055F00	0x00007400			
	12	0x20059F00	0x00007800			
Non-Modulated (2)	4.4	0x20065F00	0x00007400			
Non-Modulated (=/	14	0x20069F00	0x00007800		-	-
	16	0x20075F00	0x00007400			
	16	0x20079F00	0x00007800			
	20	0x20099F00	0x00007800			
	10	0x20049F00	0x00007600			
	40	0x20026300	0x00007800			
	12	0x20059F00	0x00007600			
Na - Madulatad(2)	4.4	0x20067700	0x00007400	00		
Non-Modulated (2)	14	0x20069F00	0x00007600	80	-	-
	16	0x20036300	0x00007800			
	16	0x20079F00	0x00007600			
	20	0x20099F00	0x00007600			

<sup>(1)</sup> The recommended PLLCTL1 and PLLCTL2 values make no assumption of the intended use of ROS, BPOS, and ROF fields within the PLL control registers. For these settings, the application should set these as appropriate for the specific application requirements.

<sup>(2)</sup> Non-Modulated settings provided show FM related bit fields as 0. When initializing the PLLCTL registers for non-modulated use, the FM related bit fields should be masked such that reset/default values are retained.

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#### Table 5-6. Validated FMPLL Settings (continued)

Mode	OSCIN Frequency (MHz)	PLLCTL1 <sup>(1)</sup>	PLLCTL2 <sup>(1)</sup>	FMPLL Output Frequency (MHz)	Modulation Bandwidth (KHz)	Modulation Depth
		0x20048B00	0x8300B844		76.92	0.50%
	10	0x20048B00	0x8300B889		76.92	1.00%
	10	0x20048B00	0x82409859		100	0.50%
		0x20048B00	0x824098B2		100	1.00%
		0x20058B00	0x8300B844		76.02	0.50%
	12	0x20058B00	0x8300B889		76.92	1.00%
	12	0x20058B00	0x82409859		100	0.50%
		0x20058B00	0x824098B2		100	1.00%
		0x20068B00	0x8300B844		76.92	0.50%
Frequency	1.4	0x20068B00	0x8300B889	56	76.92	1.00%
Modulated	14	0x20068B00	0x82409859	56	100	0.50%
		0x20068B00	0x824098B2		100	1.00%
		0x20078B00	0x8300B844		76.92	0.50%
	16	0x20078B00	0x8300B889		76.92	1.00%
	16	0x20078B00	0x82409859		100	0.50%
		0x20078B00	0x824098B2		100	1.00%
		0x20098B00	0x8300B844		76.92	0.50%
	00	0x20098B00	0x8300B889			1.00%
	20	0x20098B00	0x82409859			0.50%
		0x20098B00	0x824098B2		100	1.00%
		0x20049F00	0x8300C83B		76.92 - 100 - 76.92 -	0.50%
	10	0x20049F00	0x8300C878			1.00%
	10	0x20049F00	0x8240A84D			0.50%
		0x20049F00	0x8240A89C			1.00%
		0x20059F00	0x8300C83B			0.50%
	40	0x20059F00	0x8300C878			1.00%
	12	0x20059F00	0x8240A84D		400	0.50%
		0x20059F00	0x8240A89C		100	1.00%
		0x20069F00	0x8300C83B		76.00	0.50%
Frequency	4.4	0x20069F00	0x8300C878	64	76.92	1.00%
Modulated	14	0x20069F00	0x8240A84D	64	100	0.50%
		0x20069F00	0x8240A89C		100	1.00%
		0x20079F00	0x8300C83B		76.92	0.50%
	16	0x20079F00	0x8300C878		70.92	1.00%
	16	0x20079F00	0x8240A84D		100	0.50%
		0x20079F00	0x8240A89C		100	1.00%
		0x20099F00	0x8300C83B		76.00	0.50%
	20	0x20099F00	0x8300C878		76.92	1.00%
	20	0x20099F00	0x8240A84D		100	0.50%
		0x20099F00	0x8240A89C		100	1.00%



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#### Table 5-6. Validated FMPLL Settings (continued)

Mode	OSCIN Frequency (MHz)	PLLCTL1 <sup>(1)</sup>	PLLCTL2 <sup>(1)</sup>	FMPLL Output Frequency (MHz)	Modulation Bandwidth (KHz)	Modulation Depth
		0x20049F00	0x8300C63B		76.00	0.50%
	10	0x20049F00	0x8300C678		76.92	1.00%
	10	0x20049F00	0x8240A64D		100	0.50%
		0x20049F00	0x8240A69C		100	1.00%
		0x20059F00	0x8300C63B		76.92	0.50%
	12	0x20059F00	0x8300C678		76.92	1.00%
	12	0x20059F00	0x8240A64D		100	0.50%
		0x20059F00	0x8240A69C	100	100	1.00%
		0x20069F00	0x8300C63B	80	76.92	0.50%
Frequency	14	0x20069F00	0x8300C678		76.92	1.00%
Modulated	14	0x20069F00	0x8240A64D		100	0.50%
		0x20069F00	0x8240A69C		100	1.00%
		0x20079F00	0x8300C63B		76.92	0.50%
	16	0x20079F00	0x8300C678		76.92	1.00%
	10	0x20079F00	0x8240A64D		100	0.50%
		0x20079F00	0x8240A69C	I I I	100	1.00%
		0x20099F00	0x8300C63B		76.92	0.50%
	20	0x20099F00	0x8300C678		70.92	1.00%
	20	0x20099F00	0x8240A64D		100	0.50%
		0x20099F00	0x8240A69C		100	1.00%

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#### 5.2.4 LPO and Clock Detection

The LPOCLKDET module consists of a clock monitor (CLKDET) and 2 low-power oscillators (LPO): a low-frequency (LF) and a high-frequency (HF) oscillator. The CLKDET is a supervisor circuit for an externally supplied clock signal. In case the externally supplied clock frequency falls out of a frequency window, the clock detector flags this condition and switches to the HF LPO clock (limp mode). The OSCFAIL flag and clock switch-over remain, regardless of the behavior of the oscillator clock signal. The only way OSCFAIL can be cleared (and OSCIN be again the driving clock) is a power-on reset.

Table 5-7. LPO and Clock Detection

PARAMETER		MIN	TYP	MAX	UNIT
P. I. C	Lower threshold	1.5		5.0	MHz
invalid frequency	Higher threshold	20.0		50.0	MHz
limp mode frequency (HFosc)		7.6	12	14.0	MHz
LFosc frequency		50	90	124	kHz
HFosc frequency		7.6	12	14.0	MHz

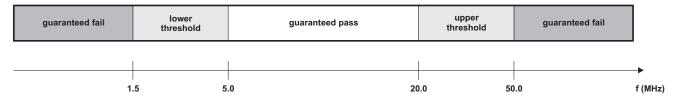


Figure 5-4. LPO and Clock Detection

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#### 5.2.5 Device Clock Domains Block Diagram

The clock domains block diagram and GCM clock source assignments are given in Figure 5-5 and Table 5-8.

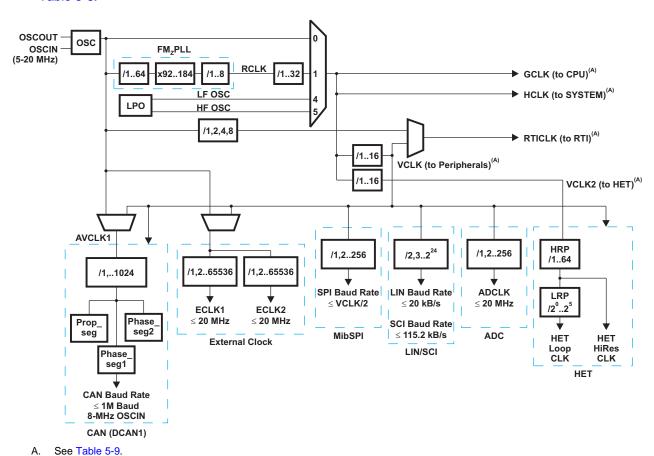


Figure 5-5. Device Clock Domains Block Diagram

**Table 5-8. GCM Clock Source Assignments** 

GCM SOURCE NUMBER	CLOCK SOURCE
0	OSCIN
1	F035 FMzPLL
2	Reserved
3	Reserved
4	LF OSC
5	HF OSC
6	Reserved
7	Reserved

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### Table 5-9. Switching Characteristics Over Recommended Operating Conditions for Clocks<sup>(1)(2)(3)(4)(5)</sup>

	PARAMETER	TEST CONDITIONS (6)	MIN	MAX	UNIT
		Pipeline mode enabled		80	
f <sub>(HCLK)</sub>	System clock frequency	Pipeline mode disabled, 0 flash wait states		28	MHz
f <sub>(PROG/ERASE)</sub>	System clock frequency Flash programming/erase			80	MHz
f <sub>(VCLK/VCLK2)</sub>	Peripheral VBUS clock frequency			f <sub>(HCLK)</sub>	MHz
$f_{(ECLK)}$	External clock output frequency for ECP Module			20	MHz
f <sub>(RCLK)</sub>	RCLK - Frequency out of PLL macro into R-divider (Post ODPLL divider)			145	MHz
		Pipeline mode enabled	12.50		
$t_{c(\text{HCLK})}$	Cycle time, system clock	Pipeline mode disabled, 0 flash wait states	35.71		ns
t <sub>c(PROG/ERASE)</sub>	Cycle time, system clock - Flash programming/erase		12.50		ns
t <sub>c(VCLK/VCLK2)</sub>	Cycle time, peripheral clock		t <sub>c(HCLK)</sub>		ns
t <sub>c(ECLK)</sub>	Cycle time, ECP module external clock output		50.0		ns
t <sub>c(RCLK)</sub>	Cycle time, RCLK minimum input cycle time out of PLL macro into R-divider		6.90		ns

- (1) f<sub>(HCLK)</sub> = f<sub>(OSC)</sub> / NR \*NF /ODPLL/PLLDIV; for details, see the PLL documentation. TI strongly recommends selection of NR and NF parameters such that NF ≤ 120 and (f<sub>(OSC)</sub> / NR \*NF) ≤ 400. f<sub>(VCLK)</sub> = f<sub>(HCLK)</sub> / X, where X = {1,2,3,4,5,6,7,8,9,10,11,12,13,14,15,16}. X is the peripheral VBUS clock divider ratio determined by the VCLKR[3:0] bits in the SYS module.
- (2) Enabling FM mode can reduce maximum rated operating frequencies. The degree of impact is application-specific and the specific settings, as well as the impact of the settings, should be discussed and agreed upon prior to using FM modes. Use of FM modes do not impact the maximum rated external clock output, f<sub>(ECLK)</sub>, for the ECP module.
- (3) Pipeline mode enabled or disabled is determined by FRDCNTL[2:0].
- (4)  $f_{(ECLK)} = f_{(VCLK)} / N$ , where N = {1 to 65536}. N is the ECP prescale value defined by the ECPCTRL.[15:0] register bits in the ECP module
- (5) ECLK output will increase radiated emissions within the system that is used. Rated emissions at the device level do not include emissions due to ECLK output.
- (6) All test conditions assume FM Mode disabled and RAM ECC enabled with 0 waitstates for RAM.

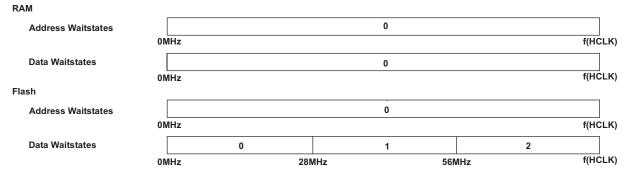


Figure 5-6. Timing - Wait States

#### **NOTE**

If FMzPLL frequency modulation is enabled, special care must be taken to ensure that the maximum system clock frequency f(HCLK) and peripheral clock frequency f(VCLK) are not exceeded. The speed of the device clocks may need be derated to accommodate the modulation depth when FMzPLL frequency modulation is enabled.

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#### 5.2.5.1 ECLK Specification

## Table 5-10. Switching Characteristics Over Recommended Operating Conditions for External Clocks<sup>(1)(2)</sup>

(see Figure 5-7)

NO.	PARAMETER		TEST CONDITIONS	MIN	MAX	UNIT
1	$t_{w(EOL)}$	Pulse duration, ECLK low	Under all prescale factor combinations (X and N)	0.5 <sub>tc(ECLK)</sub> - t <sub>f</sub>		ns
2	t <sub>w(EOH)</sub>	Pulse duration, ECLK high	Under all prescale factor combinations (X and N)	0.5t <sub>c(ECLK)</sub> - t <sub>r</sub>		ns

- X = {1,2,3,4,5,6,7,8,9,10,11,12,13,14,15,16}. X is the VBUS interface clock divider ratio determined by the CLKCNTL.[19:16] bits in the (1) SYS module.
- N = {1 to 65536}. N is the ECP prescale value defined by the ECPCNTL.[15:0] register bits in the SYS module.

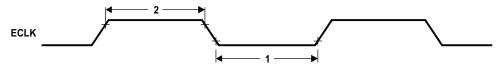


Figure 5-7. ECLK Timing Diagram

#### 5.2.6 TEST Pin Glitch Filter Timing

#### **Table 5-11. Test Pin Glitch Filter Timing**

NO.	PARAMETER		MIN	MAX	UNIT
	t <sub>f(TEST)</sub>	Filter time TEST, high pulses less than MIN will be filtered out.	40		ns



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# 5.2.7 JTAG Timing

# Table 5-12. JTAG Scan Interface Timing (JTAG Clock specification 10-MHz and 50-pF Load on TDO Output)

#### (see Figure 5-8

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NO.			MIN	MAX	UNIT
1	t <sub>c(JTAG)</sub>	Cycle time, JTAG low and high period	50		ns
2	t <sub>su(TDI/TMS</sub> - TCKr)	Setup time, TDI, TMS before TCK rise (TCKr)	5		ns
3	t <sub>h(TCKr</sub> -TDI/TMS)	Hold time, TDI, TMS after TCKr	5		ns
4	t <sub>h(TCKf</sub> -TDO)	Hold time, TDO after TCKf	5		ns
5	t <sub>d(TCKf</sub> -TDO)	Delay time, TDO valid after TCK fall (TCKf)		45	ns

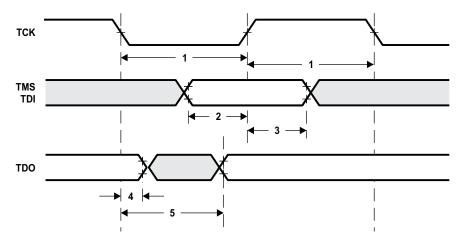


Figure 5-8. JTAG Scan Timings

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#### 5.2.8 Output Timings

#### Table 5-13. Switching Characteristics for Output Timings Versus Load Capacitance (C<sub>L</sub>)<sup>(1)</sup>

(see Figure 5-9)

	PARAMETER	MAX	UNIT	
	Adaptive impedance 4 mA pins	C <sub>L</sub> = 15 pF	4	
		C <sub>L</sub> = 50 pF	8	
t <sub>r</sub>		C <sub>L</sub> = 100 pF	15	ns
		C <sub>L</sub> = 150 pF	21	
	Adaptive impedance 4 mA pins	C <sub>L</sub> = 15 pF	5	
		C <sub>L</sub> = 50 pF	8	
t <sub>f</sub>		C <sub>L</sub> = 100 pF	12	ns
		C <sub>L</sub> = 150 pF	17	

(1) Peripheral output timings given within this document are measured in either standard buffer or impedance control mode.

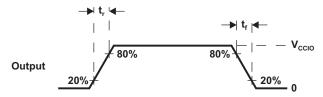


Figure 5-9. CMOS-Level Outputs

#### 5.2.9 Input Timings

Table 5-14. Timing Requirements for Input Timings<sup>(1)</sup>

(see Figure 5-10)

		MIN	MAX	UNIT
t <sub>pw</sub>	Input minimum pulse width	t <sub>c(VCLK)</sub> + 10		ns

(1)  $t_{c(VCLK)}$  = peripheral VBUS clock cycle time = 1 /  $f_{(VCLK)}$ .

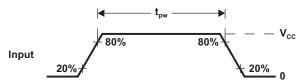


Figure 5-10. CMOS-Level Inputs

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### 5.2.10 Flash Timings

Table 5-15. Timing Requirements for Program Flash

	PARAMETER	TEST CONDITIONS	MIN	NOM	MAX	UNIT
	Flash pump stabilization time  Flash bank stabilization time	From Sleep Mode to Standby Mode	20			
		From Standby Mode to Active Mode	1			
t <sub>acc_delay</sub>		From Sleep Mode to Standby Mode	1.9			μs
		From Standby Mode to Active Mode	0.1	0.1		
t <sub>prog(32-bit)</sub>	Half-word (32-bit) programming time			37.5	300	μs
	384k-byte programming time <sup>(1)</sup>			3.7	29.5	_
<sup>t</sup> prog(Total)	448k-byte programming time <sup>(1)</sup>			4.3	34.4	S
t <sub>erase(sector)</sub>	Sector erase time			1.5	15	s
oraco(cocioi)	Write/erase cycles at TA = -40 to 125°C with 15-year Data Retention requirement				1000 <sup>(2)</sup>	cycles
N <sub>wec</sub>	Write/erase cycles at TA = -40 to 125°C EEPROM emulation requirement for 16k flash sectors in Bank 1				25000 <sup>(2)(3)</sup>	cycles

t<sub>prog(Total)</sub> programming time includes overhead of state machine, but does not include data transfer time. Flash write/erase cycles and data retention specifications are based on a validated implementation of the TI flash API. Non-TI flash API implementation is not supported. For detailed description see the F035 Flash Validation Procedure (SPNA127).

Flash write/erase cycle and data retention specifications are based on an assumed distribution of write/erase cycles over the life of the product including an even distribution over the rated temperature range and time between cycles. The EEPROM emulation bank has been qualified as outlined in the JEDEC specification JESD22-A117C.



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#### 5.3 **SPIn Master Mode Timing Parameters**

# Table 5-16. SPIn Master Mode External Timing Parameters (CLOCK PHASE = 0, SPInCLK = output, SPInSIMO = output, and SPInSOMI = input) $^{(1)(2)(3)}$

(see Figure 5-11 and Figure 5-12)

NO.			MIN	MAX	UNIT
1	t <sub>c(SPC)M</sub>	Cycle time, SPICLK <sup>(4)</sup>	90	256t <sub>c(VCLK)</sub>	
2 <sup>(5)</sup>	t <sub>w(SPCH)M</sub>	Pulse duration, SPICLK high (clock polarity = 0)	0.5t <sub>c(SPC)M</sub> - t <sub>r</sub> -8	$0.5t_{c(SPC)M} + 5$	
	t <sub>w(SPCL)M</sub>	Pulse duration, SPICLK low (clock polarity = 1)	0.5t <sub>c(SPC)M</sub> - t <sub>f</sub> -8	0.5t <sub>c(SPC)M</sub> + 5	
3 <sup>(5)</sup>	$t_{w(SPCL)M}$	Pulse duration, SPICLK low (clock polarity = 0)	0.5t <sub>c(SPC)M</sub> - t <sub>f</sub> -8	$0.5t_{c(SPC)M} + 5$	
3 <sup>(5)</sup>	t <sub>w(SPCH)M</sub>	Pulse duration, SPICLK high (clock polarity = 1)	$0.5t_{c(SPC)M} - t_r-8$	$0.5t_{c(SPC)M} + 5$	
4 <sup>(5)</sup>	t <sub>d</sub> (SIMO-SPCL)M	Delay time, SPISIMO data valid before SPICLK low (clock polarity = 0)	0.5t <sub>c(SPC)M</sub> - 10		
4(3)	t <sub>d(SIMO-SPCH)M</sub>	Delay time, SPISIMO data valid before SPICLK high (clock polarity = 1)	0.5t <sub>c(SPC)M</sub> - 10		
5 <sup>(5)</sup>	t <sub>v(SPCL-SIMO)M</sub>	Valid time, SPISIMO data valid (clock polarity = 0)	0.5t <sub>c(SPC)M</sub> - t <sub>f(SPC)</sub> -5		
	t <sub>v(SPCH-SIMO)M</sub>	Valid time, SPISIMO data valid (clock polarity = 1)	0.5t <sub>c(SPC)M</sub> - t <sub>r(SPC)</sub> -5		
	t <sub>su(SOMI-SPCL)M</sub>	Setup time, SPISOMI before SPICLK low (clock polarity = 0)	t <sub>r(SPC)</sub> +4		
6(0)	t <sub>su(SOMI-SPCH)M</sub>	Setup time, SPISOMI before SPICLK high (clock polarity = 1)	t <sub>f(SPC)</sub> +4		ns
7 <sup>(5)</sup>	t <sub>h(SPCL-SOMI)M</sub>	Hold time, SPISOMI data valid after SPICLK low (clock polarity = 0)	10		
7(0)	t <sub>h(SPCH-SOMI)M</sub>	Hold time, SPISOMI data valid after SPICLK high (clock polarity = 1)	10		
8(5)(6)		Setup time CS active until SPICLK high (clock polarity = 0)	(C2TDELAY+CSHOLD+ 2)*t <sub>c(VCLK)</sub> - t <sub>f(SPICS)</sub> + t <sub>r(SPICLK)</sub> -21	$ \begin{array}{c} (\text{C2TDELAY+CSHOLD+} \\ 2)^* t_{\text{C(VCLK)}} - t_{\text{f(SPICS)}} + \\ t_{\text{r(SPICLK)}} + 6 \end{array} $	
8(4)(4)	<sup>†</sup> C2TDELAY	Setup time CS active until SPICLK low (clock polarity = 1)	(C2TDELAY+CSHOLD+ 2)*t <sub>c(VCLK)</sub> - t <sub>f(SPICS)</sub> + t <sub>f(SPICLK)</sub> -21	(C2TDELAY+CSHOLD+ 2)*t <sub>c(VCLK)</sub> - t <sub>f(SPICS)</sub> + t <sub>f(SPICLK)</sub> +6	
9 <sup>(5)(6)</sup>		Hold time SPICLK low until CS inactive (clock polarity = 0)	$\begin{array}{c} 0.5^{\star}t_{c(SPC)M} \\ + T2CDELAY^{\star}t_{c(VCLK)} + \\ t_{c(VCLK)} - t_{f(SPICLK)} + \\ t_{r(SPICS)} - 4 \end{array}$	$\begin{array}{c} 0.5^*t_{c(SPC)M} \\ + T2CDELAY^*t_{c(VCLK)} + \\ t_{c(VCLK)} - t_{f(SPICLK)} + \\ t_{r(SPICS)} + 17 \end{array}$	
	<sup>t</sup> T2CDELAY	Hold time SPICLK high until CS inactive (clock polarity = 1)	$\begin{array}{c} 0.5^*t_{c(\text{SPC})\text{M}} \\ +\text{T2CDELAY*}t_{c(\text{VCLK})} + \\ t_{c(\text{VCLK})} - t_{r(\text{SPICLK})} + \\ t_{r(\text{SPICS})} - 4 \end{array}$	$\begin{array}{c} 0.5^*t_{c(\text{SPC})M} \\ + T2\text{CDELAY}^*t_{c(\text{VCLK})} + \\ t_{c(\text{VCLK})} - t_{r(\text{SPICLK})} + \\ t_{r(\text{SPICS})} + 17 \end{array}$	
10	t <sub>SPIENA</sub>	SPIENAn sample point	C2TDELAY * $t_{c(VCLK)}$ - $t_{f(SPICS)}$ -25	C2TDELAY * t <sub>c(VCLK)</sub>	ns

The MASTER bit (SPIGCR1.0) is set and the CLOCK PHASE bit (SPIFMTx.16) is clear.

<sup>(2)</sup>  $t_{c(VCLK)}$  = interface clock cycle time = 1 /  $f_{(VCLK)}$ .

For rise and fall timings, see Table 5-13.

When the SPI is in Master mode, the following must be true:

For PS values from 1 to 255: t ≥ (PS +1)t<sub>c(VCLK)</sub> ≥ 90 ns, where PS is the prescale value set in the SPIFMTx.[15:8] register bits.
 For PS values of 0: t<sub>c(SPC)M</sub> = 2t<sub>c(VCLK)</sub> ≥ 90 ns.
 The active edge of the SPInCLK signal referenced is controlled by the CLOCK POLARITY bit (SPIFMTx.17).
 C2TDELAY and T2CDELAY are programmed in the SPIDELAY register.



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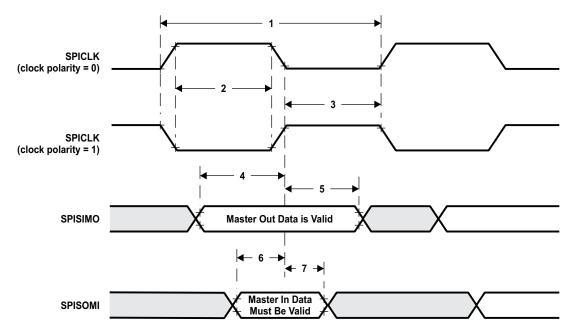


Figure 5-11. SPI Master Mode External Timing (CLOCK PHASE = 0)

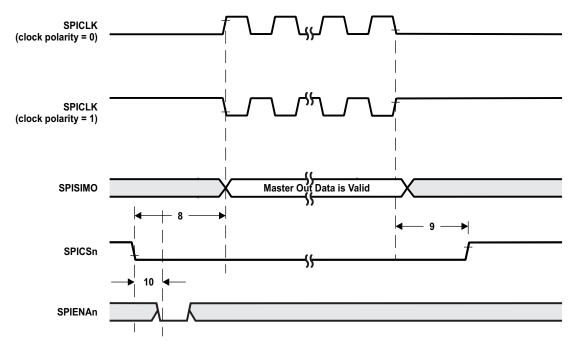


Figure 5-12. SPI Master Mode Chip Select Timing (CLOCK PHASE = 0)



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# Table 5-17. SPIn Master Mode External Timing Parameters (CLOCK PHASE = 1, SPInCLK = output, SPInSIMO = output, and SPInSOMI = input) $^{(1)(2)(3)}$

(see Figure 5-13 and Figure 5-14)

NO.			MIN	MAX	UNIT
1	t <sub>c(SPC)M</sub>	Cycle time, SPICLK <sup>(4)</sup>	90	256t <sub>c(VCLK)</sub>	
2 <sup>(5)</sup>	t <sub>w(SPCH)M</sub>	Pulse duration, SPICLK high (clock polarity = 0)	0.5t <sub>c(SPC)M</sub> - t <sub>r</sub> -8	$0.5t_{c(SPC)M} + 5$	
	t <sub>w(SPCL)M</sub>	Pulse duration, SPICLK low (clock polarity = 1)	0.5t <sub>c(SPC)M</sub> - t <sub>f</sub> -8	$0.5t_{c(SPC)M} + 5$	
3 <sup>(5)</sup>	t <sub>w(SPCL)M</sub>	Pulse duration, SPICLK low (clock polarity = 0)	0.5t <sub>c(SPC)M</sub> - t <sub>f</sub> -8	$0.5t_{c(SPC)M} + 5$	
	t <sub>w(SPCH)M</sub>	Pulse duration, SPICLK high (clock polarity = 1)	$0.5t_{c(SPC)M}$ - $t_r$ -8	$0.5t_{c(SPC)M} + 5$	
4 (5)	t <sub>v(SIMO-SPCH)M</sub>	Valid time, SPISIMO data valid before SPICLK high (clock polarity = 0)	0.5t <sub>c(SPC)M</sub> - 10		
4 <sup>(5)</sup>	t <sub>v(SIMO-SPCL)M</sub>	Valid time, SPISIMO data valid before SPICLK low (clock polarity = 1)	0.5t <sub>c(SPC)M</sub> - 10		
5 <sup>(5)</sup>	t <sub>v(SPCH-SIMO)M</sub>	Valid time, SPISIMO data valid after SPICLK high (clock polarity = 0)	0.5t <sub>c(SPC)M</sub> - t <sub>r(SPC)</sub> -5		
<b>3</b> <sup>(-)</sup>	t <sub>v(SPCL-SIMO)M</sub>	Valid time, SPISIMO data valid after SPICLK low (clock polarity = 1)	$0.5t_{c(SPC)M}$ - $t_{f(SPC)}$ -5		
6 <sup>(5)</sup>	t <sub>su(SOMI-SPCH)M</sub>	Setup time, SPISOMI before SPICLK high (clock polarity = 0)	t <sub>r(SPC)</sub> +4		ns
6(4)	t <sub>su(SOMI-SPCL)M</sub>	Setup time, SPISOMI before SPICLK low (clock polarity = 1)	t <sub>f(SPC)</sub> +4		113
7 <sup>(5)</sup>	t <sub>h(SPCH-SOMI)M</sub>	Hold time, SPISOMI data valid after SPICLK high (clock polarity = 0)	10		
<i>(</i> )	t <sub>h(SPCL-SOMI)M</sub>	Hold time, SPISOMI data valid after SPICLK low (clock polarity = 1)	10		
8 <sup>(5)(6)</sup>	t	Setup time CS active until SPICLK high (clock polarity = 0)	$\begin{array}{c} 0.5t_{c(SPC)M}\text{+}(C2TDELAY\\ +CSHOLD\text{+}2)^*t_{c(VCLK)}\text{-}\\ t_{f(SPICS)}\text{+}t_{r(SPICLK)}\text{-}21 \end{array}$	$\begin{array}{c} 0.5t_{\text{c(SPC)M}} + (\text{C2TDELAY} \\ + \text{CSHOLD} + 2)^*t_{\text{c(VCLK)}} - \\ t_{\text{f(SPICS)}} + t_{\text{r(SPICLK)}} + 6 \end{array}$	
<b>6</b> . / · /	<sup>†</sup> C2TDELAY	Setup time CS active until SPICLK low (clock polarity = 1)	$\begin{array}{c} 0.5t_{\text{C(SPC)M}}\text{+}(\text{C2TDELAY}\\ \text{+CSHOLD+2})^*t_{\text{C(VCLK)}}\text{-}\\ t_{\text{f(SPICS)}}\text{+}t_{\text{f(SPICLK)}}\text{-}21 \end{array}$	$\begin{array}{c} 0.5t_{c(SPC)M} + (C2TDELAY\\ + CSHOLD + 2)^*t_{c(VCLK)} -\\ t_{f(SPICS)} + t_{f(SPICLK)} + 6 \end{array}$	
g(5)(6)		Hold time SPICLK low CS until inactive (clock polarity = 0)	$\begin{array}{c} \text{T2CDELAY}^* t_{\text{c(VCLK)}} + \\ t_{\text{c(VCLK)}} - t_{\text{f(SPICLK)}} + \\ t_{\text{r(SPICS)}} - 4 \end{array}$	$\begin{array}{l} \text{T2CDELAY*} t_{c(\text{VCLK})} + \\ t_{c(\text{VCLK})} - t_{f(\text{SPICLK})} + \\ t_{r(\text{SPICS})} + 17 \end{array}$	
3(-7(-3)	<sup>†</sup> T2CDELAY	Hold time SPICLK high until CS inactive (clock polarity = 1)	$ \begin{array}{c} T2CDELAY^* t_{C(VCLK)} + \\ t_{c(VCLK)} - t_{r(SPICLK)} + \\ t_{r(SPICS)} \text{-} \end{array} $	$\begin{array}{c} \text{T2CDELAY*} t_{\text{c(VCLK)}} + \\ t_{\text{c(VCLK)}} - t_{\text{r(SPICLK)}} + \\ t_{\text{r(SPICS)}} + 17 \end{array}$	
10 <sup>(7)</sup>	t <sub>SPIENA</sub>	SPIENAn Sample Point	C2TDELAY * $t_{c(VCLK)}$ - $t_{f(SPICS)}$ -25	C2TDELAY * t <sub>c(VCLK)</sub>	ns

- The MASTER bit (SPIGCR1.0) is set and the CLOCK PHASE bit (SPIFMTx.16) is clear.
- $t_{c(VCLK)}$  = interface clock cycle time = 1 /  $f_{(VCLK)}$ . (2)
- For rise and fall timings, see Table 5-13.
- When the SPI is in Master mode, the following must be true:
  - For PS values from 1 to 255:  $t \ge (PS + 1)t_{c(VCLK)} \ge 90$  ns, where PS is the prescale value set in the SPIFMTx.[15:8] register bits.
- For PS values of 0: t<sub>c(SPC)M</sub> = 2t<sub>c(VCLK)</sub> ≥ 90 ns.

  The active edge of the SPInCLK signal reference is controlled by the CLOCK POLARITY bit (SPIFMTx.17).

  C2TDELAY and T2CDELAY is programmed in the SPIDELAY register.
- C2TDELAY and T2CDELAY is programmed in the SPIDELAY register.



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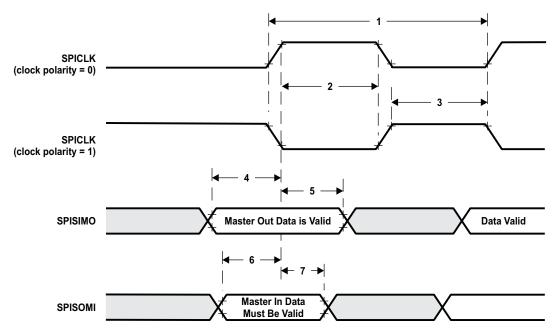


Figure 5-13. SPI Master Mode External Timing (CLOCK PHASE = 1)

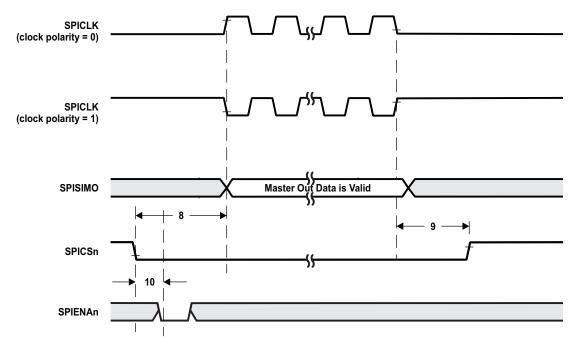


Figure 5-14. SPI Master Mode Chip Select timing (CLOCK PHASE = 1)



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#### 5.4 **SPIn Slave Mode Timing Parameters**

# Table 5-18. SPIn Slave Mode External Timing Parameters (CLOCK PHASE = 0, SPInCLK = input, SPInSIMO = input, and SPInSOMI = output) $^{(1)(2)(3)(4)}$

(see Figure 5-15 and Figure 5-16)

NO.			MIN	MAX	UNIT
1	t <sub>c(SPC)S</sub>	Cycle time, SPInCLK <sup>(5)</sup>	90		
2 <sup>(6)</sup>	t <sub>w(SPCH)S</sub>	Pulse duration, SPInCLK high (clock polarity = 0)	30		
2(")	t <sub>w(SPCL)S</sub>	Pulse duration, SPInCLK low (clock polarity = 1)	30		
3(6)	t <sub>w(SPCL)S</sub>	Pulse duration, SPInCLK low (clock polarity = 0)	30		
3(")	t <sub>w(SPCH)S</sub>	Pulse duration, SPInCLK high (clock polarity = 1)	30		
4 <sup>(6)</sup>	t <sub>d(SPCH-SOMI)S</sub>	Delay time, SPInCLK high to SPInSOMI valid (clock polarity = 0)		$t_{rf(SOMI)}$ +17	
4`′	t <sub>d(SPCL-SOMI)S</sub>	Delay time, SPInCLK low to SPInSOMI valid (clock polarity = 1)		$t_{rf(SOMI)}$ +17	
5 <sup>(6)</sup>	t <sub>v(SPCH-SOMI)S</sub>	Valid time, SPInSOMI data valid after SPInCLK high (clock polarity = 0)	0		ns
5(3)	t <sub>v(SPCL-SOMI)S</sub>	Valid time, SPInSOMI data valid after SPInCLK low (clock polarity = 1)	0		
6 <sup>(6)</sup>	t <sub>su(SIMO-SPCL)S</sub>	Setup time, SPInSIMO before SPInCLK low (clock polarity = 0)	5		
6(-)	t <sub>su(SIMO-SPCH)S</sub>	Setup time, SPInSIMO before SPInCLK high (clock polarity = 1)	5		
7 <sup>(6)</sup>	t <sub>v(SPCL-SIMO)S</sub>	Valid time, SPInSIMO data valid after SPInCLK low (clock polarity = 0)	6		
7(0)	t <sub>v(SPCH-SIMO)S</sub>	Valid time, SPInSIMO data valid after SPInCLK high (clock polarity = 1)	6		
8	t <sub>d</sub> (SPCL-SENAH)S	Delay time, SPIENAn high after last SPICLK low (clock polarity = 0)	1.5t <sub>c(VCLK)</sub>	$\begin{array}{c} 2.5t_{c(VCLK)} + \\ t_{r(ENAn)} + 20 \end{array}$	no
8	t <sub>d</sub> (SPCH-SENAH)S	Delay time, SPIENAn high after last SPICLK high (clock polarity = 1)	1.5t <sub>c(VCLK)</sub>	$\begin{array}{c} 2.5t_{c(VCLK)} + \\ t_{r(ENAn)} + 20 \end{array}$	ns
9	t <sub>d</sub> (SCSL-SENAL)S	Delay time, SPIENAn low after SPICSn low (if new data has been written to the SPI buffer)	t <sub>f(ENAn)</sub>	$t_{c(VCLK)}$ + $t_{f(ENAn)}$ +18	ns

- The MASTER bit (SPIGCR1.0) is clear and the CLOCK PHASE bit (SPIFMTx.16) is clear.
- When the SPI is in Slave mode, the following must be true:  $t_{c(SPC)S} > 2t_{c(VCLK)}$  and  $t_{c(SPC)S} \ge 90$  ns.
- For rise and fall timings, see Table 5-13.
- To list and fall trainings, see 1 state 5 1.  $t_{c(VCLK)}$  = interface clock cycle time = 1 /  $f_{(VCLK)}$ . When the SPI is in Slave mode, the following must be true:  $t_{w(SPCL)S} > t_{c(VCLK)}$ ,  $t_{w(SPCL)S} > t_{c(VCLK)}$  ns and  $t_{w(SPCH)S} > t_{c(VCLK)}$  ns and  $t_{w(SP$ (5)
- The active edge of the SPInCLK signal referenced is controlled by the CLOCK POLARITY bit (SPIFMTx.17).



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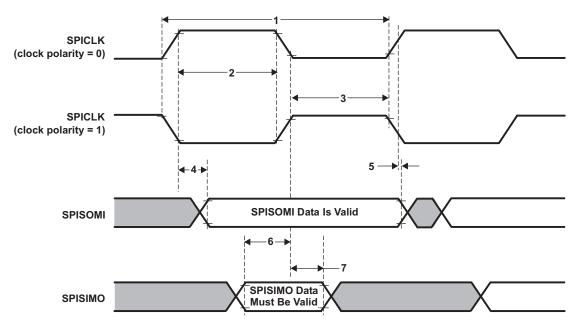


Figure 5-15. SPI Slave Mode External Timing (CLOCK PHASE = 0)

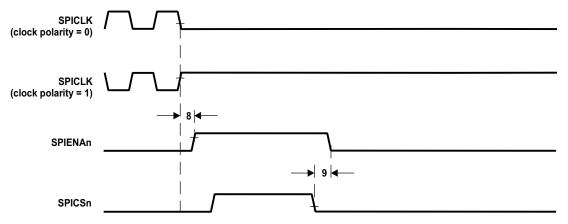


Figure 5-16. SPI Slave Mode Enable Timing (CLOCK PHASE = 0)

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# Table 5-19. SPIn Slave Mode External Timing Parameters (CLOCK PHASE = 1, SPInCLK = input, SPInSIMO = input, and SPInSOMI = output) $^{(1)(2)(3)(4)}$

(see Figure 5-17 and Figure 5-18)

NO.			MIN	MAX	UNIT
1	t <sub>c(SPC)S</sub>	Cycle time, SPInCLK <sup>(5)</sup>	90		
2(6)	t <sub>w(SPCH)S</sub>	Pulse duration, SPInCLK high (clock polarity = 0)	30		
2(-)	t <sub>w(SPCL)S</sub>	Pulse duration, SPInCLK low (clock polarity = 1)	30		
3 <sup>(6)</sup>	t <sub>w(SPCL)S</sub>	Pulse duration, SPInCLK low (clock polarity = 0)	30		
3\'/	t <sub>w(SPCH)S</sub>	Pulse duration, SPInCLK high (clock polarity = 1)	30		
4(6)	t <sub>d(SPCH-SOMI)S</sub>	Delay time, SPInSOMI data valid after SPInCLK high (clock polarity = 0)		$t_{rf(SOMI)}$ +17	
4(*)	t <sub>d(SPCL-SOMI)S</sub>	Delay time, SPInSOMI data valid after SPInCLK low (clock polarity = 1)		t <sub>rf(SOMI)</sub> +17	
5 <sup>(6)</sup>	t <sub>v(SOMI-SPCH)S</sub>	Valid time, SPInCLK high after SPInSOMI data valid (clock polarity = 0)	0		ns
5(3)	t <sub>v</sub> (SOMI-SPCL)S	Valid time, SPInCLK low after SPInSOMI data valid (clock polarity = 1)	0		
6 <sup>(6)</sup>	t <sub>su(SIMO-SPCH)S</sub>	Setup time, SPInSIMO before SPInCLK high (clock polarity = 0)	5		
6(-)	t <sub>su(SIMO-SPCL)S</sub>	Setup time, SPInSIMO before SPInCLK low (clock polarity = 1)	5		
7 <sup>(6)</sup>	t <sub>v(SPCH-SIMO)S</sub>	Valid time, SPInSIMO data valid after SPInCLK high (clock polarity = 0)	6		
107	t <sub>v(SPCL-SIMO)S</sub>	Valid time, SPInSIMO data valid after SPInCLK low (clock polarity = 1)	6		
0	t <sub>d</sub> (SPCH-SENAH)S	Delay time, SPIENAn high after last SPICLK high (clock polarity = 0)	1.5t <sub>c(VCLK)</sub>	$2.5t_{c(VCLK)}$ + $tr(ENAn)$ + $20$	
8	t <sub>d</sub> (SPCL-SENAH)S	Delay time, SPIENAn high after last SPICLK low (clock polarity = 1)	1.5t <sub>c(VCLK)</sub>	$\begin{array}{c} 2.5t_{c(VCLK)} + \\ t_{r(ENAn)} + 20 \end{array}$	ns
9	t <sub>d</sub> (SCSL-SENAL)S	Delay time, SPIENAn low after SPICSn low (if new data has been written to the SPI buffer)	t <sub>f(ENAn)</sub>	$t_{c(VCLK)}$ + $t_{f(ENAn)}$ +18	ns
10	t <sub>d(SCSL-SOMI)S</sub>	Delay time, SOMI valid after SPICSn low (if new data has been written to the SPI buffer)	t <sub>c(VCLK)</sub>	2t <sub>c(VCLK)</sub> + t <sub>rf(SOMI)</sub> +17	ns

- (1) The MASTER bit (SPInCTRL2.3) is cleared and the CLOCK PHASE bit (SPInCTRL2.0) is set.
- When the SPI is in Slave mode, the following must be true:  $t_{c(SPC)S} > 2t_{c(VCLK)}$  and  $t_{c(SPC)S} \ge 90$  ns.
- For rise and fall timings, see Table 5-13.
- To the and rail trillings, see Table 5 To.  $t_{c(VCLK)}$  = interface clock cycle time = 1 /f(VCLK). When the SPI is in Slave mode, the following must be true:  $t_{w(SPCL)S} > t_{c(VCLK)}$ ,  $t_{w(SPCL)S} > t_{c(VCLK)}$  as and  $t_{w(SPCH)S} > t_{c(VCLK)}$  as and  $t_{w(SPCH)S} > t_{c(VCLK)}$ .
- The active edge of the SPInCLK signal referenced is controlled by the CLOCK POLARITY bit (SPIFMTx.17).



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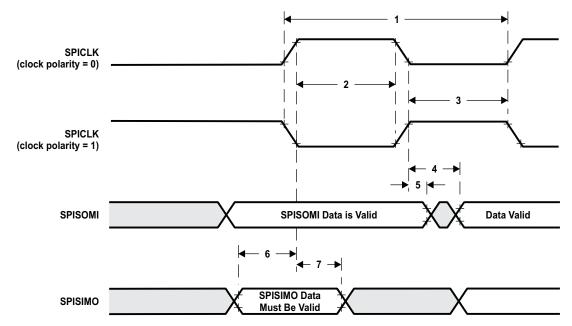


Figure 5-17. SPI Slave Mode External Timing (CLOCK PHASE = 1)

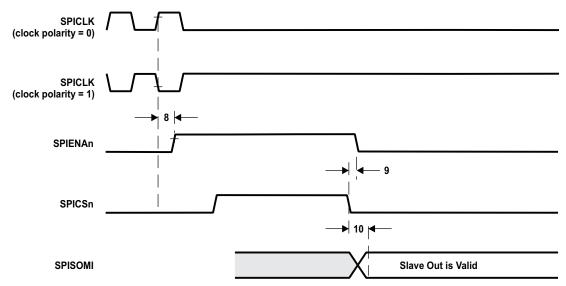


Figure 5-18. SPI Slave Mode Enable Timing (CLOCK PHASE = 1)

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#### 5.5 **CAN Controller (DCANn) Mode Timings**

### Table 5-20. Dynamic Characteristics for the CANnSTX and CANnSRX Pins

	PARAMETER	MIN MAX	UNIT
t <sub>d(CANnSTX)</sub>	Delay time, transmit shift register to CANnSTX pin <sup>(1)</sup>	15	ns
t <sub>d(CANnSRX)</sub>	Delay time, CANnSRX pin to receive shift register	6	ns

<sup>(1)</sup> These values do not include rise/fall times of the output buffer.

#### **High-End Timer (HET) Timings**

## Table 5-21. Dynamic Characteristics for the HET Pins

	PARAMETER	MIN	MAX	UNIT
t <sub>opw</sub> (HET)	Output pulse width, this is the minimum pulse width that can be generated <sup>(1)</sup>	1/f <sub>(VCLK2)</sub>		ns
t <sub>ipw</sub> (HET)	Input pulse width, this is the minimum pulse width that can be captured (2)	1/f <sub>(VCLK2)</sub>		ns

 $<sup>\</sup>begin{array}{ll} \text{(1)} & t_{opw}(\text{HET})_{min} = \text{HRP}_{(min)} = \text{hr}_{(min)} \ / \ \text{VCLK2}. \\ \text{(2)} & t_{ipw}(\text{HET}) = \text{LRP}_{(min)} = \text{hr}_{(min)} \ ^* \ \text{lr}_{(min)} \ / \ \text{VCLK2}. \end{array}$ 

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#### 5.7 Multi-Buffered A-to-D Converter (MibADC)

The multi-buffered A-to-D converter (MibADC) has a separate power bus for its analog circuitry that enhances the A-to-D performance by preventing digital switching noise on the logic circuitry which could be present on VSS and VCC from coupling into the A-to-D analog stage. All A-to-D specifications are given with respect to ADREFLO unless otherwise noted.

Resolution 10 bits (1024 values)

Monotonic Assured

Output conversion code 00h to 3FFh [00 for  $V_{AI} \le AD_{REFLO}$ ; 3FF for  $V_{AI} \ge A_{DREFHI}$ ]

### Table 5-22. MibADC Recommended Operating Conditions<sup>(1)</sup>

		MIN	MAX	UNIT
AD <sub>REFHI</sub>	A-to-D high -voltage reference source	3.0	V <sub>CCAD</sub>	V
AD <sub>REFLO</sub>	A-to-D low-voltage reference source	V <sub>SSAD</sub>	0.3	V
V <sub>AI</sub>	Analog input voltage	AD <sub>REFLO</sub>	AD <sub>REFHI</sub>	V
I <sub>AIC</sub>	Analog input clamp current <sup>(2)</sup> (V <sub>AI</sub> < V <sub>SSAD</sub> - 0.3 or V <sub>AI</sub> > V <sub>CCAD</sub> + 0.3)	- 2	2	mA

<sup>(1)</sup> For  $V_{CCAD}$  and  $V_{SSAD}$  recommended operating conditions, see .

# Table 5-23. MibADC Operating Characteristics Over Full Range of Recommended Operating Conditions<sup>(1)</sup>

PARAMETER		DESCRIPTION/CONDITIONS	MIN	NOM	MAX	UNIT	
R <sub>mux</sub>	Analog input mux on-resistance	See Figure 5-19		125	1.5K	Ω	
$R_{\text{samp}}$	ADC sample switch on-resistance	See Figure 5-19		150	1.5K	Ω	
C <sub>mux</sub>	Input mux capacitance	See Figure 5-19			16	pF	
C <sub>samp</sub>	ADC sample capacitance	See Figure 5-19			8	pF	
I <sub>AIL</sub>	Analog input leakage current	Input leakage per ADC input pin	-200		200	nA	
I <sub>ADREFHI</sub>	AD <sub>REFHI</sub> input current	AD <sub>REFHI</sub> = 3.6 V, AD <sub>REFLO</sub> = V <sub>SSAD</sub>			5	mA	
CR	Conversion range over which specified accuracy is maintained	ADREFHI - ADREFLO	3		3.6	V	
E <sub>DNL</sub>	Differential non-linearity error	Difference between the actual step width and the ideal value (see Figure 5-20).			± 2	LSB	
E <sub>INL</sub>	Integral non-linearity error	Maximum deviation from the best straight line through the MibADC. MibADC transfer characteristics, excluding the quantization error (see Figure 5-21).			± 2	LSB	
E <sub>TOT</sub>	Total error/Absolute accuracy	Maximum value of the difference between an analog value and the ideal midstep value (see Figure 5-22).			± 2	LSB	

(1)  $1 - LSB = (AD_{REFHI} - AD_{REFLO})/2^{10}$  for the MibADC.

<sup>(2)</sup> Input currents into any ADC input channel outside the specified limits could affect conversion results of other channels.

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#### MibADC Input Model 5.7.1

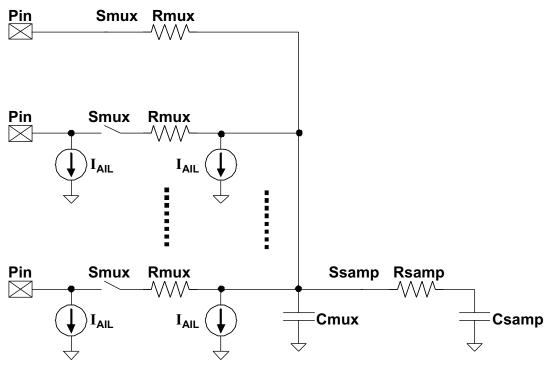


Figure 5-19. MibADC Input Equivalent Circuit

#### Table 5-24. Multi-Buffer ADC Timing Requirements

	PARAMETER	MIN	NOM	MAX	UNIT
t <sub>c(ADCLK)</sub>	Cycle time, MibADC clock	0.05			μs
t <sub>d(SH)</sub>	Delay time, sample and hold time	1			μs
t <sub>d(C)</sub>	Delay time, conversion time	0.55			μs
t <sub>d(SHC)</sub> <sup>(1)</sup>	Delay time, total sample/hold and conversion time	1.55			μs

(1) This is the minimum sample/hold and conversion time that can be achieved. These parameters are dependent on many factors.

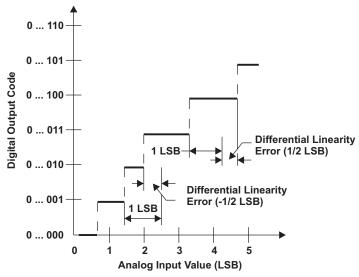


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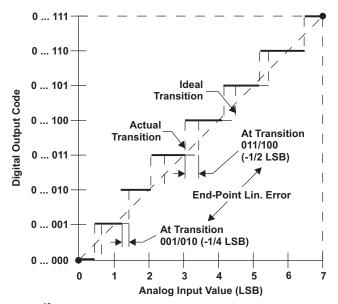
The differential non-linearity error shown in Figure 5-20 (sometimes referred to as differential linearity) is the difference between an actual step width and the ideal value of 1 LSB.



 $1 \text{ LSB} = (AD_{REFHI} - AD_{REFLO})/2^{10}$ 

Figure 5-20. Differential Non-linearity (DNL)

The integral non-linearity error shown in Figure 5-21 (sometimes referred to as linearity error) is the deviation of the values on the actual transfer function from a straight line.



 $1 LSB = (AD_{REFHI} - AD_{REFLO})/2^{10}$ 

Figure 5-21. Integral Non-linearity (INL) Error

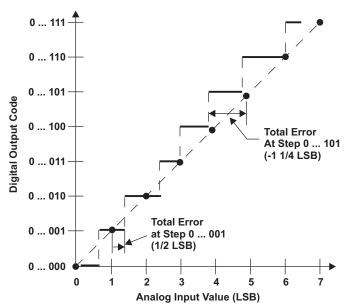
The absolute accuracy or total error of an MibADC as shown in Figure 5-22 is the maximum value of the difference between an analog value and the ideal midstep value.

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A. 1 LSB =  $(AD_{REFHI\mu} - AD_{REFLO})/2^{10}$ 

Figure 5-22. Absolute Accuracy (Total) Error



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# 6 Revision History

This data sheet revision history highlights the technical changes made to the device or the datasheet.

Date	Additions, Deletions, And Modifications	Revision			
August 2011	Added descriptions for the ENZ pin.	Α			
December 2011	Corrected number of GIO pins available from 8 to 4 in Device Characteristics table				
	Updated LBIST section to include support for STCCLK = HCLK				
	Added additional detail about MBIST cycle counts				
	Operating Conditions and electrical specs upated with characterized values				
	Added upper limit to Vreg ramp specification				
Removed support for low power modes					
	Added TEST Pin Glitch Filter Timing specification				
	Added note about back to back write/erase cycling in the EEPROM emulation bank.				
	Added FMPLL validated settings table				
January 2012	Updated programming times in the Flash Timings table.				
Corrected programming word size from 16-bit to 32-bit in the Flash Timings table to accurate reflect the default FSM configuration.		С			
	Added assumed use case and qualification standards for EEPROM emulation use in an application in the Flash Timings table.				

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#### **Mechanical Data**

#### 7.1 **Thermal Data**

Table 7-1 shows the thermal resistance characteristics for the PQFP - PZ mechanical packages.

**Table 7-1. Thermal Resistance Characteristics** (S-PQFP Package) [PZ]

PARAMETER	°C/W
$R_{\theta JA}$	48
R <sub>eJC</sub>	5

#### 7.2 **Packaging Information**

The following packaging information reflects the most current released data available for the designated device(s). This data is subject to change without notice and without revision of this document.



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PACKAGE OPTION ADDENDUM

23-Sep-2014

#### PACKAGING INFORMATION

Orderable Device	Status	Package Type	-		-	Eco Plan	Lead/Ball Finish	MSL Peak Temp	Op Temp (°C)	Device Marking	Samples
	(1)		Drawing		Qty	(2)	(6)	(3)		(4/5)	
S4MF03107SPZQQ1	ACTIVE	LQFP	PZ	100	90	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR	-40 to 125	TMS470 MF03107SPZQQ1	Samples
S4MF04207SPZQQ1	ACTIVE	LQFP	PZ	100	90	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR	-40 to 125	TMS470 MF04207SPZQQ1	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design. PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): Tl's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between

the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight

in homogeneous material)

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "-" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead/Ball Finish - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width

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PACKAGE OPTION ADDENDUM

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In no event shall T1's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

Addendum-Page 2

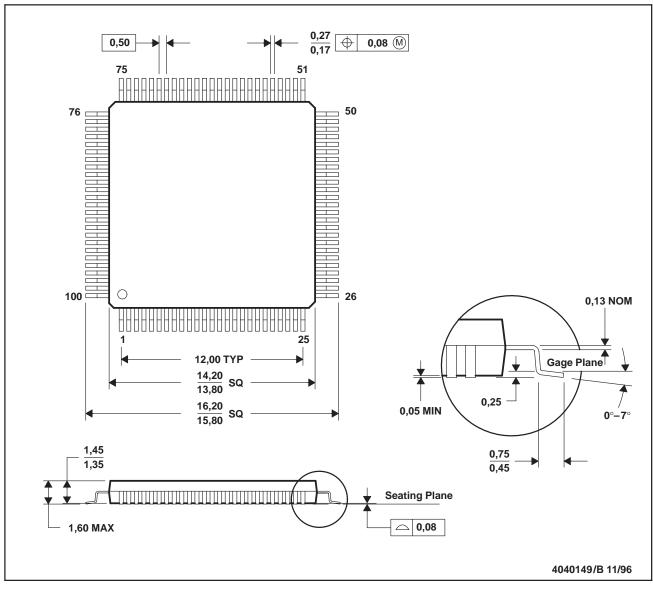
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### **MECHANICAL DATA**

MTQF013A - OCTOBER 1994 - REVISED DECEMBER 1996

#### PZ (S-PQFP-G100)

#### PLASTIC QUAD FLATPACK



NOTES: A. All linear dimensions are in millimeters.

B. This drawing is subject to change without notice.

C. Falls within JEDEC MS-026





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