

## Excellent Integrated System Limited

Stocking Distributor

Click to view price, real time Inventory, Delivery & Lifecycle Information:

[Texas Instruments](#)  
[LM4940TS/NOPB](#)

For any questions, you can email us directly:

[sales@integrated-circuit.com](mailto:sales@integrated-circuit.com)



**LM4940**

www.ti.com

SNAS219C –OCTOBER 2003–REVISED MAY 2013

## LM4940 Boomer® Audio Power Amplifier Series 6W Stereo Audio Power Amplifier

Check for Samples: [LM4940](#)

### FEATURES

- Click and Pop Circuitry Eliminates Noise During Turn-On and Turn-Off Transitions
- Low Current, Active-Low Shutdown Mode
- Low Quiescent Current
- Stereo 6W Output,  $R_L = 4\Omega$
- Short Circuit Protection
- Unity-Gain Stable
- External Gain Configuration Capability

### APPLICATIONS

- Flat Panel Monitors
- Flat Panel TVs
- Computer Sound Cards

### KEY SPECIFICATIONS

- Quiescent Power Supply Current: 40mA (max) (SE)
- $P_{OUT}$   
 $V_{DD} = 14.4V$ ,  $R_L = 4\Omega$ , 10% THD+N: 6 W (typ)
- Shutdown Current: 40 $\mu$ A (typ)

### DESCRIPTION

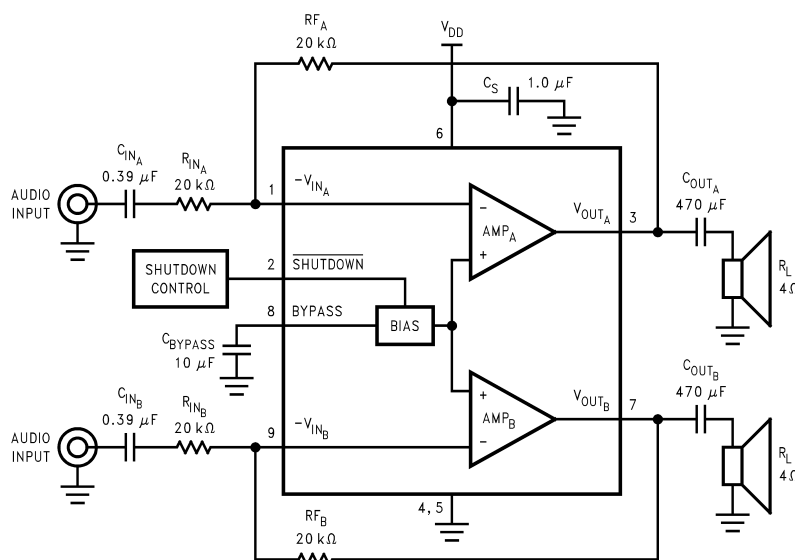
The LM4940 is a dual audio power amplifier primarily designed for demanding applications in flat panel monitors and TVs. It is capable of delivering 6 watts per channel to a 4 $\Omega$  load with less than 10% THD+N while operating on a 14.4V<sub>DC</sub> power supply.

Boomer™ audio power amplifiers were designed specifically to provide high quality output power with a minimal amount of external components. The LM4940 does not require bootstrap capacitors or snubber circuits. Therefore, it is ideally suited for display applications requiring high power and minimal size.

The LM4940 features a low-power consumption active-low shutdown mode. Additionally, the LM4940 features an internal thermal shutdown protection mechanism along with short circuit protection.

The LM4940 contains advanced pop and click circuitry that eliminates noises which would otherwise occur during turn-on and turn-off transitions.

The LM4940 is a unity-gain stable and can be configured by external gain-setting resistors.



**Figure 1. Typical Stereo Audio Amplifier Application Circuit**



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

Boomer is a trademark of Texas Instruments, Inc.

All other trademarks are the property of their respective owners.

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of the Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

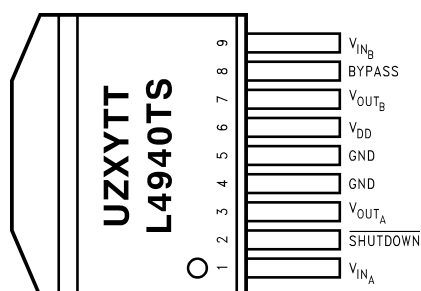
Copyright © 2003–2013, Texas Instruments Incorporated

## LM4940

SNAS219C –OCTOBER 2003–REVISED MAY 2013

[www.ti.com](http://www.ti.com)

### CONNECTION DIAGRAM



U = Wafer Fab Code, Z = Assembly Plant Code, XY = Date Code, TT = Die Traceability

**Plastic Package, DDPAK  
Top View  
See Package Number KTW**



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

## ABSOLUTE MAXIMUM RATINGS<sup>(1)(2)(3)</sup>

Supply Voltage (pin 6, referenced to GND, pins 4 and 5)		18.0V
Storage Temperature		–65°C to +150°C
Input Voltage	Pins 3 and 7	–0.3V to $V_{DD} + 0.3V$
	Pins 1, 2, 8, and 9	–0.3V to 9.5V
Power Dissipation <sup>(4)</sup>		Internally limited
ESD Susceptibility <sup>(5)</sup>		2000V
ESD Susceptibility <sup>(6)</sup>		200V
Junction Temperature		150°C
Thermal Resistance	$\theta_{JC}$ (KTW)	4°C/W
	$\theta_{JA}$ (KTW) <sup>(4)</sup>	20°C/W
	$\theta_{JC}$ (NEC)	4°C/W
	$\theta_{JA}$ (NEC) <sup>(4)</sup>	20°C/W

- (1) All voltages are measured with respect to the GND pin, unless otherwise specified.
- (2) Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is functional, but do not ensure specific performance limits. Electrical Characteristics state DC and AC electrical specifications under particular test conditions which ensure performance limits. This assumes that the device is within the Operating Ratings. Specifications are not ensured for parameters where no limit is given, however, the typical value is a good indication of device performance.
- (3) If Military/Aerospace specified devices are required, please contact the Texas Instruments Sales Office/ Distributors for availability and specifications.
- (4) The maximum power dissipation must be derated at elevated temperatures and is dictated by  $T_{JMAX}$ ,  $\theta_{JA}$ , and the ambient temperature,  $T_A$ . The maximum allowable power dissipation is  $P_{DMAX} = (T_{JMAX} - T_A) / \theta_{JA}$  or the given in Absolute Maximum Ratings, whichever is lower. For the LM4940 typical application (shown in Figure 1) with  $V_{DD} = 12V$ ,  $R_L = 4\Omega$  stereo operation the total power dissipation is 3.65W.  $\theta_{JA} = 20^\circ C/W$  for both DPAK and TO–220 packages mounted to 16in<sup>2</sup> heatsink surface area.
- (5) Human body model, 100pF discharged through a 1.5k $\Omega$  resistor.
- (6) Machine Model, 220pF–240pF discharged through all pins.

## OPERATING RATINGS

Temperature Range ( $T_{MIN} \leq T_A \leq T_{MAX}$ )	$-40^\circ C \leq T_A \leq 85^\circ C$
Supply Voltage	$10V \leq V_{DD} \leq 16V$

## LM4940



SNAS219C –OCTOBER 2003–REVISED MAY 2013

www.ti.com

### ELECTRICAL CHARACTERISTICS $V_{DD} = 12V^{(1)(2)}$

The following specifications apply for  $V_{DD} = 12V$ ,  $A_V = 10$ ,  $R_L = 4\Omega$ ,  $f = 1kHz$  unless otherwise specified. Limits apply for  $T_A = 25^\circ C$ .

Symbol	Parameter	Conditions	LM4940		Units (Limits)
			Typical <sup>(3)</sup>	Limit <sup>(4)(5)</sup>	
$I_{DD}$	Quiescent Power Supply Current	$V_{IN} = 0V$ , $I_O = 0A$ , No Load	16	40	mA (max)
$I_{SD}$	Shutdown Current	$V_{SHUTDOWN} = GND^{(6)}$	40	100	$\mu A$ (max)
$V_{SDIH}$	Shutdown Voltage Input High			2.0 $V_{DD}/2$	V (min) V (max)
$V_{SDIL}$	Shutdown Voltage Input Low			0.4	V (max)
$P_O$	Output Power	Single Channel			W (min)
		THD+N = 1%	3.1	2.8	
		THD+N = 10%	4.2		
		$V_{DD} = 14.4V$ , THD+N = 10%	6.0		
THD+N	Total Harmonic Distortion + Noise	$P_O = 1W_{RMS}$ , $A_V = 10$ , $f = 1kHz$	0.15		%
$\epsilon_{OS}$	Output Noise	A-Weighted Filter, $V_{IN} = 0V$ , Input Referred	10		$\mu V$
$X_{TALK}$	Channel Separation	$P_O = 1W$	70		dB
PSRR	Power Supply Rejection Ratio	$V_{RIPPLE} = 200mV_{p-p}$ $f_{RIPPLE} = 1kHz$	56		dB

- (1) All voltages are measured with respect to the GND pin, unless otherwise specified.
- (2) Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is functional, but do not ensure specific performance limits. Electrical Characteristics state DC and AC electrical specifications under particular test conditions which ensure performance limits. This assumes that the device is within the Operating Ratings. Specifications are not ensured for parameters where no limit is given, however, the typical value is a good indication of device performance.
- (3) Typicals are measured at  $25^\circ C$  and represent the parametric norm.
- (4) Limits are specified to AOQL (Average Outgoing Quality Level).
- (5) Datasheet min/max specification limits are specified by design, test, or statistical analysis.
- (6) Shutdown current is measured in a normal room environment. The Shutdown pin should be driven as close as possible to GND for minimum shutdown current.

### TYPICAL APPLICATION

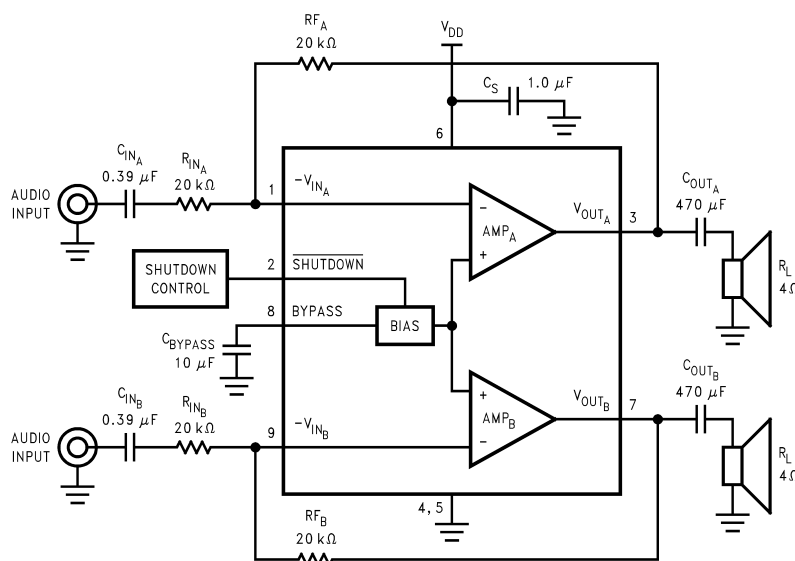


Figure 2. Typical Stereo Audio Amplifier Application Circuit

## EXTERNAL COMPONENTS DESCRIPTION

Refer to [Figure 1](#).

Components		Functional Description
1.	$R_{IN}$	This is the inverting input resistance that, along with $R_F$ , sets the closed-loop gain. Input resistance $R_{IN}$ and input capacitance $C_{IN}$ form a high pass filter. The filter's cutoff frequency is $f_C = 1/(2\pi R_{IN} C_{IN})$ .
2.	$C_{IN}$	This is the input coupling capacitor. It blocks DC voltage at the amplifier's inverting input. $C_{IN}$ and $R_{IN}$ create a highpass filter. The filter's cutoff frequency is $f_C = 1/(2\pi R_{IN} C_{IN})$ . Refer to the <a href="#">SELECTING EXTERNAL COMPONENTS</a> section for an explanation of determining $C_{IN}$ 's value.
3.	$R_F$	This is the feedback resistance that, along with $R_i$ , sets closed-loop gain.
4.	$C_S$	The supply bypass capacitor. Refer to the <a href="#">POWER SUPPLY BYPASSING</a> section for information about properly placing, and selecting the value of, this capacitor.
5.	$C_{BYPASS}$	This capacitor filters the half-supply voltage present on the BYPASS pin. Refer to the <a href="#">SELECTING EXTERNAL COMPONENTS</a> for information about properly placing, and selecting the value of, this capacitor.
6.	$C_{OUT}$	This is the output coupling capacitor. It blocks the nominal $V_{DD}/2$ voltage present at the output and prevents it from reaching the load. $C_{OUT}$ and $R_L$ form a high pass filter whose cutoff frequency is $f_C = 1/(2\pi R_L C_{OUT})$ . Refer to the <a href="#">SELECTING EXTERNAL COMPONENTS</a> section for an explanation of determining $C_{OUT}$ 's value.

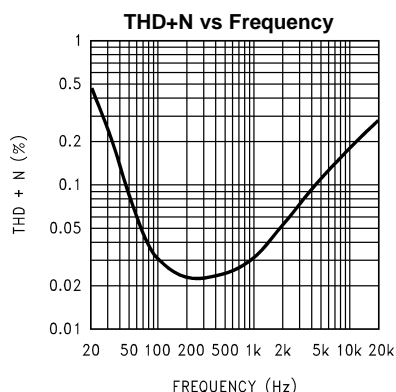
## LM4940



SNAS219C –OCTOBER 2003–REVISED MAY 2013

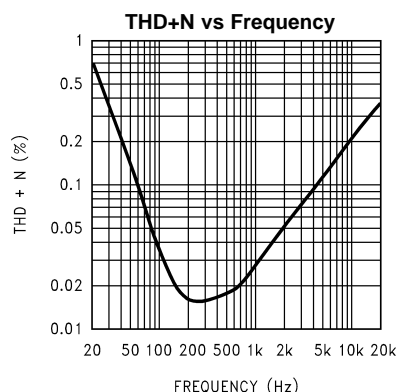
www.ti.com

### TYPICAL PERFORMANCE CHARACTERISTICS



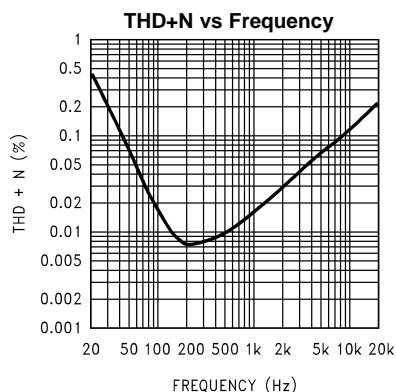
$V_{DD} = 12V$ ,  $R_L = 4\Omega$ , SE operation,  
both channels driven and loaded (average shown)  
 $P_{OUT} = 1W$ ,  $A_V = 1$

Figure 3.



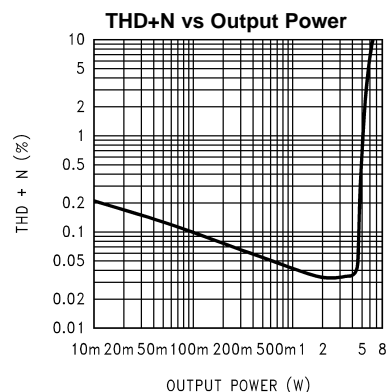
$V_{DD} = 12V$ ,  $R_L = 4\Omega$ , SE operation,  
both channels driven and loaded (average shown),  
 $P_{OUT} = 2.5W$ ,  $A_V = 1$

Figure 4.



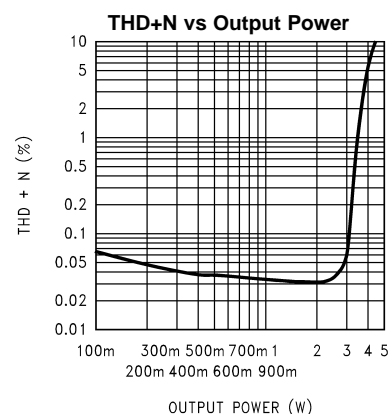
$V_{DD} = 12V$ ,  $R_L = 8\Omega$ , SE operation,  
both channels driven and loaded (average shown),  
 $P_{OUT} = 1W$ ,  $A_V = 1$

Figure 5.



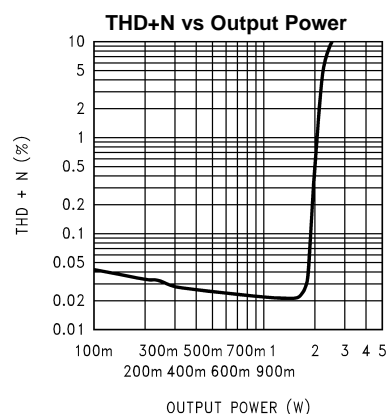
$V_{DD} = 14.4V$ ,  $R_L = 4\Omega$ , SE operation,  $A_V = 1$   
single channel driven/single channel measured,  
 $f_{IN} = 1kHz$

Figure 6.



$V_{DD} = 12V$ ,  $R_L = 4\Omega$ , SE operation,  $A_V = 1$   
single channel driven/single channel measured,  
 $f_{IN} = 1kHz$

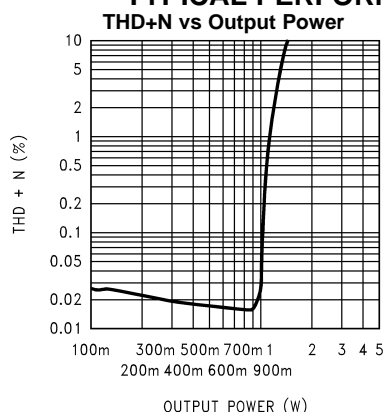
Figure 7.



$V_{DD} = 12V$ ,  $R_L = 8\Omega$ , SE operation,  $A_V = 1$   
single channel driven/single channel measured,  
 $f_{IN} = 1kHz$

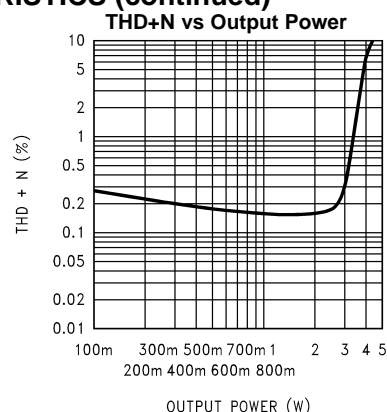
Figure 8.

## TYPICAL PERFORMANCE CHARACTERISTICS (continued)



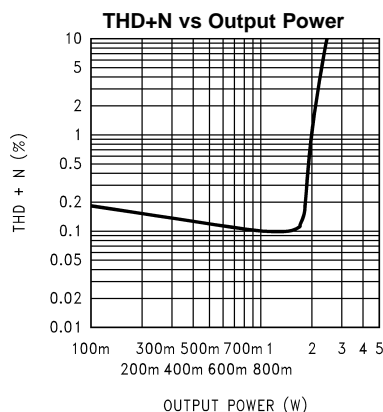
$V_{DD} = 12V$ ,  $R_L = 16\Omega$ , SE operation,  $A_V = 1$   
single channel driven/single channel measured,  
 $f_{IN} = 1kHz$

Figure 9.



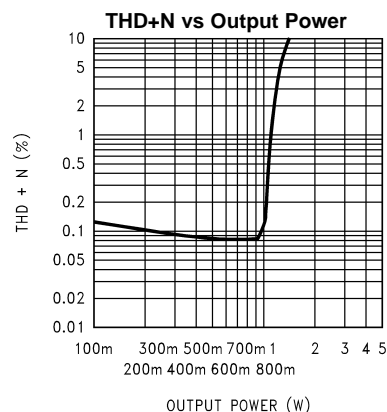
$V_{DD} = 12V$ ,  $R_L = 4\Omega$ , SE operation,  $A_V = 10$   
single channel driven/single channel measured,  
 $f_{IN} = 1kHz$

Figure 10.



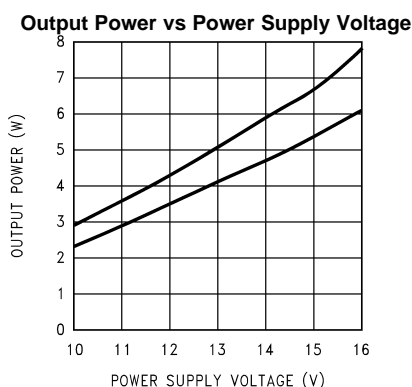
$V_{DD} = 12V$ ,  $R_L = 8\Omega$ , SE operation,  $A_V = 10$   
single channel driven/single channel measured,  
 $f_{IN} = 1kHz$

Figure 11.



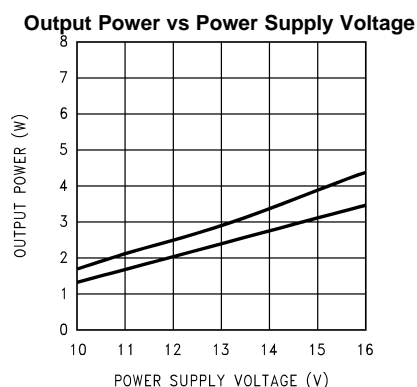
$V_{DD} = 12V$ ,  $R_L = 16\Omega$ , SE operation,  $A_V = 10$   
single channel driven/single channel measured,  
 $f_{IN} = 1kHz$

Figure 12.



$R_L = 4\Omega$ , SE operation,  $f_{IN} = 1kHz$ ,  
both channels driven and loaded (average shown),  
at (from top to bottom at 12V): THD+N = 10%,  
THD+N = 1%

Figure 13.



$R_L = 8\Omega$ , SE operation,  $f_{IN} = 1kHz$ ,  
both channels driven and loaded (average shown),  
at (from top to bottom at 12V): THD+N = 10%,  
THD+N = 1%

Figure 14.



## LM4940

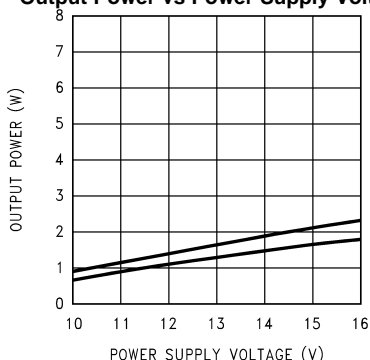


SNAS219C –OCTOBER 2003–REVISED MAY 2013

www.ti.com

### TYPICAL PERFORMANCE CHARACTERISTICS (continued)

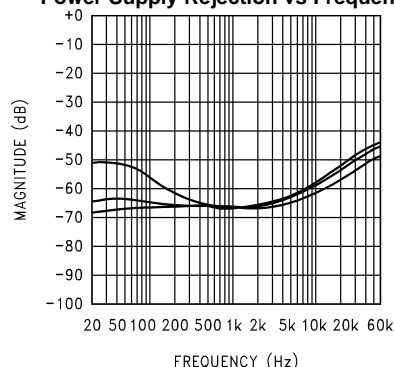
**Output Power vs Power Supply Voltage**



$R_L = 16\Omega$ , SE operation,  $f_{IN} = 1\text{kHz}$ ,  
both channels driven and loaded (average shown),  
at (from top to bottom at 12V): THD+N = 10%,  
THD+N = 1%

Figure 15.

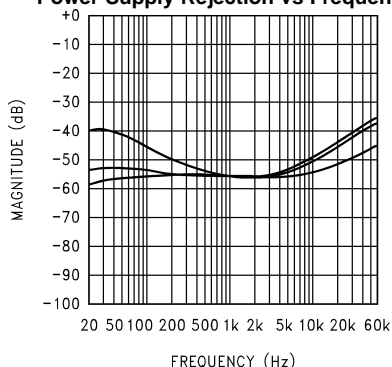
**Power Supply Rejection vs Frequency**



$V_{DD} = 12\text{V}$ ,  $R_L = 8\Omega$ , SE operation,  
 $V_{RIPPLE} = 200\text{mV}_{p-p}$ , at (from top to bottom at 60Hz):  
 $C_{BYPASS} = 1\mu\text{F}$ ,  $C_{BYPASS} = 4.7\mu\text{F}$ ,  $C_{BYPASS} = 10\mu\text{F}$

Figure 16.

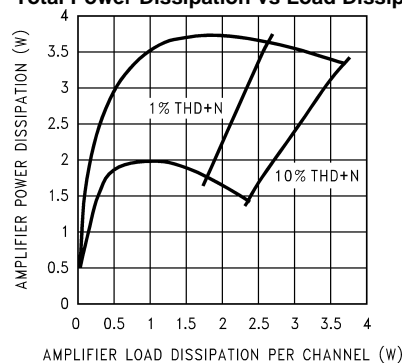
**Power Supply Rejection vs Frequency**



$V_{DD} = 12\text{V}$ ,  $R_L = 8\Omega$ , SE operation,  $V_{RIPPLE} = 200\text{mV}_{p-p}$ ,  
 $A_V = 10$ , at (from top to bottom at 60Hz):  
 $C_{BYPASS} = 1\mu\text{F}$ ,  $C_{BYPASS} = 4.7\mu\text{F}$ ,  $C_{BYPASS} = 10\mu\text{F}$

Figure 17.

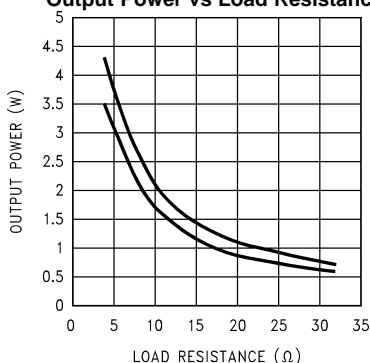
**Total Power Dissipation vs Load Dissipation**



$V_{DD} = 12\text{V}$ , SE operation,  $f_{IN} = 1\text{kHz}$ ,  
at (from top to bottom at 1W):  
 $R_L = 4\Omega$ ,  $R_L = 8\Omega$

Figure 18.

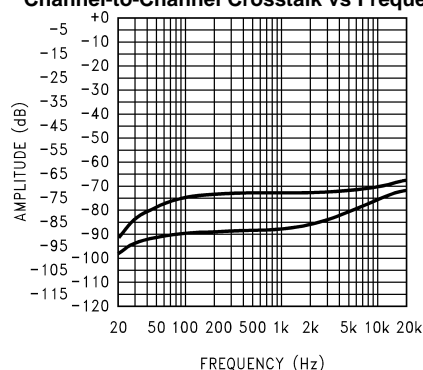
**Output Power vs Load Resistance**



$V_{DD} = 12\text{V}$ , SE operation,  $f_{IN} = 1\text{kHz}$ ,  
both channels driven and loaded,  
at (from top to bottom at 15Ω):  
THD+N = 10%, THD+N = 1%

Figure 19.

**Channel-to-Channel Crosstalk vs Frequency**

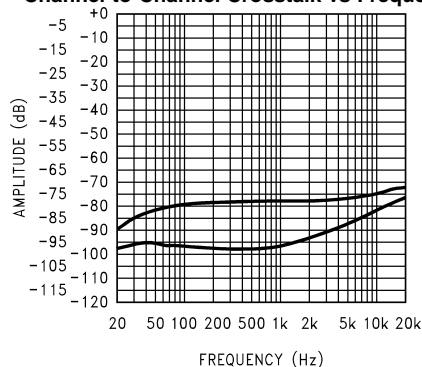


$V_{DD} = 12\text{V}$ ,  $R_L = 4\Omega$ ,  $P_{OUT} = 1\text{W}$ , SE operation,  
at (from top to bottom at 1kHz):  $V_{INB}$  driven,  
 $V_{OUTA}$  measured;  $V_{INA}$  driven,  $V_{OUTB}$  measured

Figure 20.

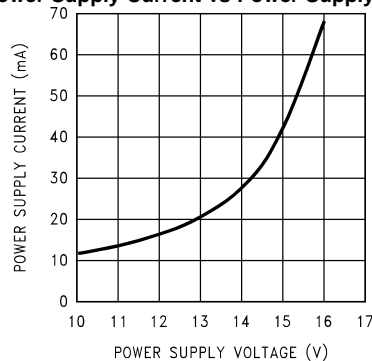
## TYPICAL PERFORMANCE CHARACTERISTICS (continued)

**Channel-to-Channel Crosstalk vs Frequency**



$V_{DD} = 12V$ ,  $R_L = 8\Omega$ ,  $P_{OUT} = 1W$ , SE operation,  
at (from top to bottom at 1kHz):  $V_{INB}$  driven,  
 $V_{OUTA}$  measured;  $V_{INA}$  driven,  $V_{OUTB}$  measured  
**Figure 21.**

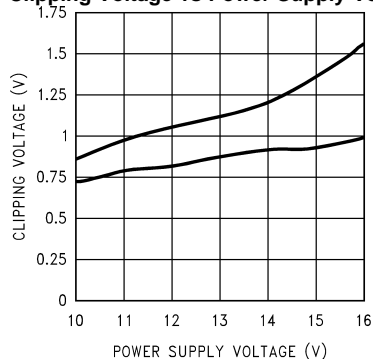
**Power Supply Current vs Power Supply Voltage**



$R_L = 4\Omega$ , SE operation  
 $V_{IN} = 0V$ ,  $R_{SOURCE} = 50\Omega$

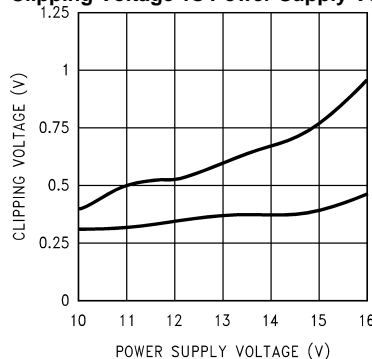
**Figure 22.**

**Clipping Voltage vs Power Supply Voltage**



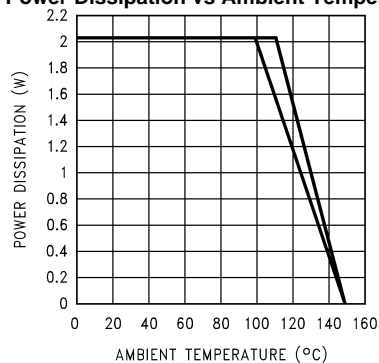
$R_L = 4\Omega$ , SE operation,  $f_{IN} = 1kHz$   
both channels driven and loaded,  
at (from top to bottom at 13V):  
negative signal swing, positive signal swing  
**Figure 23.**

**Clipping Voltage vs Power Supply Voltage**



$R_L = 8\Omega$ , SE operation,  $f_{IN} = 1kHz$   
both channels driven and loaded,  
at (from top to bottom at 13V):  
negative signal swing, positive signal swing  
**Figure 24.**

**Power Dissipation vs Ambient Temperature**



$V_{DD} = 12V$ ,  $R_L = 8\Omega$  (SE),  $f_{IN} = 1kHz$ ,  
(from top to bottom at 120°C):  
16in<sup>2</sup> copper plane heatsink area,  
8in<sup>2</sup> copper plane heatsink area

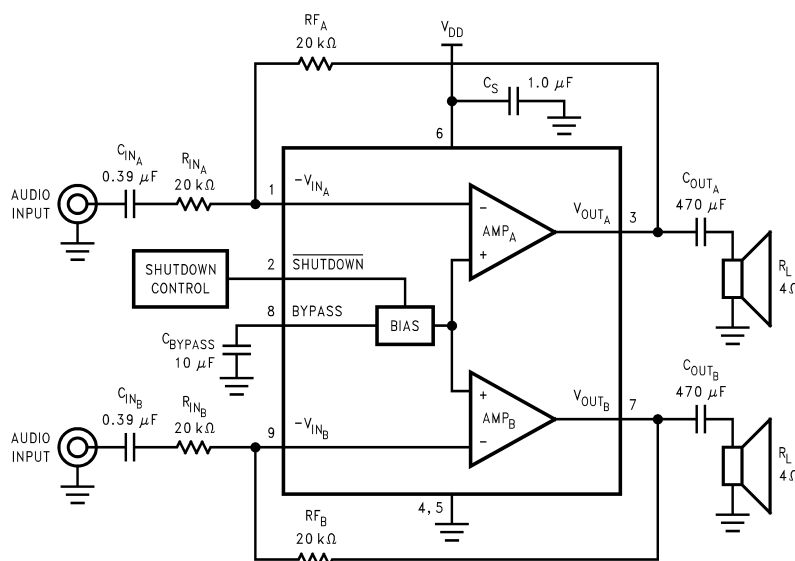
**Figure 25.**

## LM4940

SNAS219C –OCTOBER 2003–REVISED MAY 2013

www.ti.com

### APPLICATION INFORMATION



**Figure 26. Typical LM4940 Stereo Amplifier Application Circuit**

### HIGH VOLTAGE BOOMER WITH INCREASED OUTPUT POWER

Unlike previous 5V Boomer amplifiers, the LM4940 is designed to operate over a power supply voltages range of 10V to 15V. Operating on a 12V power supply, the LM4940 will deliver 3.1W per channel into 4Ω loads with no more than 1% THD+N.

### POWER DISSIPATION

Power dissipation is a major concern when designing a successful single-ended amplifier. Equation 1 states the maximum power dissipation point for a single-ended amplifier operating at a given supply voltage and driving a specified output load.

$$P_{\text{DMAX-SE}} = (V_{\text{DD}})^2 / (2\pi^2 R_L): \text{ Single Ended} \quad (1)$$

The LM4940's dissipation is twice the value given by Equation 1 when driving two SE loads. For a 12V supply and two 8Ω SE loads, the LM4940's dissipation is 1.82W.

The maximum power dissipation point (twice the value given by Equation 1 must not exceed the power dissipation given by Equation 2:

$$P_{\text{DMAX}}' = (T_{\text{JMAX}} - T_A) / \theta_{\text{JA}} \quad (2)$$

The LM4940's  $T_{\text{JMAX}} = 150^\circ\text{C}$ . In the KTW package, the LM4940's  $\theta_{\text{JA}}$  is  $20^\circ\text{C/W}$  when the metal tab is soldered to a copper plane of at least  $16\text{in}^2$ . This plane can be split between the top and bottom layers of a two-sided PCB. Connect the two layers together under the tab with a 5x5 array of vias. For the NEC package, use an external heatsink with a thermal impedance that is less than  $20^\circ\text{C/W}$ . At any given ambient temperature  $T_A$ , use Equation 3 to find the maximum internal power dissipation supported by the IC packaging. Rearranging Equation 3 and substituting  $P_{\text{DMAX}}$  for  $P_{\text{DMAX}}'$  results in Equation 4. This equation gives the maximum ambient temperature that still allows maximum stereo power dissipation without violating the LM4940's maximum junction temperature.

$$T_A = T_{JMAX} - P_{DMAX-SE} \theta_{JA} \quad (3)$$

For a typical application with a 12V power supply and two 4Ω SE loads, the maximum ambient temperature that allows maximum stereo power dissipation without exceeding the maximum junction temperature is approximately 113°C for the KTW package.

$$T_{JMAX} = P_{DMAX-SE} \theta_{JA} + T_A \quad (4)$$

Equation 4 gives the maximum junction temperature  $T_{JMAX}$ . If the result violates the LM4940's 150°C, reduce the maximum junction temperature by reducing the power supply voltage or increasing the load resistance. Further allowance should be made for increased ambient temperatures.

The above examples assume that a device is operating around the maximum power dissipation point. Since internal power dissipation is a function of output power, higher ambient temperatures are allowed as output power or duty cycle decreases.

If the result of Equation 3 is greater than that of Equation 4, then decrease the supply voltage, increase the load impedance, or reduce the ambient temperature. Further, ensure that speakers rated at a nominal 4Ω do not fall below 3Ω. If these measures are insufficient, a heat sink can be added to reduce  $\theta_{JA}$ . The heat sink can be created using additional copper area around the package, with connections to the ground pins, supply pin and amplifier output pins. Refer to the **TYPICAL PERFORMANCE CHARACTERISTICS** curves for power dissipation information at lower output power levels.

## POWER SUPPLY VOLTAGE LIMITS

Continuous proper operation is ensured by never exceeding the voltage applied to any pin, with respect to ground, as listed in the **Absolute Maximum Ratings** section.

## POWER SUPPLY BYPASSING

As with any power amplifier, proper supply bypassing is critical for low noise performance and high power supply rejection. Applications that employ a voltage regulator typically use a 10μF in parallel with a 0.1μF filter capacitors to stabilize the regulator's output, reduce noise on the supply line, and improve the supply's transient response. However, their presence does not eliminate the need for a local 1.0μF tantalum bypass capacitance connected between the LM4940's supply pins and ground. Do not substitute a ceramic capacitor for the tantalum. Doing so may cause oscillation. Keep the length of leads and traces that connect capacitors between the LM4940's power supply pin and ground as short as possible. Connecting a 10μF capacitor,  $C_{BYPASS}$ , between the BYPASS pin and ground improves the internal bias voltage's stability and improves the amplifier's PSRR. The PSRR improvements increase as the bypass pin capacitor value increases. Too large, however, increases turn-on time and can compromise the amplifier's click and pop performance. The selection of bypass capacitor values, especially  $C_{BYPASS}$ , depends on desired PSRR requirements, click and pop performance (as explained in the section, **SELECTING EXTERNAL COMPONENTS**), system cost, and size constraints.

## MICRO-POWER SHUTDOWN

The LM4940 features an active-low shutdown mode that disables the amplifier's bias circuitry, reducing the supply current to 40μA (typ). Connect SHUTDOWN to a voltage between 2V to  $V_{DD}/2$  for normal operation. Connect SHUTDOWN to GND to disable the device. A voltage that is greater than GND can increase shutdown current.

## LM4940

SNAS219C –OCTOBER 2003–REVISED MAY 2013

www.ti.com

### SELECTING EXTERNAL COMPONENTS

#### Input Capacitor Value Selection

Two quantities determine the value of the input coupling capacitor: the lowest audio frequency that requires amplification and desired output transient suppression.

As shown in [Figure 26](#), the input resistor ( $R_{IN}$ ) and the input capacitor ( $C_{IN}$ ) produce a high pass filter cutoff frequency that is found using [Equation 5](#).

$$f_c = 1/2\pi R_i C_i \quad (5)$$

As an example when using a speaker with a low frequency limit of 50Hz,  $C_i$ , using [Equation 5](#) is 0.159 $\mu$ F. The 0.39 $\mu$ F  $C_{INA}$  shown in [Figure 26](#) allows the LM4940 to drive high efficiency, full range speaker whose response extends below 30Hz.

#### Output Coupling Capacitor Value Selection

The capacitors  $C_{OUTA}$  and  $C_{OUTB}$  that block the  $V_{DD}/2$  output DC bias voltage and couple the output AC signal to the amplifier loads also determine low frequency response. These capacitors, combined with their respective loads create a highpass filter cutoff frequency. The frequency is also given by [Equation 5](#).

Using the same conditions as above, with a 4 $\Omega$  speaker,  $C_{OUT}$  is 820 $\mu$ F (nearest common value).

#### Bypass Capacitor Value

Besides minimizing the input capacitor size, careful consideration should be paid to value of  $C_{BYPASS}$ , the capacitor connected to the BYPASS pin. Since  $C_{BYPASS}$  determines how fast the LM4940 settles to quiescent operation, its value is critical when minimizing turn-on pops. The slower the LM4940's outputs ramp to their quiescent DC voltage (nominally  $V_{DD}/2$ ), the smaller the turn-on pop. Choosing  $C_{BYPASS}$  equal to 10 $\mu$ F along with a small value of  $C_{IN}$  (in the range of 0.1 $\mu$ F to 0.39 $\mu$ F), produces a click-less and pop-less shutdown function. As discussed above, choosing  $C_{IN}$  no larger than necessary for the desired bandwidth helps minimize clicks and pops.

### OPTIMIZING CLICK AND POP REDUCTION PERFORMANCE

The LM4940 contains circuitry that eliminates turn-on and shutdown transients ("clicks and pops"). For this discussion, turn-on refers to either applying the power supply voltage or when the micro-power shutdown mode is deactivated.

As the  $V_{DD}/2$  voltage present at the BYPASS pin ramps to its final value, the LM4940's internal amplifiers are configured as unity gain buffers and are disconnected from the  $AMP_A$  and  $AMP_B$  pins. An internal current source charges the capacitor connected between the BYPASS pin and GND in a controlled manner. Ideally, the input and outputs track the voltage applied to the BYPASS pin. The gain of the internal amplifiers remains unity until the voltage applied to the BYPASS pin.

The gain of the internal amplifiers remains unity until the voltage on the bypass pin reaches  $V_{DD}/2$ . As soon as the voltage on the bypass pin is stable, the device becomes fully operational and the amplifier outputs are reconnected to their respective output pins. Although the BYPASS pin current cannot be modified, changing the size of  $C_{BYPASS}$  alters the device's turn-on time. Here are some typical turn-on times for various values of  $C_{BYPASS}$ :

$C_B$ ( $\mu$ F)	$T_{ON}$ (ms)
1.0	120
2.2	120
4.7	200
10	440

In order eliminate "clicks and pops", all capacitors must be discharged before turn-on. Rapidly switching  $V_{DD}$  may not allow the capacitors to fully discharge, which may cause "clicks and pops".

There is a relationship between the value of  $C_{IN}$  and  $C_{BYPASS}$  that ensures minimum output transient when power is applied or the shutdown mode is deactivated. Best performance is achieved by setting the time constant created by  $C_{IN}$  and  $R_i + R_f$  to a value less than the turn-on time for a given value of  $C_{BYPASS}$  as shown in the table above.

## AUDIO POWER AMPLIFIER DESIGN

### Audio Amplifier Design: Driving 3W into a 4Ω load

The following are the desired operational parameters:

Power Output	3W <sub>RMS</sub>
Load Impedance	4Ω
Input Level	0.3V <sub>RMS</sub> (max)
Input Impedance	20kΩ
Bandwidth	100Hz–20kHz ± 0.25dB

The design begins by specifying the minimum supply voltage necessary to obtain the specified output power. One way to find the minimum supply voltage is to use the Output Power vs Power Supply Voltage curve in the [TYPICAL PERFORMANCE CHARACTERISTICS](#) section. Another way, using [Equation 6](#), is to calculate the peak output voltage necessary to achieve the desired output power for a given load impedance. To account for the amplifier's dropout voltage, two additional voltages, based on the Clipping Dropout Voltage vs Power Supply Voltage in the [TYPICAL PERFORMANCE CHARACTERISTICS](#) curves, must be added to the result obtained by [Equation 6](#). The result is [Equation 7](#).

$$V_{\text{opeak}} = \sqrt{2R_L P_O} \quad (6)$$

$$V_{DD} = V_{\text{OUTPEAK}} + V_{\text{ODTOP}} + V_{\text{ODBOT}} \quad (7)$$

The [Figure 13](#) graph for an 8Ω load indicates a minimum supply voltage of 11.8V. The commonly used 12V supply voltage easily meets this. The additional voltage creates the benefit of headroom, allowing the LM4940 to produce an output power of 3W without clipping or other audible distortion. The choice of supply voltage must also not create a situation that violates of maximum power dissipation as explained above in the [POWER DISSIPATION](#) section. After satisfying the LM4940's power dissipation requirements, the minimum differential gain needed to achieve 3W dissipation in a 4Ω BTL load is found using [Equation 8](#).

$$A_V \geq \sqrt{(P_O R_L)} / (V_{IN}) = V_{\text{orms}} / V_{\text{inrms}} \quad (8)$$

Thus, a minimum gain of 11.6 allows the LM4940's to reach full output swing and maintain low noise and THD+N performance. For this example, let  $A_V = 12$ . The amplifier's overall BTL gain is set using the input ( $R_{IN_A}$ ) and feedback (R) resistors of the first amplifier in the series BTL configuration. Additionally,  $A_{V-BTL}$  is twice the gain set by the first amplifier's  $R_{IN}$  and  $R_f$ . With the desired input impedance set at 20kΩ, the feedback resistor is found using [Equation 9](#).

$$R_f / R_{IN} = A_V \quad (9)$$

The value of  $R_f$  is 240kΩ. The nominal output power is 3W.

## LM4940

SNAS219C –OCTOBER 2003–REVISED MAY 2013

[www.ti.com](http://www.ti.com)

The last step in this design example is setting the amplifier's -3dB frequency bandwidth. To achieve the desired  $\pm 0.25\text{dB}$  pass band magnitude variation limit, the low frequency response must extend to at least one-fifth the lower bandwidth limit and the high frequency response must extend to at least five times the upper bandwidth limit. The gain variation for both response limits is  $0.17\text{dB}$ , well within the  $\pm 0.25\text{dB}$ -desired limit. The results are an

$$f_L = 100\text{Hz} / 5 = 20\text{Hz} \quad (10)$$

and

$$f_H = 20\text{kHz} \times 5 = 100\text{kHz} \quad (11)$$

As mentioned in the [SELECTING EXTERNAL COMPONENTS](#) section,  $R_{INA}$  and  $C_{INA}$ , as well as  $C_{OUT}$  and  $R_L$ , create a highpass filter that sets the amplifier's lower bandpass frequency limit. Find the coupling capacitor's value using [Equation 14](#).

$$C_{IN} = 1 / 2\pi R_{IN} f_L \quad (12)$$

The result is

$$1 / (2\pi \times 20\text{k}\Omega \times 20\text{Hz}) = 0.398\mu\text{F} = C_{IN} \quad (13)$$

and

$$1 / (2\pi \times 4\Omega \times 20\text{Hz}) = 1989\mu\text{F} = C_{OUT} \quad (14)$$

Use a  $0.39\mu\text{F}$  capacitor for  $C_{IN}$  and a  $2000\mu\text{F}$  capacitor for  $C_{OUT}$ , the closest standard values.

The product of the desired high frequency cutoff ( $100\text{kHz}$  in this example) and the differential gain  $A_V$ , determines the upper passband response limit. With  $A_V = 12$  and  $f_H = 100\text{kHz}$ , the closed-loop gain bandwidth product (GBWP) is  $1.2\text{MHz}$ . This is less than the LM4940's  $3.5\text{MHz}$  GBWP. With this margin, the amplifier can be used in designs that require more differential gain while avoiding performance restricting bandwidth limitations.

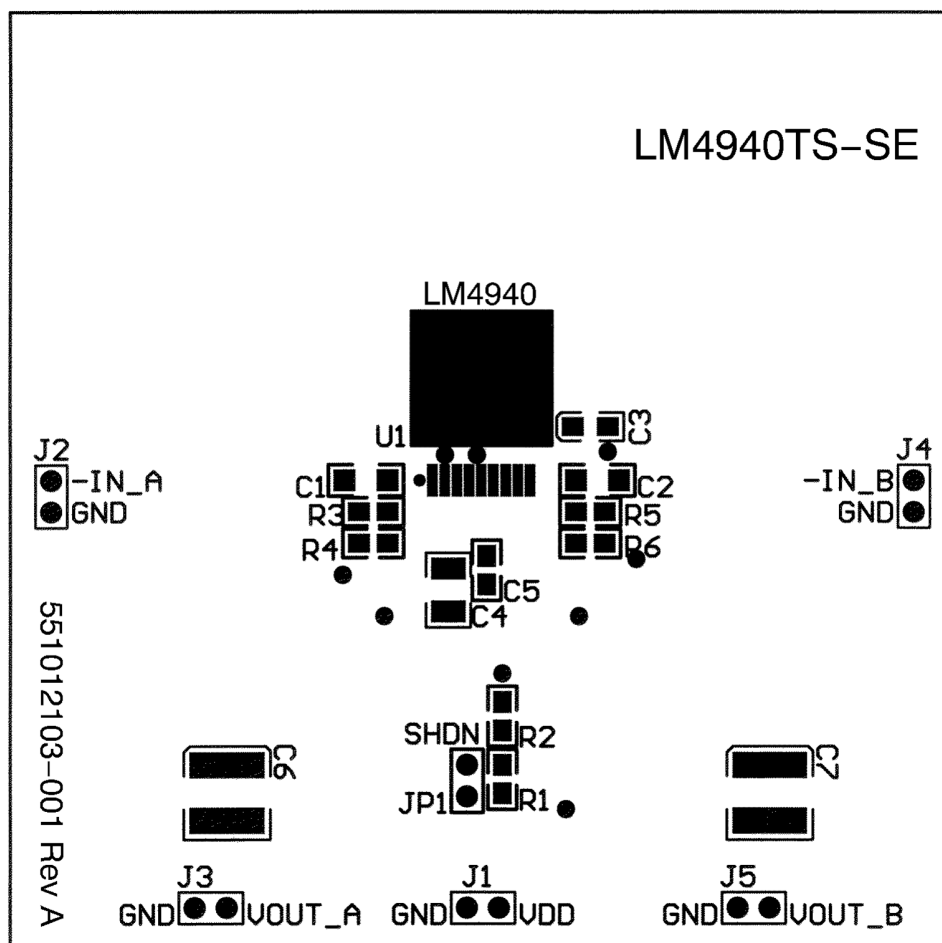
### RECOMMENDED PRINTED CIRCUIT BOARD LAYOUT

[Figure 27](#) through [Figure 29](#) show the recommended two-layer PC board layout that is optimized for the DDPACK-packaged LM4940 and associated external components. This circuit board is designed for use with an external  $12\text{V}$  supply and  $4\Omega$  (min) speakers.

This circuit board is easy to use. Apply  $12\text{V}$  and ground to the board's  $V_{DD}$  and GND pads, respectively. Connect a speaker between the board's  $OUT_A$  and  $OUT_B$  outputs and their respective GND terminals.



## Demonstration Board Layout



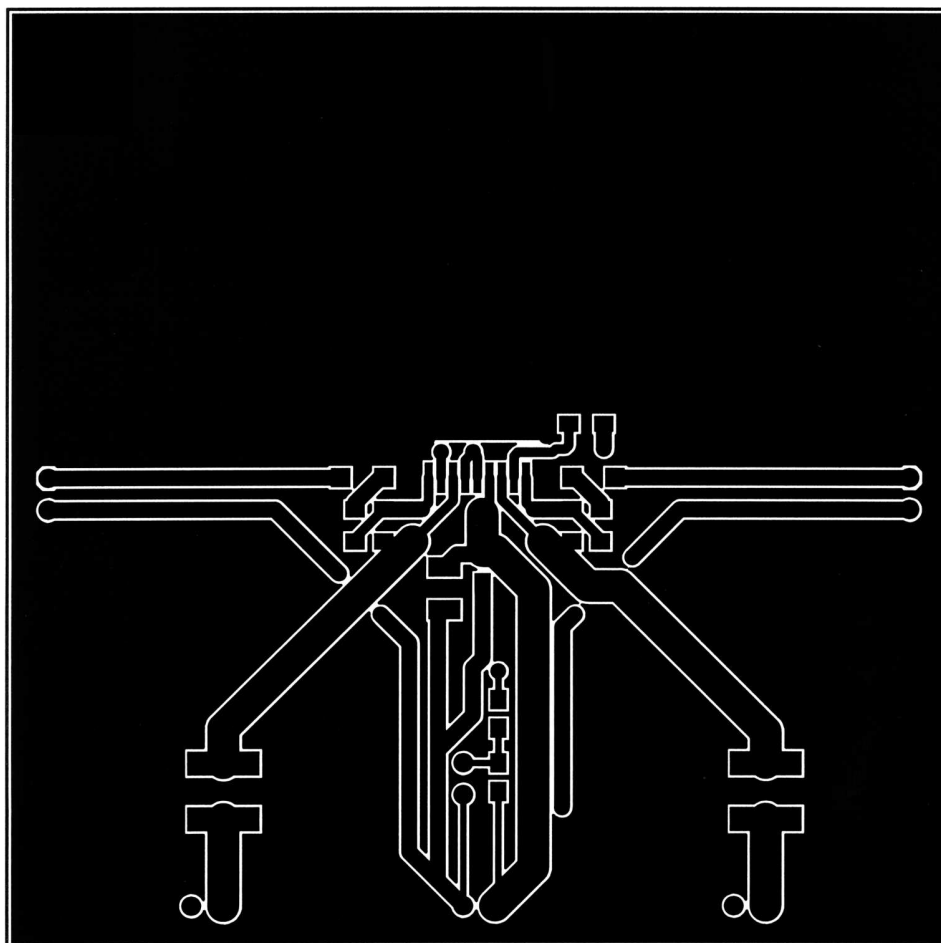
**Figure 27. Recommended KTW PCB Layout:  
Top Silkscreen**



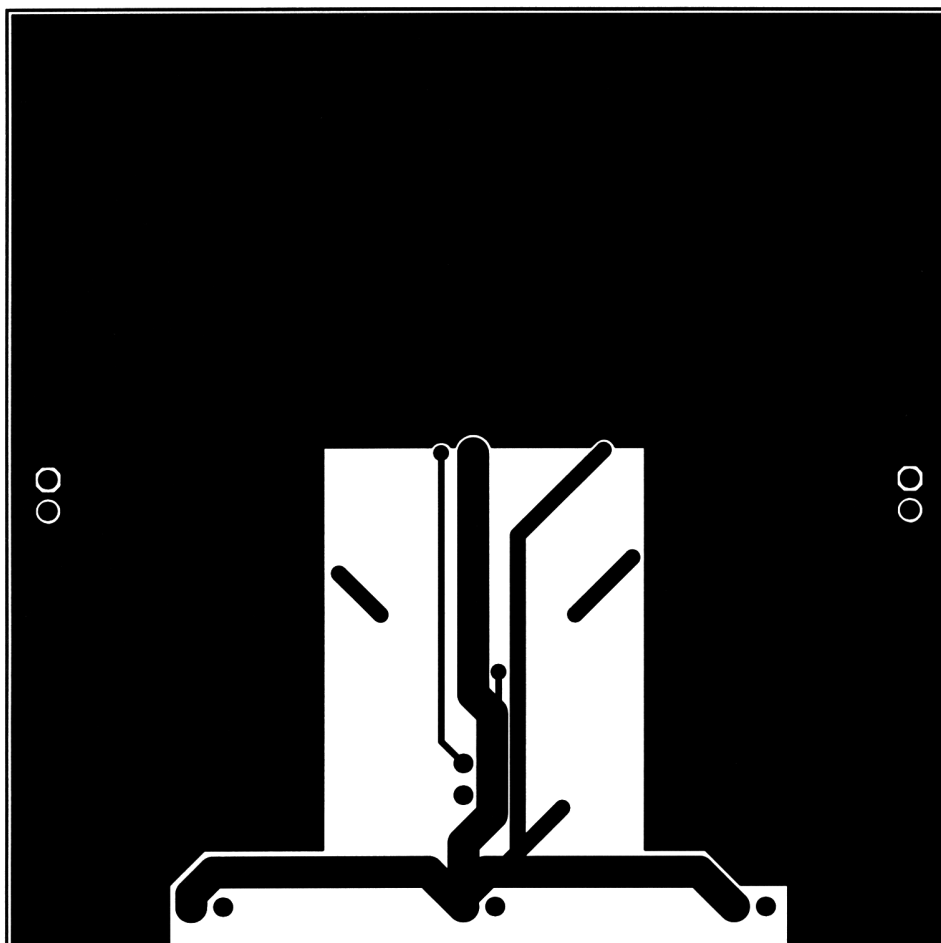
## LM4940

SNAS219C –OCTOBER 2003–REVISED MAY 2013

[www.ti.com](http://www.ti.com)



**Figure 28. Recommended KTW PCB Layout:  
Top Layer**



**Figure 29. Recommended KTW PCB Layout:  
Bottom Layer**

## LM4940

SNAS219C –OCTOBER 2003–REVISED MAY 2013

[www.ti.com](http://www.ti.com)

### REVISION HISTORY

Changes from Revision B (May 2013) to Revision C	Page
--	------

- |  |                    |
|--|--------------------|
| • Changed layout of National Data Sheet to TI format ..... | <a href="#">17</a> |
|--|--------------------|



PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish	MSL Peak Temp (3)	Op Temp (°C)	Top-Side Markings (4)	Samples
LM4940TS/NOPB	ACTIVE	DDPAK/ TO-263	KTW	9	45	Pb-Free (RoHS Exempt)	CU SN	Level-3-245C-168 HR	-40 to 85	L4940TS	<a href="#">Samples</a>
LM4940TSX/NOPB	ACTIVE	DDPAK/ TO-263	KTW	9	500	Pb-Free (RoHS Exempt)	CU SN	Level-3-245C-168 HR	-40 to 85	L4940TS	<a href="#">Samples</a>

(1) The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check <http://www.ti.com/productcontent> for the latest availability information and additional product content details.

**TBD:** The Pb-Free/Green conversion plan has not been defined.

**Pb-Free (RoHS):** TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

**Pb-Free (RoHS Exempt):** This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

**Green (RoHS & no Sb/Br):** TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) Multiple Top-Side Markings will be inside parentheses. Only one Top-Side Marking contained in parentheses and separated by a "-" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Top-Side Marking for that device.

**Important Information and Disclaimer:** The information provided on this page represents TI's knowledge and belief as of the date that it is provided. TI bases its knowledge and belief on information provided by third parties, and makes no representation or warranty as to the accuracy of such information. Efforts are underway to better integrate information from third parties. TI has taken and continues to take reasonable steps to provide representative and accurate information but may not have conducted destructive testing or chemical analysis on incoming materials and chemicals. TI and TI suppliers consider certain information to be proprietary, and thus CAS numbers and other limited information may not be available for release.

In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

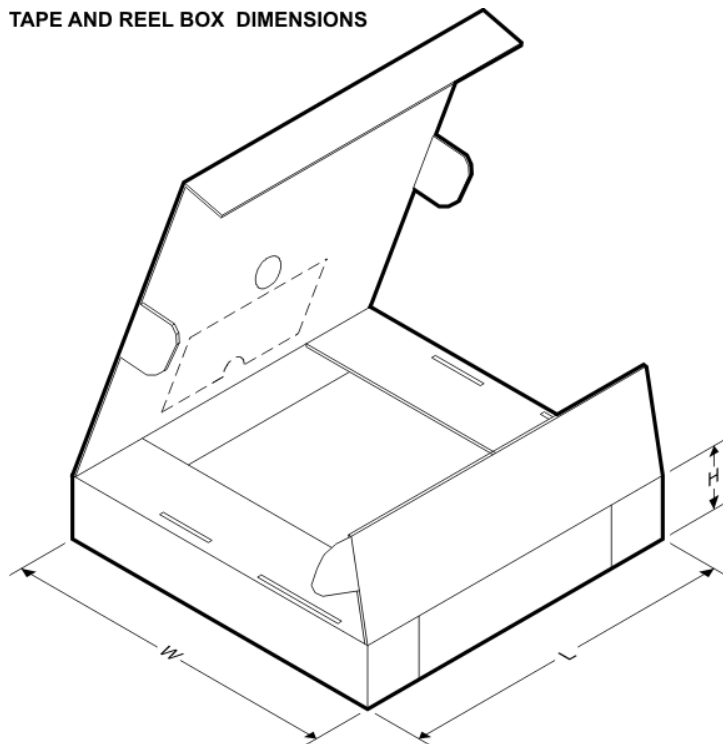
**TAPE AND REEL INFORMATION**



\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
LM4940TSX/NOPB	DDPAK/TO-263	KTW	9	500	330.0	24.4	10.75	14.85	5.0	16.0	24.0	Q2

**TAPE AND REEL BOX DIMENSIONS**

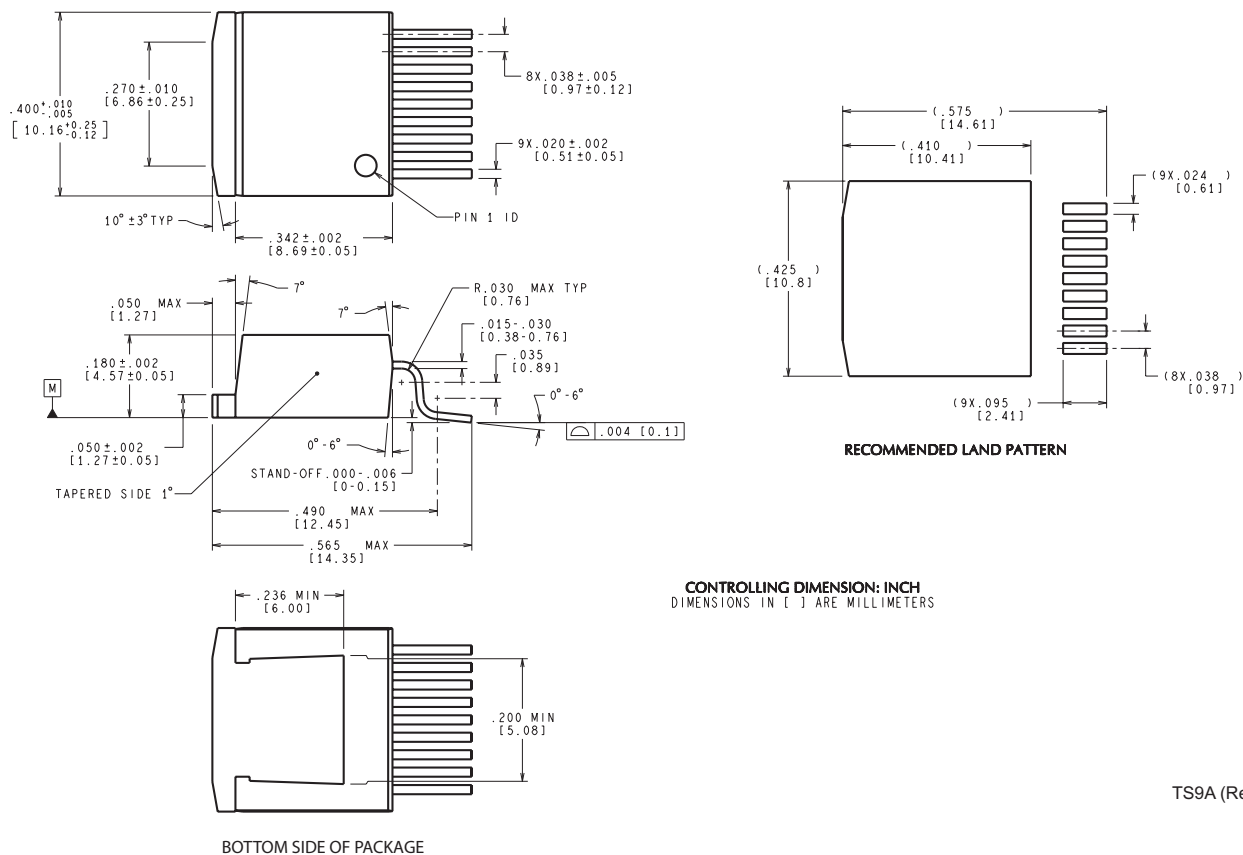


\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
LM4940TSX/NOPB	DDPAK/TO-263	KTW	9	500	367.0	367.0	45.0

## MECHANICAL DATA

KTW0009A



TS9A (Rev B)

## IMPORTANT NOTICE

Texas Instruments Incorporated and its subsidiaries (TI) reserve the right to make corrections, enhancements, improvements and other changes to its semiconductor products and services per JESD46, latest issue, and to discontinue any product or service per JESD48, latest issue. Buyers should obtain the latest relevant information before placing orders and should verify that such information is current and complete. All semiconductor products (also referred to herein as "components") are sold subject to TI's terms and conditions of sale supplied at the time of order acknowledgment.

TI warrants performance of its components to the specifications applicable at the time of sale, in accordance with the warranty in TI's terms and conditions of sale of semiconductor products. Testing and other quality control techniques are used to the extent TI deems necessary to support this warranty. Except where mandated by applicable law, testing of all parameters of each component is not necessarily performed.

TI assumes no liability for applications assistance or the design of Buyers' products. Buyers are responsible for their products and applications using TI components. To minimize the risks associated with Buyers' products and applications, Buyers should provide adequate design and operating safeguards.

TI does not warrant or represent that any license, either express or implied, is granted under any patent right, copyright, mask work right, or other intellectual property right relating to any combination, machine, or process in which TI components or services are used. Information published by TI regarding third-party products or services does not constitute a license to use such products or services or a warranty or endorsement thereof. Use of such information may require a license from a third party under the patents or other intellectual property of the third party, or a license from TI under the patents or other intellectual property of TI.

Reproduction of significant portions of TI information in TI data books or data sheets is permissible only if reproduction is without alteration and is accompanied by all associated warranties, conditions, limitations, and notices. TI is not responsible or liable for such altered documentation. Information of third parties may be subject to additional restrictions.

Resale of TI components or services with statements different from or beyond the parameters stated by TI for that component or service voids all express and any implied warranties for the associated TI component or service and is an unfair and deceptive business practice. TI is not responsible or liable for any such statements.

Buyer acknowledges and agrees that it is solely responsible for compliance with all legal, regulatory and safety-related requirements concerning its products, and any use of TI components in its applications, notwithstanding any applications-related information or support that may be provided by TI. Buyer represents and agrees that it has all the necessary expertise to create and implement safeguards which anticipate dangerous consequences of failures, monitor failures and their consequences, lessen the likelihood of failures that might cause harm and take appropriate remedial actions. Buyer will fully indemnify TI and its representatives against any damages arising out of the use of any TI components in safety-critical applications.

In some cases, TI components may be promoted specifically to facilitate safety-related applications. With such components, TI's goal is to help enable customers to design and create their own end-product solutions that meet applicable functional safety standards and requirements. Nonetheless, such components are subject to these terms.

No TI components are authorized for use in FDA Class III (or similar life-critical medical equipment) unless authorized officers of the parties have executed a special agreement specifically governing such use.

Only those TI components which TI has specifically designated as military grade or "enhanced plastic" are designed and intended for use in military/aerospace applications or environments. Buyer acknowledges and agrees that any military or aerospace use of TI components which have **not** been so designated is solely at the Buyer's risk, and that Buyer is solely responsible for compliance with all legal and regulatory requirements in connection with such use.

TI has specifically designated certain components as meeting ISO/TS16949 requirements, mainly for automotive use. In any case of use of non-designated products, TI will not be responsible for any failure to meet ISO/TS16949.

### Products

Audio	<a href="http://www.ti.com/audio">www.ti.com/audio</a>
Amplifiers	<a href="http://amplifier.ti.com">amplifier.ti.com</a>
Data Converters	<a href="http://dataconverter.ti.com">dataconverter.ti.com</a>
DLP® Products	<a href="http://www.dlp.com">www.dlp.com</a>
DSP	<a href="http://dsp.ti.com">dsp.ti.com</a>
Clocks and Timers	<a href="http://www.ti.com/clocks">www.ti.com/clocks</a>
Interface	<a href="http://interface.ti.com">interface.ti.com</a>
Logic	<a href="http://logic.ti.com">logic.ti.com</a>
Power Mgmt	<a href="http://power.ti.com">power.ti.com</a>
Microcontrollers	<a href="http://microcontroller.ti.com">microcontroller.ti.com</a>
RFID	<a href="http://www.ti-rfid.com">www.ti-rfid.com</a>
OMAP Applications Processors	<a href="http://www.ti.com/omap">www.ti.com/omap</a>
Wireless Connectivity	<a href="http://www.ti.com/wirelessconnectivity">www.ti.com/wirelessconnectivity</a>

### Applications

Automotive and Transportation	<a href="http://www.ti.com/automotive">www.ti.com/automotive</a>
Communications and Telecom	<a href="http://www.ti.com/communications">www.ti.com/communications</a>
Computers and Peripherals	<a href="http://www.ti.com/computers">www.ti.com/computers</a>
Consumer Electronics	<a href="http://www.ti.com/consumer-apps">www.ti.com/consumer-apps</a>
Energy and Lighting	<a href="http://www.ti.com/energy">www.ti.com/energy</a>
Industrial	<a href="http://www.ti.com/industrial">www.ti.com/industrial</a>
Medical	<a href="http://www.ti.com/medical">www.ti.com/medical</a>
Security	<a href="http://www.ti.com/security">www.ti.com/security</a>
Space, Avionics and Defense	<a href="http://www.ti.com/space-avionics-defense">www.ti.com/space-avionics-defense</a>
Video and Imaging	<a href="http://www.ti.com/video">www.ti.com/video</a>

### TI E2E Community

[e2e.ti.com](http://e2e.ti.com)